# LAMBDA ADVANCED ANALOG INC.

## ATR2800D Series Hybrid - High Reliability DC/DC Converters

## DESCRIPTION

The ATR2800D Series of DC/DC converters feature high power density and an extended temperature range for use in military and industrial applications. Designed to MIL-STD-704D input requirements, these devices have nominal 28VDC inputs with  $\pm$ 12V and  $\pm$ 15V dual outputs to satisfy a wide range of requirements. The circuit design incorporates a pulse width modulated single forward topology operating in the feed-forward mode at a nominal switching frequency of 550kHz. Input to output isolation is achieved through the use of transformers in the forward and feedback circuits.

The advanced feedback design provides fast loop response for superior line and load transient characteristics and offers greater reliability and radiation tolerance than devices incorporating optical feedback circuits.

Three standard temperature grades are offered with screening options. Refer to Part Number section. They can be provided in a standard plug-in package for PC mounting or in a flanged package for more severe environments.

These converters are manufactured in a facility certified to MIL-PRF-38534. All processes used to manufacture these converters have been qualified to enable Lambda Advanced Analog to deliver compliant devices. Four screening grades are available to satisfy a wide range of requirements. The CH grade converters are fully compliant to MIL-PRF-38534 class H. The HB grade converters are processed to full class H screening but do not have class H element evaluation as required by MIL-PRF-38534. Both grades are fully tested and operate over the full military temperature range without derating of output power. The ES version is a full temperature device without the full class H screening or element evaluation. The non-suffix device is a low cost limited temperature range option. Variations in electrical, mechanical and screening can be accommodated. Extensive computer simulation using complex modeling enables rapid design modification to be provided. Contact Lambda Advanced Analog with specific requirements.

#### **FEATURES**

- 16 40 VDC input range (28VDC nominal)
- ±12V and ±15V outputs available
- Indefinite short circuit and overload protection
- 35W/in<sup>3</sup> power density
- 30 watt output power
- Fast loop response for superior transient characteristics
- Operating temperature range from -55°C to +125°C
- Popular industry standard pin-out
- Resistance seam welded case for superior long term hermeticity
- Ceramic feed-thru pins
- External synchronization
- High Efficiency
- Shutdown from external signal
- Military screening

## **SPECIFICATIONS**

TCASE = -55°C to +85°C, VIN = +28 V ±5% unless otherwise specified.

#### **ABSOLUTE MAXIMUM RATINGS**

Input Voltage	-0.5 V to +5	0 V
Power Output	Internally lin	nited, 36 W typical
Soldering	300°C for 10	) seconds
Temperature Range <sup>1</sup>	Operating	-55°C + 85°C case
	Storage	-65°C+135°C

		ATR2812D		ATR2815D				
Parameter	Parameter Conditions		Тур	Мах	Min	Тур	Max	Units
STATIC CHARACTERISTIC	S							
OUTPUT Voltage Current <sup>5</sup> Ripple Accuracy Power <sup>1</sup>	VIN = 16 to 40VDC lout = 0 to Full Load Full Load, 20kHz to 2MHz TCASE = 25°C, Full Load	±11.76 0.0 ±11.88 30	±12.00 40 ±12.00	±12.24 ±1.25 85 ±12.12	±14.70 0.0 ±14.85 30	±15.00 40 ±15.00	±15.30 ±1.0 85 ±15.15	VDC ADC mV p-p VDC W
REGULATION Line Load	$V_{IN} = 16 \text{ to } 40 \text{ VDC}$ lout = 0 to Full Load			75 120			75 150	mV mV
	VIN = 16, 28 40 VDC			±5			±5	%
Voltage Range Current Ripple Current	No Load, pin 2 = open Inhibited, pin 2 tied to pin 10 Full Load	16.0	28.0	40.0 75 18 50	16.0	28.0	40.0 75 18 50	VDC mADC mADC mA p-p
FEFICIENCY	$T_{CASE} = +25^{\circ}C$							
	FullLoad		82			82		%
CAPACITIVE LOAD	No effect on performance Tc = 25°C Total for both outputs			100			100	μF
LOAD FAULT POWER DISSIPATION	Short Circuit Overload, Tc = +25°C			9 14			9 14	W W
SWITCHING FREQUENCY	louτ = F.L.	500		600	500		600	kHz
SYNC FREQUENCY RANGE <sup>7</sup>		500		700	500		700	kHz
ISOLATION	OLATION Input to Output @ 500 VDC				100			MΩ
DYNAMIC CHARACTERIS STEP LOAD CHANGES Output Transient Recovery <sup>2</sup>	TICS 50% Load to 100% Load No Load to Half Load 50% Load to 100% Load		±100 ±250 25			±100 ±250 25		mVpk mVpk µs
	No Load to 50% Load		500			500		μs
STEP LINE CHANGES Output Transient Recovery <sup>2</sup> TURN-ON	Store Load to No Load   FEP LINE CHANGES   Output Input step 16 to 40 VDC   Transient Input step 40 to 16 VDC   Recovery2 Input step 16 to 40 VDC   Input step 40 to 16 VDC   JRN-ON VIN = 16 to 40 VDC		+180 -600 5 5			+180 -600 5 5		mVpk mVpk ms ms
Overshoot	IOUT = 0 to Full Load			600			600	mVpk
	$V_{IN} = 16 \text{ to } 40 \text{ Vpc}$		14	25		14	25	ms

