



Datasheet

DS000622

AS7026GG

Biosensor

v2-00 • 2019-Sep-17

Content Guide

1	General Description.....	3	7.1	Register Overview.....	17
1.1	Key Benefits & Features	3	7.2	I ² C.....	94
1.2	Applications	4	8	Application Information	105
1.3	Block Diagram	4	8.1	Application Examples.....	105
2	Ordering Information.....	5	9	Package Drawings & Markings.	106
3	Pin Assignment	6	10	Revision Information.....	107
3.1	Pin Diagram.....	6	11	Legal Information	108
3.2	Pin Description	6			
4	Absolute Maximum Ratings.....	8			
5	Electrical Characteristics.....	9			
6	Functional Description.....	15			
6.1	Optical Analog Front End	15			
7	Register Description.....	17			

1 General Description

The operation of the AS7026GG is based on photoplethysmography (PPG) and electrocardiogram (ECG). PPG is the most used HRM method, which measures the pulse rate by sampling light modulated by the blood vessels, which expand and contract as blood pulses through them. ECG is the reference for any measurement of the bio potential generated by the heart.

The AS7026GG is supported by algorithms converting the PPG and ECG readings into digital HRM, HRV and continuous blood pressure values. The cuff-less blood pressure measurement has been optimized to full the IEEE Std 1708-2014.

The module includes the LEDs, photo-sensor, analog front-end (AFE) and sequencer as well as application software. In addition to HRM/HRV and blood pressure, the module also enables skin temperature and skin resistivity measurements by providing interfaces to external sensors.

The AS7026GG's low-power design and small form factor is particularly well suited to application in fitness bands, smart watches, sports watches and smart patches, in which board space is limited and in which users look for extended, multi-day intervals between battery recharges.

1.1 Key Benefits & Features

The benefits and features of AS7026GG, Biosensor, are listed below:

Figure 1:
Added Value of Using AS7026GG

Benefits	Features
Address all skin types	Improved optical path
Allows smallest application size e.g. narrow HRM measurement band	Single device integrated optical solution
Electrocardiogram ECG with dry electrodes	Embedded low noise analog front end for ECG signals acquisition
Enabling blood pressure measurements	Synchronized PPG and ECG acquisition
Good HRM measurement quality	Low noise analog optical front end
Additional information for end user	Analog electrical front end (e.g. for temperature sensing using a NTC or galvanic skin resistivity (GSR))
Integrated interference filter	Reduce negative effect of strong sunlight
Long operating time	Hardware sequencer to offload processor Adjustable LED driver with current control

Benefits	Features
Works reliably with ambient light	Embedded ambient light rejection system

