



ALO20F36 Single Output 8th Brick: 3.3V at 20A output Wide Range (19V - 60V Input Range)

The ALO20F36 series is Astec's latest addition to its 8th brick standard products offering. It operates from an input range of 19V to 60V suited for wireless/base station applications that are configured for a 24V or 48V input voltages. The ALO20F36 series is designed to provide 3.3V output and handle up to 20A of load current. It comes with industry standard features such as Input UVLO; Output Enable; non-latching OCP, OVP and OTP; Output Trim; Differential Remote Sense pins.



Special Features

- **Industry Standard 8th Brick Footprint**
- Low Ripple and Noise
- Regulation to zero load
- **High Capacitive Load Start-up**
- **Fixed Switching Frequency**
- **Industry standard features: Input UVLO;** Enable; non-latching OVP, OCP and OTP; **Output Trim. Differential Remote Sense**
- **Meets Basic Insulation**

Environmental Specifications

- -40°C to 85°C Operating Temperature
- -40°C to 125°C Storage Temperature
- MTBF > 1 million hours

Electrical Parameters

Input

Input Range 19-60 VDC **Input Surge** 100V / 100ms

Control

Enable TTL compatible (Positive or Negative Logic Enable Options)

Output

Load Current Up to 20A max Line/Load Regulation 0.1% Vo Typical **Ripple and Noise** 50mV_{P-P} typical

Output Voltage

Adjust Range $\pm 10\% V_{O}$

3% Typical deviation **Transient Response**

> 50% to 75% Load Change 120µs settling time (Typ)

 $+10\%V_{0}$

Remote Sense

125% Io,max **Over Current**

Protection

Over Voltage 125% Vo, nom

Protection

Over Temperature 110 °C

Protection

Safety

UL + cUL 60950-1, Recognized EN60950-1 through TUV-PS





Electrical Specifications

ABSOLUTE MAXIMUM RATINGS

Stresses in excess of the absolute maximum ratings can cause permanent damage to the converter. Functional operation of the device is converter is not implied at these or any other conditions in excess of those given in the operational section of the specs. Exposure to absolute maximum ratings for extended period can adversely affect device reliability.

| Parameter | Device | Symbol | Min | Typical | Max | Unit |
|---------------------------|----------|-------------|------|---------|-----|------|
| Input Voltage | | | | | | |
| Continuous | All | Vin | -0.3 | - | 60 | Vdc |
| Transient (100ms) | | Vin trans | - | - | 100 | |
| I/O Isolation | | | | | | |
| Input-to-Output | All | - | 1500 | - | - | Vdc |
| Operating Temperature | All | T_A | -40 | - | 85 | °C |
| Storage Temperature | All | T_{STG} | -55 | - | 125 | °C |
| Operating Humidity | All | - | 10 | - | 85 | % |
| Max Voltage at Enable Pin | All | | -0.6 | - | 25 | Vdc |
| Max Output Power | F (3.3V) | $P_{O,MAX}$ | - | - | 66 | W |

INPUT SPECIFICATION

| Parameter | Device | Symbol | Min | Typical | Max | Unit |
|-----------------------------------|---------|--------------|------|---------|------|-------|
| Operating Input Voltage Range | All | $V_{\rm IN}$ | 19 | 24/48 | 60 | Vdc |
| Input Under-Voltage Lock-out | | | | | | |
| T_ON Threshold | All | | 18.5 | - | 20 | Vdc |
| T_OFF Threshold | | | 16.0 | - | 18.4 | |
| Max Input Current ¹ | F (3V3) | Iin_{max} | - | - | 4.5 | A |
| | | | | | | |
| Standing Loss | All | | - | 1.5 | 3 | W |
| | | | | | | |
| Input Ripple Current ² | All | I_{I1} | - | 10 | 30 | mAp-p |
| | | | | | | |
| External Input Bypass Cap | All | C_{IN} | | 220 | | μF |
| | | | | | | |

Note:

- 1. Module is not internally fused. External fuse is recommended (e.g. Littlefuse $^{@}$ 465 Series / 1.5 x $I_{IN,MAX}$ typical).
- 2. See Figure 1 for the input ripple current test setup.