Notes:

1. Above +85°C case temperature, derate output power linearly to 0 at +115°C case.

2. Recovery time is measured from the initiation of the input transient to where VOUT has returned to within ±1% of VOUT at 50% load.

3. Turn-on delay time measurement is for either an application of power at the input or a signal at the inhibit pin.

Load current split equally between +Vout and -Vout.
Up to 90% of Full Power is available from either output provided. The total power output does not exceed 30 watts.

6. 3W load on output under test, 3W to 27W on other output.

7. Sync input signal: V  $_{\rm IL}$  = -0.5V Min,  $\rm ~V_{IH}$  = 2.5V Min,  $\rm ~10\%$  to 90% duty cycle

0.8V Max 11.5V Max

## **SPECIFICATIONS**

TCASE = -55°C to +125°C, VIN = +28 V ±5% unless otherwise specified.

#### **ABSOLUTE MAXIMUM RATINGS**

Input Voltage <sup>1</sup>
Power Output
Soldering
Temperature Range <sup>1</sup>

-0.5 V to +50 V Internally limited, 36 W typical 300°C for 10 seconds Operating -55°C +125°C case Storage -65°C+135°C

		ATR2812D/ES		ATR2815D/ES				
Parameter	Conditions	Min	Тур	Max	Min	Тур	Max	Units
STATIC CHARACTERISTIC	S							
OUTPUT Voltage Current <sup>5</sup> Ripple Accuracy Power <sup>1</sup>	Vin = 16 to 40Vbc lout = 0 to Full Load Full Load, 20kHz to 2MHz Tcase = 25°C, Full Load	±11.76 0.0 ±11.88 30	±12.00 40 ±12.00	±12.24 ±1.25 85 ±12.12	±14.70 0.0 ±14.85 30	±15.00 40 ±15.00	±15.30 ±1.0 85 ±15.15	VDC ADC mV p-p VDC W
REGULATION Line Load	$V_{IN} = 16 \text{ to } 40 \text{ VDC}$ lout = 0 to Full Load			75 120			75 150	mV mV
	Vin = 16, 28 40 VDC			±5			±5	%
INPUT Voltage Range Current Ripple Current	No Load, pin 2 = open Inhibited, pin 2 tied to pin 10 Full oad	16.0	28.0	40.0 75 18 50	16.0	28.0	40.0 75 18 50	VDC mADC mADC mA p-p
EFFICIENCY	TCASE = +25°C	80	82		79	82		%
CAPACITIVE LOAD	No effect on performance, Tc = 25°C Total for both outputs			100			100	μF
LOAD FAULT POWER DISSIPATION	Short circuit Overload, Tc = +25°C			9 14			9 14	W W
SWITCHING FREQUENCY	Ιουτ = F.L.	500		600	500		600	kHz
SYNC FREQUENCY RANGE <sup>7</sup>		500		700	500		700	kHz
ISOLATION	Input to Output @ 500 VDC	100			100			MΩ
DYNAMIC CHARACTERIST STEP LOAD CHANGES Output Transient	50% Load to 100% Load		±100			±100		mVpk
Recovery <sup>2</sup>	50% Load to 100% Load No Load to 50% Load 50% Load to No Load		±250 25 500 3			±250 25 500 3		μs μs ms
STEP LINE CHANGES Output Transient Recovery <sup>2</sup>	Input step 16 to 40 VDC Input step 40 to 16 VDC Input step 16 to 40 VDC Input step 40 to 16 VDC		+180 -600 5 5			+180 -600 5 5		mVpk mVpk ms ms
TURN-ON Overshoot Delay <sup>3</sup>	VIN = 16 to 40 VDC IOUT = 0 to Full Load		0 14	600 25		0 14	600 25	mVpk ms
LOAD FAULT RECOVERY	VIN = 16 to 40 VDC		14	25		14	25	ms

Notes:

Above +125°C case temperature, derate output power linearly to 0 at +135°C case.
Recovery time is measured from the initiation of the input transient to where VOUT has returned to within ±1% of VOUT at 50% load.