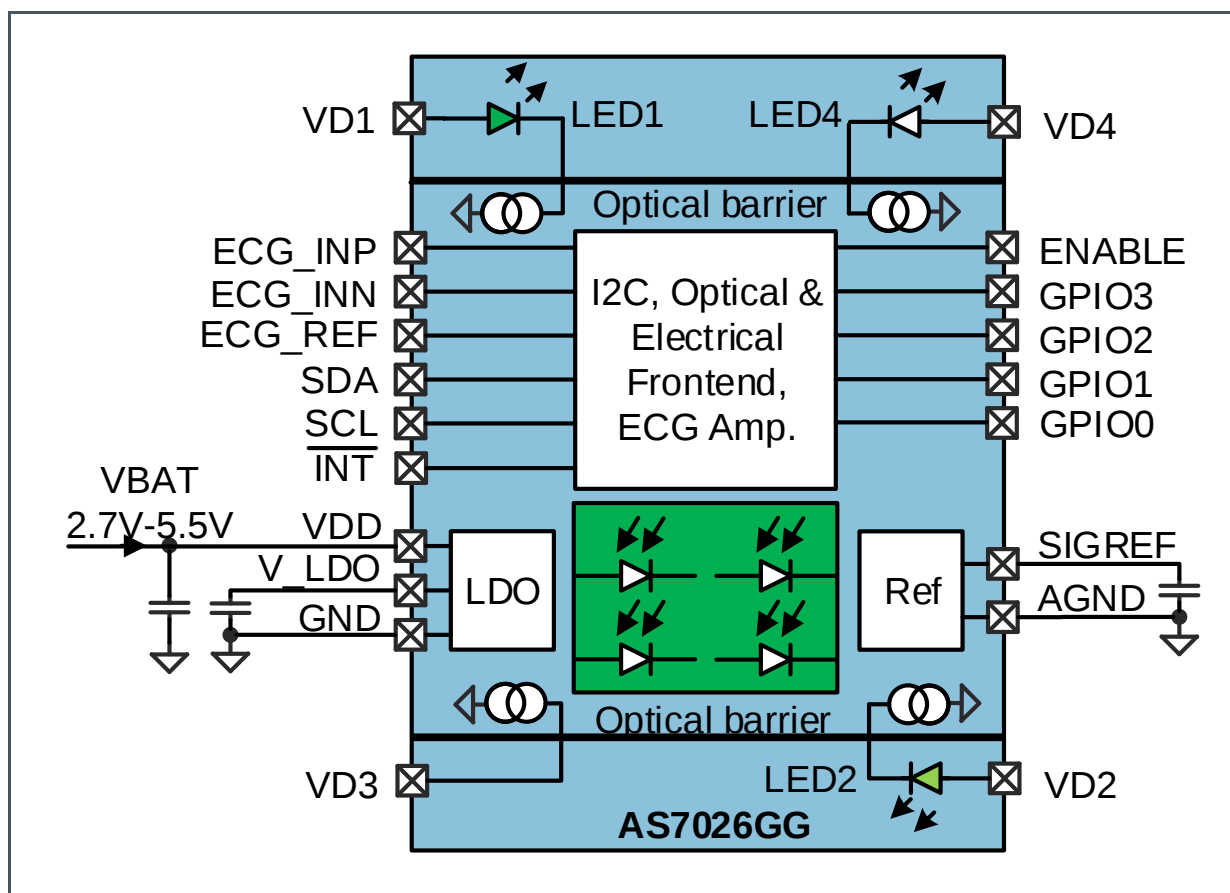
1.2 Applications

- Optical sensor platform
- Fitness band
- Smart watch
- Heart rate monitor
- Cuff-less blood pressure measurements

1.3 Block Diagram

The functional blocks of this device are shown below:

Figure 2 :
Functional Blocks of AS7026GG



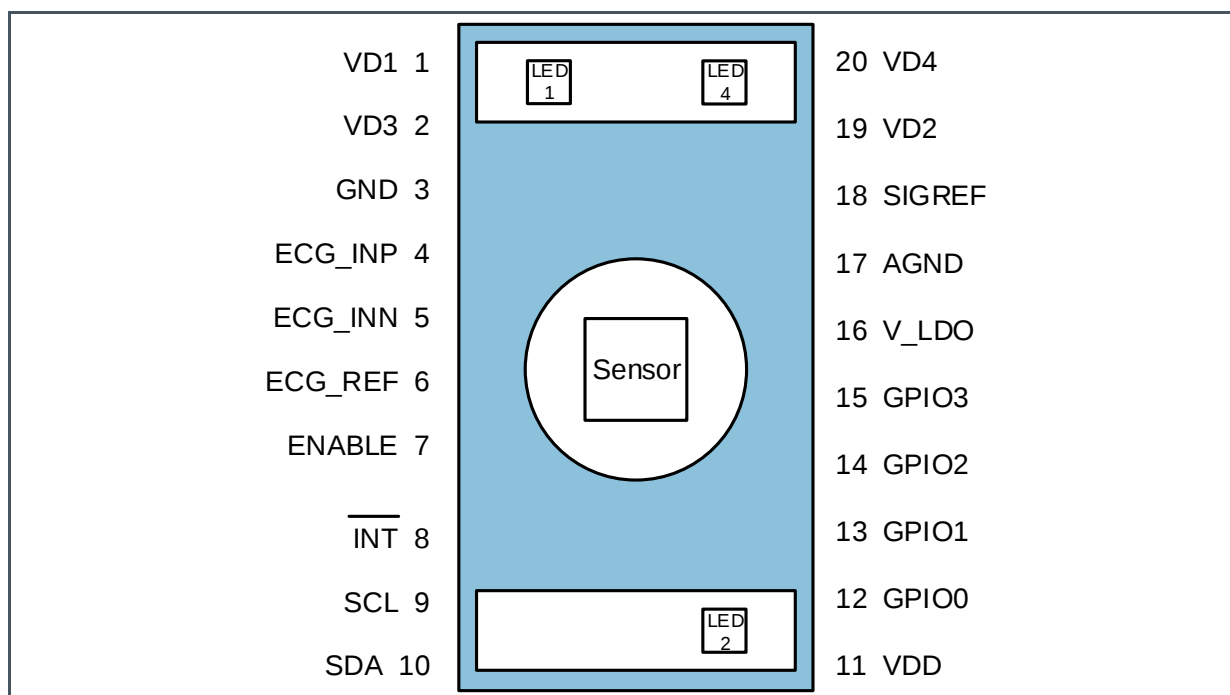
2 Ordering Information

Ordering Code	Package	Marking	Delivery Form	Delivery Quantity
AS7026GG-COLT	OLGA-20	n.a.	Tape & Reel	5000 pcs/reel
AS7026GG-COLM	OLGA-20	n.a.	Tape & Reel	1000 pcs/reel

3 Pin Assignment

3.1 Pin Diagram

Figure 3:
AS7026GG Optical Module Pin-Out –Top View (not to scale)



3.2 Pin Description

Figure 4:
Pin Description of AS7026GG

Pin Number	Pin Name	Pin Type ⁽¹⁾	Description
1	VD1	AI	Supply voltage for LED D1
2	VD3	AI	Connection to current sink 3
3	GND	G	Power supply ground. All voltages are referenced to GND.
4	ECG_INP	AI	ECG amplifier positive input
5	ECG_INN	AI	ECG amplifier negative input
6	ECG_REF	AO	ECG amplifier reference output

Pin Number	Pin Name	Pin Type ⁽¹⁾	Description
7	ENABLE	DI	Enable input for AS7026GG. Active high. Setting this input to low resets all internal registers and the AS7026GG enters power down mode. Setting it high allows operation of the AS7026GG. If ENABLE is not used (AS7026GG always enabled), connect to VDD.
8	INT	DO	Open drain interrupt output pin. Active low.
9	SCL	DI	I ² C serial clock input terminal – the device does not use clock stretching therefore SCL is only an input terminal.
10	SDA	DI	I ² C serial data I/O terminal – open drain.
11	VDD	P	Supply voltage. Connect a 2.2 µF capacitor to GND.
12	GPIO0	GPIO	General purpose input/output
13	GPIO1	GPIO	General purpose input/output
14	GPIO2	GPIO	General purpose input/output
15	GPIO3	GPIO	General purpose input/output
16	V_LDO	AO	1.9 V output voltage. Connect 2.2 µF capacitor to GND (e.g. 0402 sized capacitor GRM153R60J225ME95 or 0201 sized GRM033R60J225ME47 from Murata – needs to have >1 µF with 1.0 V voltage bias); do not load externally
17	AGND	GND	Analog ground. Connect to low noise GND
18	SIGREF	AO	Analog reference output. Connect 2.2 µF capacitor to GND (e.g. 0402 sized capacitor GRM153R60J225ME95 or 0201 sized GRM033R60J225ME47 from Murata – needs to have >1 µF specified for 1.0 V voltage bias); do not load externally The typical operating voltage on this pin is 0.6 V (sigref_en=1)
19	VD2	AI	Supply voltage for LED D2
20	VD4	AI	Supply voltage for LED D4

(1)	DI	Digital Input
	DO	Digital Output
	AI	Analog Input
	AO	Analog Output
	GPIO	General Purpose IO
	P	Power Supply

4 Absolute Maximum Ratings

Stresses beyond those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. These are stress ratings only. Functional operation of the device at these or any other conditions beyond those indicated under “Operating Conditions” is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Figure 5
Absolute Maximum Ratings of AS7026GG