Electrical Specifications (continued)

OUTPUT SPECIFICATIONS

| Parameter | Device | Symbol | Min | Typical | Max | Unit |
|--|----------|--------------------|------|---------|--------|-----------------|
| Output Voltage Set point | F (3V3) | $V_{O,SET}$ | 3.2 | 3.30 | 3.36 | Vdc |
| $V_{IN} = V_{IN,MIN}$ to $V_{IN,MAX}$ | | | | | | |
| $I_{O} = I_{O,MAX}$ | | | | | | |
| Output Regulation | | | | | | |
| Line | All | - | - | 0.1 | 0.5 | % |
| $V_{IN} = V_{IN,MIN}$ to $V_{IN,MAX}$ | | | | | | |
| Load | | - | - | 0.1 | 0.5 | |
| $V_{IN} = V_{IN,NOM}$ | | | | | | |
| $I_{\rm O} = I_{\rm O,MIN}$ to $I_{\rm O,MAX}$ | | | | 0.27 | 0.7 | |
| Temp | | - | - | 0.25 | 0.5 | |
| $V_{IN} = V_{IN,NOM}$; $I_O = I_{O,MAX}$ $T_a = -40$ °C to 85 °C | | | | | | |
| Output Ripple and Noise ³ | F (3V3) | - | - | 50 | 100 | mVp-p |
| Peak-to-Peak | | | | | | |
| $I_O = I_{O,MAX}; V_{IN} = V_{IN,NOM};$ | | | | | | |
| $BWL = 20 \text{ MHz}; T_A = 25 ^{\circ}\text{C}$ | | | | | | |
| 2 4 | | | | | | |
| Output Current ⁴ | F (3V3) | I_{O} | 0 | - | 20 | A |
| Output Current-limit Inception ⁵ | F (3V3) | I _{O,OCP} | 22 | - | 30.5 | A |
| $V_{O} = 90\% \ V_{O,NOM}; T_{A} = 25 \ ^{\circ}C$ | (- ' - ' | 0,001 | | | | |
| $V_{IN} = V_{IN,NOM}$ | | | | | | |
| Non-latching / auto-recovery | | | | | | |
| External Load Capacitance | | | | | | |
| $I_{O} = I_{O,MAX}$, resistive load | F (3V3) | C_{EXT} | - | - | 10,000 | μF |
| | | | | | | |
| ESR | | | 8 | - | - | $m\Omega$ |
| Efficiency | F (3.3V) | η | 88.5 | 91 | | % |
| $V_{IN} = V_{IN,NOM}$; $I_O = I_{O,MAX}$ | | , | | | | |
| $T_A = 25$ °C; | | | | | | |
| Output Over Voltage Protection ⁵ | F (3.3V) | $V_{O,OVP}$ | 3.9 | 4.1 | 4.7 | V |
| Non-latching / autorecovery | | | | | | |
| Over Temperature Protection | | | | | | |
| Autorecovery | All | | 110 | - | 120 | °C |
| Input to Output Turn-On Delay | All | - | - | - | 15 | ms |
| $V_{IN} = V_{IN,NOM,} I_O = I_{O,MAX}$ | | | | | | |
| Enable to Output Turn-On Delay | All | - | - | - | 15 | ms |
| $V_{\rm IN} = V_{\rm IN,NOM,} I_{\rm O} = I_{\rm O,MAX}$ | | | | | | |
| Switching Frequency | All | F_{SW} | 340 | 400 | 460 | kHz |
| Output Voltage Remote Sensing ⁶ | All | - | - | - | 10 | %V _O |
| Output Voltage Trim Range ⁷ | All | | 90 | | 110 | %V _O |





Electrical Specifications (continued)

OUTPUT SPECIFICATIONS

| Parameter | Device | Symbol | Min | Typical | Max | Unit |
|--|--------|--------|-------|---------|------|------|
| Dynamic Response | | | | | | |
| $di/dt = 0.1 A/\mu s$ | | | | | | |
| Peak Deviation | All | - | - | 3 | 4 | % |
| $\Delta I_{\rm O} = 50\%$ to 75% of $Io_{\rm max}$ | | | | | | |
| Settling Time | | | | | | |
| $Vref = Vo_{nom}$ | All | - | - | - | 200 | μs |
| Peak Deviation | | | | | | |
| $\Delta I_{\rm O} = 50\%$ to 25% of Io_{max} | All | - | - | 3 | 4 | % |
| Settling Time | All | - | - | - | 200 | μs |
| $Vref = Vo_{nom}$ | | | | | | • |
| Output Enable ON/OFF | | | | | | |
| Open collector TTL compatible | | | | | | |
| Positive Enable: Mod-ON | All | - | 2.95 | - | 20 | V |
| Mod-OFF | All | - | -0.50 | - | 1.20 | V |
| | | | | | | |
| Negative Enable: Mod-ON | All | - | -0.50 | - | 1.20 | V |
| Mod-OFF | All | - | 2.95 | - | 20 | V |