3. Turn-on delay time measurement is for either an application of power at the input or a signal at the inhibit pin.

4. Load current split equally between +Vout and -Vout.

6. 3W load on output under test, 3W to 27W on other output provided. The total p 6. 3W load on output under test, 3W to 27W on other output. 7. Sync input signal:  $V_{\mu} = -0.5V$  Min,  $V_{\mu} = 2.5V$  Min, 10% to 90% duty cycle

0.8V Max 11.5V Max

<sup>5.</sup> Up to 90% of Full Power is available from either output provided. The total power output does not exceed 30 watts.

## **SPECIFICATIONS**

TCASE = -55°C to +125°C, VIN = +28 V  $\pm$ 5% unless otherwise specified.

#### **ABSOLUTE MAXIMUM RATINGS**

Input Voltage
Power Output
Soldering
Temperature Range <sup>1</sup>

-0.5 V to +50 V Internally limited, 36 W typical 300°C for 10 seconds Operating -55°C +125°C case Storage -65°C+135°C

		ATR2812D/HB		ATR2815D/HB				
Parameter	Conditions	Min	Тур	Max	Min	Тур	Max	Units
STATIC CHARACTERISTIC	S							
OUTPUT	VIN = 16 to 40VDC							
Voltage	IOUT = 0 to Full Load	±11.76	±12.00	±12.24	±14.70	±15.00	±15.30	VDC
Current⁵		0.0		±1.25	0.0		±1.0	ADC
Ripple	Full Load, 20kHz to 2MHz	44.00	40	85	44.05	40	85	mV p-p
Accuracy Bower1	$ICASE = 25^{\circ}C$ , Full Load	±11.88	±12.00	±12.12	±14.85	±15.00	±15.15	
Power		30			30			VV
REGULATION	Vin 16 to 10 Vino			75			75	m\/
	VIN = 16 t0 40 VDC			120			150	m\/
	$V_{\rm IN} = 16.2840 \text{VDC}$			120			150	0/
	VIN = 10, 28 40 VDC			±0			±5	70
INPUT Voltage Range		16.0	28.0	10.0	16.0	28.0	10.0	Vpc
Current	No Load pin 2 = open	10.0	20.0	75	10.0	18	75	mADC
Odiform	Inhibited, pin 2 tied to pin 10			18		10	18	mADC
Ripple Current	FullLoad		25	50		25	50	mA p-p
EFFICIENCY	TCASE = +25°C							
		80	82		79	82		%
	FullLoad							
CAPACITIVE LOAD	No effect on performance, Tc = 25°C			100			100	μF
	Total for both outputs							
LOAD FAULT POWER	Short circuit			9			9	W
DISSIPATION	Overload Tc = +25°C			14			14	W
SWITCHING FREQUENCY	IOUT = F.L.	500		600	500		600	kHz
SYNC FREQUENCY RANGE <sup>7</sup>		500		700	500		700	kHz
ISOLATION	Input to Output @ 500 VDC	100			100			MΩ
	ICS							
Output <sup>4</sup>	50% Load to 100% Load		+100	+450		+100	+150	m\/nk
Transient	No Load to Half Load		+250	+760		+250	±430 +750	mVpk
Hallololl			1200	100		1200	100	muph
Recovery <sup>2</sup>	50% Load to 100% Load		25	70		25	70	μs
-	No Load to 50% Load		500	1500		500	1500	μs
	50% Load to No Load		3	5		3	5	ms
STEP LINE CHANGES								
Output	Input step 16 to 40 VDC		+180	1200		+180	1500	mVpk
Transient	Input step 40 to 16 VDC		-600	-1500		-600	-1500	mVpk
Recovery <sup>2</sup>	Input step 16 to 40 VDC		5	10		5	10	ms
			5	10		5	10	1115
I UKN-ON				600		_	600	m\/nl/
	1001 = 0 to Full LOad		14	25		1/	25	me
	V/w 16 to 10 V/cc		14	20		14	20	1115 mo
LUADFAULTRECOVERY	VIN = 16 to 40 VDC		14	25		14	25	ms

Notes:

1. Above +125°C case temperature, derate output power linearly to 0 at +135°C case.

2. Recovery time is measured from the initiation of the input transient to where VOUT has returned to within ±1% of VOUT at 50% load.