Symbol	Parameter	Min	Max	Unit	Comments
Electrical Parameters					
V_{SUP} / V_{GND}	Supply Voltage to Ground		6	V	
V_{IN}	Input Pin Voltage to Ground pins GPIO0/1/2/3	-0.3	VDD+0.3V max. 6V	V	Diode to VDD
$V_{IN-OTHER}$	Input Pin Voltage to Ground pins SCL/SDA/INT/ENABLE and VD1/VD2/VD3/VD4	-0.3	5.5	V	No internal diode to VDD or V_LDO
$V_{VD1/2/3/4_INTERNAL}$	Voltage between internal pin of VD1-VD4 to VDD		VDD+0.3V	V	Internal diode between current source (internal node at anode of the LED if the pin has a LED otherwise VD1/2/3/4 pin) and VDD
V_{IN-LDO}	Input Pin Voltage to Ground for pin V_LDO	-0.3	2.0 V	V	Diode to VDD
V_{IN-LDO_DIODE}	Input Pin Voltage to Ground pins for ECG_INP/ECG_INN/ECG_REF/SIGREF	-0.3	2.0 V		Diode to V_LDO
$V_{GND-AGND}$	Analog to power ground voltage difference	-0.3	+0.3	V	
I_{SCR}	Input Current (latch-up immunity)		± 100	mA	JEDEC JESD78D Nov 2011
Electrostatic Discharge					
ESD_{HBM}	Electrostatic Discharge HBM		± 2	kV	JS-001-2014
Temperature Ranges and Storage Conditions					
T_{STRG}	Storage Temperature Range	- 40	85	°C	
T_{BODY}	Package Body Temperature		260	°C	IPC/JEDEC J-STD-020 ⁽¹⁾
RH_{NC}	Relative Humidity (non-condensing)	5	85	%	
MSL	Moisture Sensitivity Level		3		Maximum floor life time of 168h

- (1) The reflow peak soldering temperature (body temperature) is specified according to IPC/JEDEC J-STD-020 “Moisture/Reflow Sensitivity Classification for Nonhermetic Solid State Surface Mount Devices.” The lead finish for Pb-free leaded packages is “Matte Tin” (100 % Sn)

5 Electrical Characteristics

All limits are guaranteed. The parameters with Min and Max values are guaranteed with production tests or SQC (Statistical Quality Control) methods.

VDD=2.7 V to 5.5 V, typ. values are at T_{AMB}=25 °C (unless otherwise specified).

Figure 6:
Electrical Characteristics of AS7026GG

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
VDD	Supply voltage		2.7	3.8	5.5	V
T _{AMB}	Operating free-air temperature		-30		70	°C
IDD	Supply current	ENABLE=VDD, ldo_en=0; osc_en=0; internal LDO operating in low power mode – only I ² C communication possible, no blocks shall be enabled ⁽¹⁾		22		μA
		ENABLE=VDD, ldo_en=1; osc_en=0; internal LDO operating and bandgap running – I ² C communication possible, analog blocks can be enabled ⁽¹⁾		32		μA
		ENABLE=VDD, ldo_en=1, osc_en=1; internal LDO operating and bandgap and oscillator running – I ² C communication possible, analog blocks can be enabled		86		μA
		SIGREF buffer (sigref_en=1)		52		μA
		transimpedance amplifier (pd_amp_en=1)		110		μA
		Optical front end operating (one channel)		200		μA

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
		Gain stage (ofe1_gain_en=1 or ofe2_gain_en=1)		75		μA
		ADC sampling at 20 Hz with 64 μs settling time		4.5		μA
		ECG amplifier and frontend (need SIGREF enabled)		190		μA
		ECG leakage compensation (ecg_low_leakage_en=1), low pass filter, high pass filter and gain stage		151		μA
		Power down, no I ² C communication possible ENABLE=GND ⁽²⁾		0.5		μA
VOL	GPI00-3, INT, SDA output low voltage	With 3 mA load With 6 mA load	0 0		0.4 0.8	V
VOH	GPI00-3 output high voltage	With 3 mA load	2.3		VDD	V
VIH	GPI00-3, SCL, SDA, ENABLE input high voltage		1.25			V
VIL	GPI00-3, SCL, SDA, ENABLE input low voltage				0.54	V
ILEAK1	GPI00-3, SCL, SDA, ENABLE, INT		-1		1	μA
ILEAK2	VD1, VD2 VD3, VD4		-3		3	μA
E_f2M	Tolerance of internal 2 MHz oscillator	0 °C to 70 °C, VDD<5.0 V	-2		2	%
		-30 °C to 70 °C	-4		2	%
ECG Amplifier and Filter						
ILEAK_ECG	ECG pins leakage current	Lab evaluation shows <±20 nA maximum leakage current. Not production tested.		±1		nA