Note:

- 3. See Figure 2 for the Output Ripple and Noise Test Measurement Setup.
- 4. Output derating applies at elevated temperature. See Figures 16 and 17.
- 5. OCP and OVP are both auto-recovery. The converter will shutdown and attempt to restart until the fault is removed. There is a 25ms lockout period between restart attempts. Note also that the OCP threshold will be reduced proportionally with output voltage trim up and/or remote sense compensation. The percent rise in output voltage will be proportional to the reduction in OCP current limit inception.
- 6. The sense pins can be used to compensate for any voltage drops (per indicated max limits) that may occur along the connection between the output pins to the load. Pin 7 (+Sense) and Pin 5 (-Sense) should be connected to Pin 8 (+Vout) and Pin 4 (Return) respectively at the point where regulation is desired.
- 7. See Equation 1 and 2 for the output trim function. The combination of remote sense and trim adjust cannot exceed 100% Vo. Whenever the output voltage is increased, the output current must be derated so as not to exceed the maximum output power.
- 8. Minimum Enable pin disable time is 100ms. Shorter disable durations may cause output to overshoot beyond specification upon restart (Enable On).

SAFETY AGENCY / MATERIAL RATING / ISOLATION

| Parameter | Device | | | | | | | |
|---------------------------------|--------|------------------------|--|------|------|-------|--|--|
| Safety Approval | All | UL/cUL | UL/cUL 60950-1, 3rd Edition – Recognized | | | | | |
| | | EN 60950-1 through TUV | | | | | | |
| Material Flammability Rating | All | UL94V-0 | | | | | | |
| Parameter | Device | Symbol Min Typical Max | | | Unit | | | |
| Input to Output Capacitance | All | | - | 1000 | - | pF | | |
| Input to output Resistance | All | | - | 10 | - | Mohms | | |
| Input to Output Insulation Type | All | | - Basic - | | | | | |





Electrical Specifications (continued)

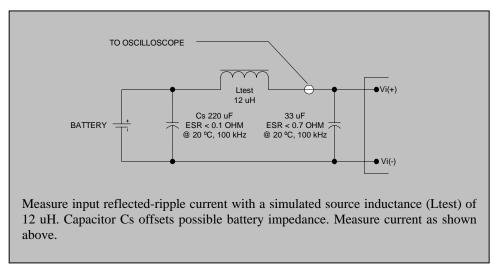


Figure 1. Input Reflected Ripple Current Measurement Setup.

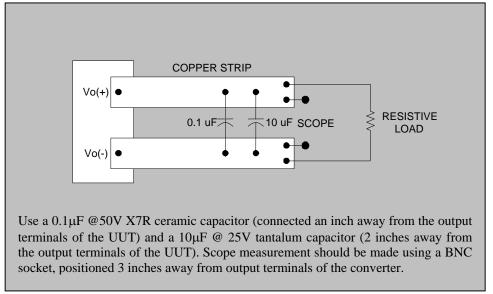


Figure 2. Peak to Peak Output Noise Measurement Setup.





Basic Operation and Features

INPUT UNDER VOLTAGE LOCKOUT

To prevent any instability to the converter, which may affect the end system, the converter have been designed to turn-on once V_{IN} is in the voltage range of 18.5-20.0 VDC. Likewise, it has also been programmed to turn-off when V_{IN} drops down to 16.0-18.4 VDC.

OUTPUT VOLTAGE ADJUST/TRIM

The converter comes with a TRIM pin (PIN 6), which is used to adjust the output by as much as 90% to 110% of its set point. This is achieved by connecting an external resistor as described below.

To **INCREASE** the output, external R_{adj_up} resistor should be connected between TRIM PIN (Pin6) and +SENSE PIN (Pin 7). Please refer to Equation (1) for the required external resistance and output adjust relationship.

Equation (1a):

$$R_{\text{adj_up}} = \left[\frac{5.1 \times \text{Vo}_{\text{set}} \times (100 + \Delta\%)}{1.225 \times \Delta\%} - \frac{510}{\Delta\%} - 10.2 \right] \quad K\Omega$$

To **DECREASE** the output, external R_{adj_down} resistor should be connected between TRIM pin (Pin 6) and -SENSE PIN (Pin 5). Please refer to Equation (2) for the required external resistance and output adjust relationship.