3. Turn-on delay time measurement is for either an application of power at the input or a signal at the inhibit pin.

4. Load current split equally between +Vout and -Vout.

5. Up to 90% of Full Power is available from either output provided. The total power output does not exceed 30 watts.
6. 3W load on output under test, 3W to 27W on other output.

7. Sync input signal: V  $_{\rm IL}$  = -0.5V Min,  $\rm ~V_{IH}$  = 2.5V Min,  $\rm ~10\%$  to 90% duty cycle

0.8V Max 11.5V Max

## **BLOCK DIAGRAM (Single Output)**



## **APPLICATION INFORMATION**

#### **Inhibit Function**

Connecting the inhibit input (Pin 2) to input common (Pin 10) will cause the converter to shut down. It is recommended that the inhibit pin be driven by an open collector device capable of sinking at least 400 $\mu$ A of current. The open circuit voltage of the inhibit input is 11.5 ±1 VDC.

#### **EMI Filter**

An optional EMI filter (AFC461) will reduce the input ripple current to levels below the limits imposed by MIL-STD-461B CEO3.

### **Device Synchronization**

Whenever multiple DC/DC converters are utilized in a single system, significant low frequency noise may be generated due to slight difference in the switching frequencies of the converters (beat frequency noise). Because of the low frequency nature of this noise (typically less than 10 kHz), it is difficult to filter out and may interfere with proper operation of sensitive systems (communications, radar or telemetry). Lambda Advanced Analog provides synchronization of multiple ATR converters to match switching frequency of the converter to the frequency fo the system clock, thus eliminating this type of noise.

## **PIN DESIGNATION**

Pin 1	Positive input	Pin 10	Input common
Pin 2	Inhibit input	Pin 9	Sync.
Pin 3	Positive output	Pin 8	Case ground
Pin 4	Output common	Pin 7	N/C
Pin 5	Negative output	Pin 6	N/C

## PART NUMBER



## Available Screening Levels and Process Variations for ATR Series

Requirement	MIL-STD-883 Method	<b>No</b> Suffix	<b>ES</b> Suffix	<b>HB</b> Suffix	<b>CH</b> Suffix
Temperature Range		-20 to +85°C	-55°C to +125°C	-55°C to +125°C	-55°C to +125°C
Element Evaluation					MIL-PRF-38534
Internal Visual	2017	*	Yes	Yes	Yes
Temperature Cycle	1010		Cond B	Cond C	Cond C
Constant Acceleration	2001		500g	Cond A	Cond A
Burn-in	1015	96hrs @ 125°C	96hrs @ 125°C	160hrs @ 125°C	160hrs @ 125°C
Final Electrical (Group A)	MIL-PRF-38534	25°C	25°C	-55, +25, +125°C	-55, +25, +125°C
Seal, Fine & Gross	1014	*	Cond A, C	Cond A, C	Cond A, C
External Visual	2009	*	Yes	Yes	Yes

\* Per Commercial Standards



#### **Thermal Management**

Assuming that there is no forced air flow, the package termperature rise above ambient ( $\Delta T$ ) may be calculated using the following expression:

$$\Delta T \approx 80 \text{ A}^{-0.7} \text{ p}^{-0.85}$$
 ( °C)

where A = the effective surface area in square inches (including heat sink if used;) P = power dissipation in watts.

The total surface area of the ATR standard package is 7.34 square inches. If a worse case full load efficiency of 78% is assumed, then the case temperature rise can be calculated as follows:

$$P = P_{OUT} \left[ \frac{1}{Eff} - 1 \right] = 30 \left[ \frac{1}{78} - 1 \right] = 8.5 W$$
$$\Delta T = 80 \ (7.34)^{-0.7} \ (8.5)^{0.85} = 122^{\circ}C$$

Hence, if  $T_{AMBIENT} = +25^{\circ}C$ , the DC/DC converter case temperature will be approximately 147°C if no heat sink or air flow is provided.

To calculate the heat sink area required to maintain a specific case temperature rise, the above equation may be manipulated as follows:

A <sub>HEAT SINK</sub> = 
$$\left[ \Delta T \right]^{-1.43} - A_{PKG}$$

As an example, if a maximum case temperature rise of 50°C above ambient is desired, then the required effective heat sink area is:

A <sub>HEAT SINK</sub> =  $[50 \\ 80 (8.5)^{-1.43} - 7.34 = 19.1 \text{ in.}^2$ 

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MIL-PRF-38534 Certified

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