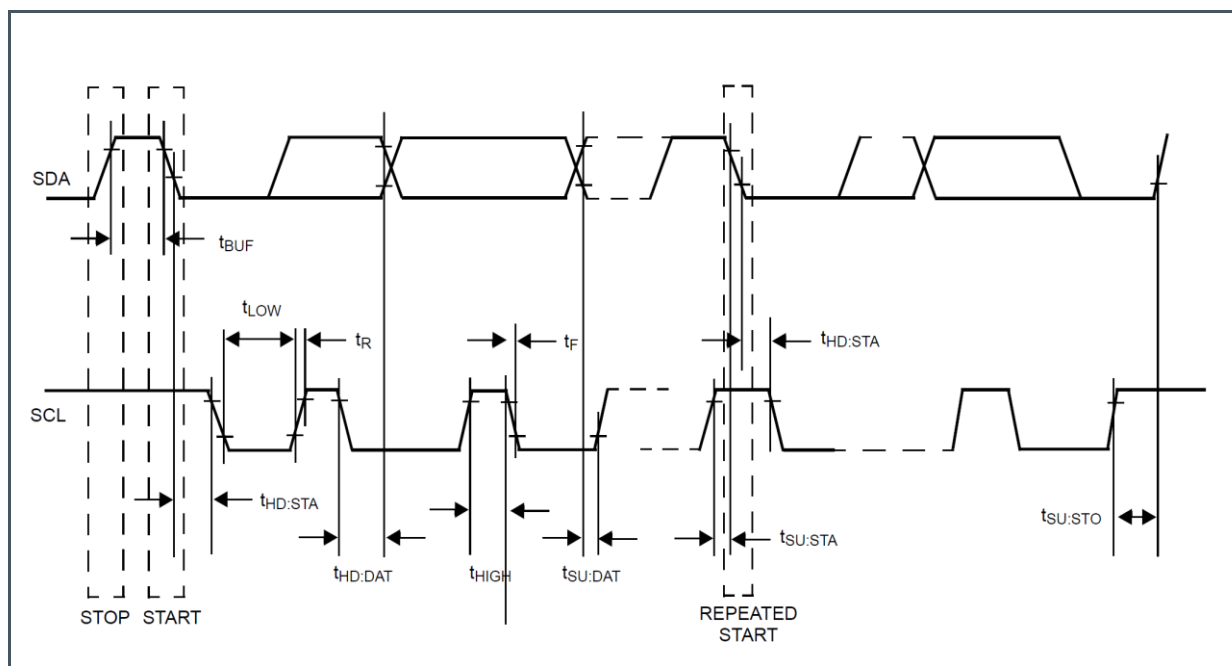
Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V _{NOISE_ECG}	Integrated noise	ADC sampling at 400 Hz; low pass filter set to 40 Hz, PPG channel operating in parallel		20		μA
CMRR _{ECG}	Common mode rejection ratio	Measured at 50 Hz and 100 Hz		73		dB
LED						
I _{LED}	Allowed operating LED current range		0		50	mA
		1/10 duty cycle @ 1 kHz			100	mA
V _{FLED}	Forward voltage	Green LED, add compliance voltage of LED driver, I _{LED} =10 mA, add compliance voltage of LED driver (V _{Dmin}) to obtain minimum voltage on the pin to drive the current at T _{AMB} =25 °C		3.1	33	V
		IR LED, I _{LED} =20 mA		1.4		
λ _p	Peak wavelength at I _{LED} =20mA	Green LED		527		nm
		Yellow LED		590		
		IR LED		940		
LED Driver						
I _{LED1/2/3/4}	LED output current range	LED current is adjustable with 10 bits – registers curr1/2/3/4	0		100	mA
	Tolerance	At 35 mA output current (currX[9:0]=166 h, X=1...4), VDD<5.0 V	-7		7	%
V _{Dmin}	Output voltage compliance			0.3		V
V _{Dmax}					5.5	V
Photodiode						
Re _{PD1-4}	Irradiance responsivity photodiode PD1...PD4	λ _P =550 nm, 4 photodiodes used pd1/2/3/4=1, gain_g=4x, gain_en=1, pd_ampres = 7 MΩ		45.9		mV/ (μW/cm ²)
	Irradiance responsivity photodiode B	λ _P =940 nm, gain_g=4x, gain_en=1, pd_ampres=7 MΩ		0.3		