Equation (2):

Radj_down =
$$\left(\frac{510}{\Delta\%} - 10.2\right) \cdot k\Omega$$

Where: Δ % = percent change in output voltage:

$$\Delta\% = \left(\frac{\text{Vo}_{\text{desired}} - \text{Vo}_{\text{set}}}{\text{Vo}_{\text{set}}}\right) \times 100$$

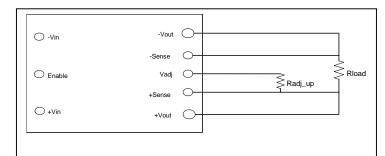


Figure 3. External resistor configuration to increase the outputs

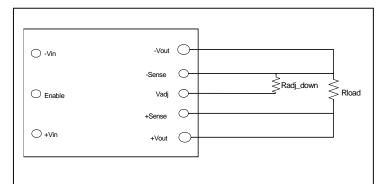


Figure 4. External resistor configuration to increase the outputs

OUTPUT ENABLE

The converter comes with an Enable pin (PIN 2), which is primarily used to turn ON/OFF the converter. Both a Positive (no "N" suffix required) and a Negative (suffix "N" required) Enable Logic options are being offered. Please refer to Table 2 for the Part Numbering Scheme.

For Positive Enable, the converter is turned on when the Enable pin is at logic HIGH or left open. The unit turns off when the Enable pin is at logic LOW or directly connected to $-V_{IN}$. On the other hand, the Negative Enable version turns unit on when the Enable pin is at logic LOW or directly connected to $-V_{IN}$. The unit turns off when the Enable pin is at Logic HIGH.





Basic Operation and Features (continued)

OUTPUT OVER VOLTAGE PROTECTION (OVP)

The Over Voltage Protection circuit is non-latching - auto recovery mode. The output of the converter is terminated under an OVP fault condition (Vo > OVP threshold). The converter will attempt to restart until the fault is removed. There is a 25ms lockout period between restart attempts.

OVER CURRENT PROTECTION (OCP)

The Over Current Protection is non-latching - auto recovery mode. The converter shuts down once the output current reaches the OCP range. The converter will attempt to restart until the fault is removed. There is a 25ms lockout period between restart attempts.

OVER TEMPERATURE PROTECTION (OTP)

The Over Temperature Protection circuit will shutdown the converter once the average PCB temperature (See Figure 90B for OTP reference sense point) reaches the OTP range. This feature prevents the unit from overheating and consequently going into thermal runaway, which may further damage the converter and the end system. Such overheating may be an effect of operation outside the given power thermal derating conditions. Restart is possible once the temperature of the sensed location drops to less than 110°C.

REMOTE SENSE

The remote sense pins can be used to compensate for any voltage drops (per indicated max limits) that may occur along the connection between the output pins to the load. Pin 7 (+Sense) and Pin 5 (-Sense) should be connected to Pin 8 (+Vout) and Pin 4 (Return) respectively at the point where regulation is desired. The combination of remote sense and trim adjust cannot exceed 110% of V_O . When output voltage is trimmed up (through remote sensing <u>and/or</u> trim pin), output current must be derated and maximum output power must not be exceeded.





Performance Curves

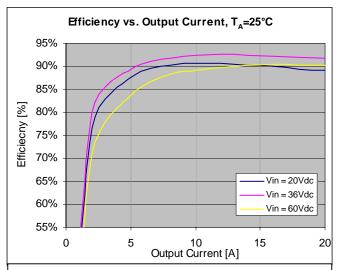


Figure 6. Efficiency vs. Load current at various line voltages, $T_A = 25$ °C.

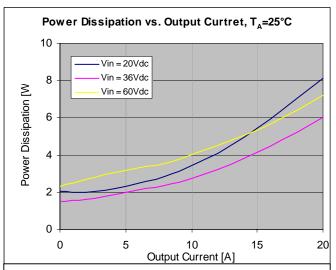


Figure 7. Power Dissipation vs. Load current at various line voltages, $T_A = 25$ °C.

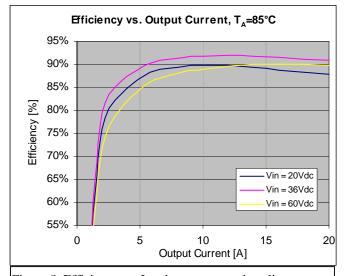


Figure 8. Efficiency vs. Load current at various line voltages, $T_A = 85$ °C.

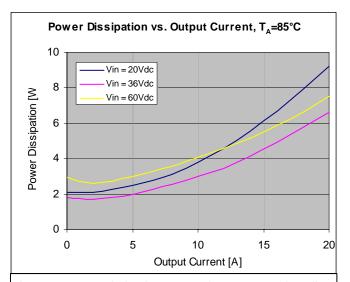


Figure 9. Power Dissipation vs. Load current at various line voltages, $T_A = 85$ °C.





Performance Curves

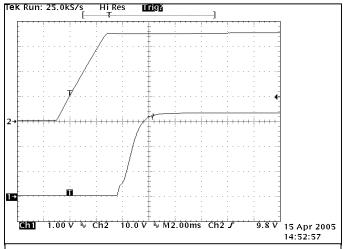


Figure 10. Output startup (CH1) at V_{IN} = 36Vdc, I_O = 20A, C_O = 10 μ F, T_A = 25°C.

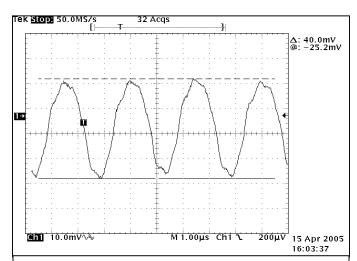


Figure 11. Output ripple at $V_{IN} = 36Vdc$, $I_O = 20A$, $T_A = 25$ °C. See Fig 2 for test setup.

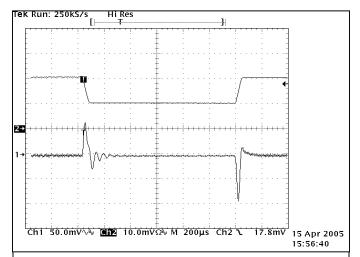


Figure 12. Output transient response (CH1) at $V_{IN} = 36Vdc$, 25% to 50% step change, $T_A = 25^{\circ}C$, $C_O = 10\mu F$.

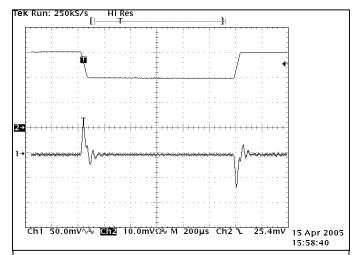


Figure 13. Output transient response (CH1) at $V_{IN} = 36 \text{Vdc}$, 25% to 50% step change, $T_A = 25^{\circ}\text{C}$, $C_O = 10 \mu\text{F}$.





Performance Curves

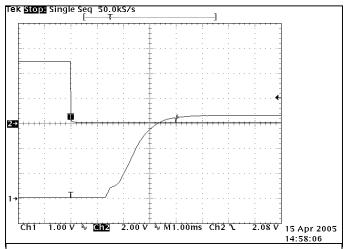


Figure 14. Enable to output turn on characteristic. $V_{IN} = 36V$, $T_A = 25$ °C.

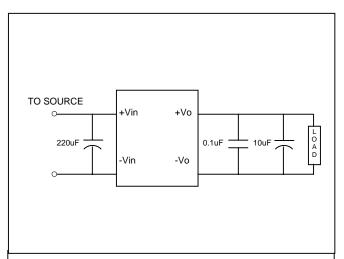


Figure 15. Typical application circuit.

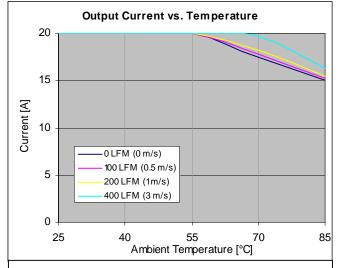


Figure 16. Output current vs. ambient temperature, $V_{IN} = 24 \text{Vdc} (T_J \le 120^{\circ}\text{C})$.

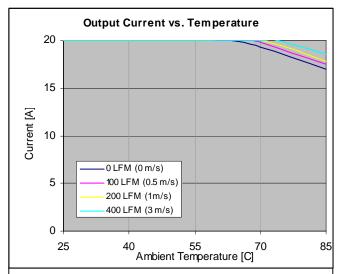


Figure 17. Output current vs. ambient temperature, $V_{IN} = 48 V dc (T_J \le 120^{\circ}C)$.





Input Filter for FCC Class B Conducted Noise

A reference design for an input filter that can provide FCC Class B conducted noise levels is shown below (See Figure 18). Two common mode connected inductors are used in the circuit along with balanced bypass capacitors to shunt common mode currents into the ground plane. Shunting noise current back to the converter reduces the amount of energy reaching the input LISN for measurement.