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
I_d	Dark current	$E_e=0$, $T_{AMB}=25\text{ }^{\circ}\text{C}$	0		1	nA
I_{os}	Extrapolated offset current	$T_{AMB}=25\text{ }^{\circ}\text{C}$	-1		1	nA
ADC						
V_{ref}	ADC reference voltage			1.6		V
Nbit	Resolution		14			Bit
INL	Relative accuracy	$T_{AMB}=25\text{ }^{\circ}\text{C}$	-8		8	LSB
DNL ⁽³⁾	Differential nonlinearity	$T_{AMB}=25\text{ }^{\circ}\text{C}$		1.5		LSB
	Offset error	$T_{AMB}=25\text{ }^{\circ}\text{C}$	-8		8	LSB
	Gain error	$T_{AMB}=25\text{ }^{\circ}\text{C}$	-8		8	LSB
SNR	Signal-to-noise ratio	$F_{sample} = 1\text{ kHz}$, $F_{signal}=100\text{ Hz}$		80		dB
THD	Total harmonic distortion	$F_{sample} = 1\text{ kHz}$, $F_{signal}=100\text{ Hz}$		-70		dB
T_{conv}	Conversion rate	14-bit resolution			50	ksps
V_{in}	Input voltage range		0		V_{ref}	V
I²C Mode Timings						
f_{SCLK}	SCL Clock Frequency		0		400	kHz
t_{BUF}	Bus Free Time Between a STOP and START Condition		1.3			μs
$t_{HD:STA}$	Hold Time (Repeated) START Condition ⁽⁴⁾		0.6			μs
t_{LOW}	LOW Period of SCL Clock		1.3			μs
t_{HIGH}	HIGH Period of SCL Clock		0.6			μs

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$t_{SU:STA}$	Setup Time for a Repeated START Condition		0.6			μs
$t_{HD:DAT}$	Data Hold Time ⁽⁵⁾		0		0.9	μs
$t_{SU:DAT}$	Data Setup Time ⁽⁶⁾					
t_R	Rise Time of Both SDA and SCL Signals		20		300	ns
t_F	Fall Time of Both SDA and SCL Signals		20		300	ns
$t_{SU:STO}$	Setup Time for STOP Condition		0.6			μs
C_B	Capacitive Load for Each Bus Line	CB — total capacitance of one bus line in pF			500	pF
$C_{I/O}$	I/O Pin Capacitance (SDA, SCL)				10	pF

- (1) GPIO0-3 configured to draw minimum current (software dependent).
- (2) AS7026GG I²C interface active also in power down mode.
- (3) Specified only typical value for DNL to reduce production test time.
- (4) After this period, the first clock pulse is generated.
- (5) A device must internally provide a hold time of at least 300 ns for the SDA signal (referred to the VIHMIN of the SCL signal) to bridge the undefined region of the falling edge of SCL.
- (6) Fast-mode device can be used in a standard-mode system, but the requirement $t_{SU:DAT} = t_{SU:DAT} = 250$ ns must then be met. This is automatically the case if the device does not stretch the LOW period of the SCL signal. If such a device does stretch the LOW period of the SCL signal, it must output the next data bit to the SDA line $t_R \text{ max} + t_{SU:DAT} = 1000 + 250 = 1250$ ns before the SCL line is released.