The application circuit shown has an earth ground (frame ground) connected to the converter output (-) terminal. Such a configuration is common practice to accommodate safety agency requirements. Grounding an output terminal results in much higher conducted emissions as measured at the input LISN because a hard path for common mode current back to the LISN is created by the frame ground. "Floating" loads generally result in much lower measured emissions. The electrical equivalent of a floating load, for EMI measurement purposes, can be created by grounding the converter output (load) through a suitably sized inductor(s) while maintaining the necessary safety bonding.

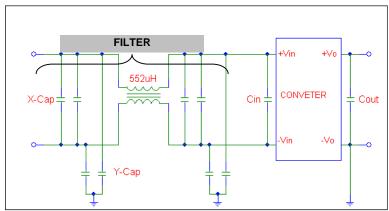


Figure 18: Class B Filter Circuit

PARTS LIST

| CKT CODE | DESCRIPTION |
|----------------------|-------------------|
| Common | CTX01-15091 |
| Common Mode Choke | Cooper Electronic |
| | Technologies |
| X-Cap | 0.47 μF X 4pcs |
| Y-Cap | 22 nF X 4 pcs |
| C _{IN} | 220 uF |





Mechanical Specifications

| Parameter | Device | Symbol | | Min | Тур | Max | Unit | | | |
|----------------|----------------|-----------|----|-------------|-------------|--------------|-----------|------|---|--|
| Dimension | All | L | 2 | .28 [57.91] | 2.30 [58.4] | 2.32 [58.9] | in [mm] | | | |
| | | W | 0. | .88 [22.35] | 0.90 [22.9] | 0.92 [23.36] | in [mm] | | | |
| | | Н | 0 | 0.38 [9.65] | 0.40 [10.1] | 0.42 [10.66] | in [mm] | | | |
| Weight | All | | - | | 22.68 [0.8] | - | g [oz] | | | |
| PIN ASSIGNMENT | PIN ASSIGNMENT | | | | | | | | | |
| 1 | - | $+V_{IN}$ | | | 5 | -SENS | SE | | | |
| 2 | EN | ABLE | | 6 | | 6 TR | | TRIN | A | |
| 3 | | $-V_{IN}$ | 7 | | 7 | | SE | | | |
| 4 | | -Vo | | 8 | | 8 +Vo | |) | | |

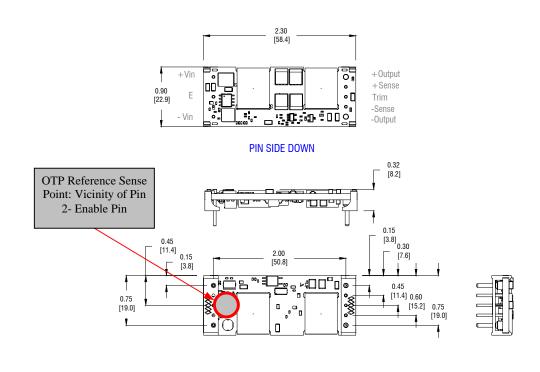


Figure 19. Mechanical Outline





Mechanical Specifications

SOLDERING CONSIDERATIONS

The RoHS-compliant terminal pin finish of the converter are compatible with both Pb and Pb-free wave soldering techniques. The modules are recommended to be preheated for 20-30 sec at 110°C and wave soldered at 260°C for Pb solder and 270°C max for Pb-free solder, for less than 10 sec.

TABLE 2: PART NUMBERING SCHEME

| | O/P Current | O/P Voltage | I/P Voltage | Enable Logic | | PIN Length | RoHS |
|-----|-------------|-------------|-------------|---|---|---|------------------------------|
| ALO | 20 | F | 36 | N | _ | 6 | L |
| | 20 A | F = 3.3 V | 19 – 60 Vdc | N = Negative Logic BLANK = Positive Logic | | 6 = 3.7mm nom Blank = 5mm nom (default) | L = RoHS 6 Blank = RoHS 5 |

Note: 1) For Through Hole termination:

- Std pin length is 5mm nominal (min: 0.189 [4.8]; max: 0.205 [5.2] / in [mm])
- "-6" option is 3.7mm nominal (min: 0.137 [3.5]; max: 0.152 [3.9] / in [mm])
- Pins 4&8 diameter: \emptyset = 0.062 [1.57], others: \emptyset = 0.04 [1.0] (6X)

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