Figure 7:
I²C Mode Timing Diagram

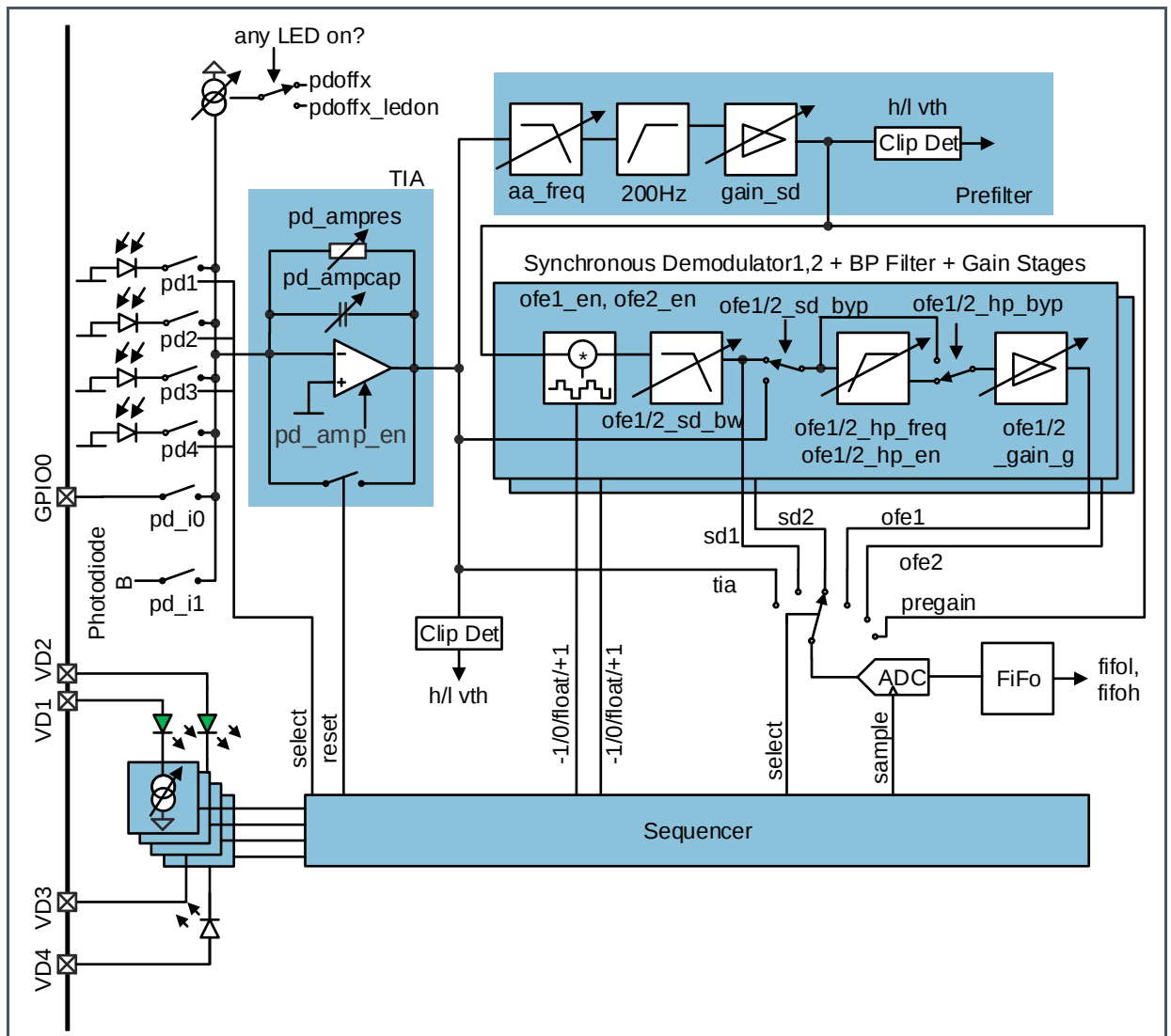


I²C Mode Timing Diagram: This figure shows the different timings required for I²C communication

6 Functional Description

6.1 Optical Analog Front End

Figure 8:
Optical Analog Front End



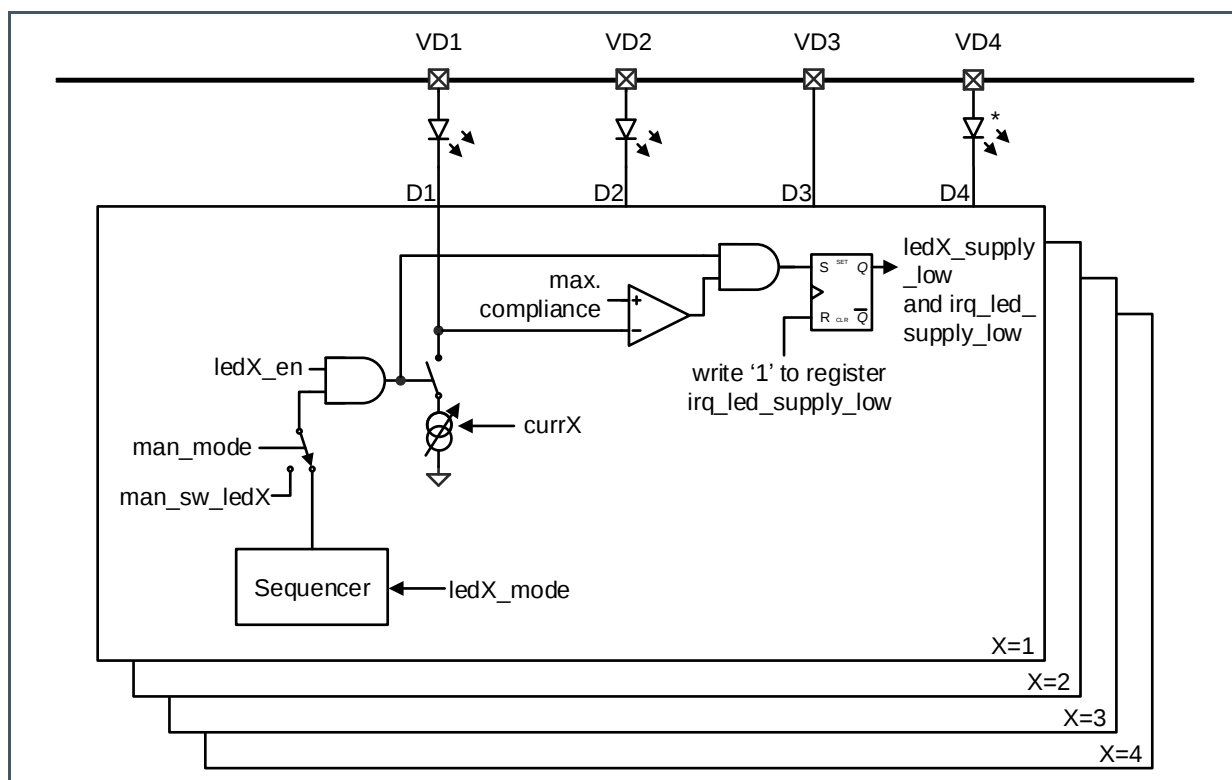
6.1.1 LEDs

Two green LEDs are used with anode on pin VD1 and VD2. A IR LED is connected with anode to pin VD4. VD3 allows direct access to the current sink 3

6.1.2 LED-Driver

The LED-driver outputs can be controlled manually or by the built in sequencer. See section 7.1.7 Optical Front End Operating Modes

Figure 9:
LED Drivers



7 Register Description

7.1 Register Overview

Figure 10:
Register Overview

Addr	Name	<D7>	<D6>	<D5>	<D4>	<D3>	<D2>	<D1>	<D0>
Register type 1									
0x10	LED_CFG	Not used	sigref_en	sigref_vo ltage[1]	sigref_vo ltage[0]	led4_en	led3_en	led2_en	led1_en
0x12	LED1_CURR L	Curr1[1:0]		Not used	Not used	Not used	Not used	Not used	Not used
0x13	LED1_CURR H	Curr1[9:2]							
0x14	LED2_CURR L	Curr2[1:0]		Not used	Not used	Not used	Not used	Not used	Not used
0x15	LED2_CURR H	Curr2[9:0]							
0x16	LED3_CURR L	Curr3[1:0]		Not used	Not used	Not used	Not used	Not used	Not used
0x17	LED3_CURR H	Curr3[9:2]							
0x18	LED4_CURR L	Curr4[1:0]		Not used	Not used	Not used	Not used	Not used	Not used
0x19	LED4_CURR H	Curr4[9:2]							
0x2C	LED12_MOD E	Man- sw_led2	Led2_mode[2:0]			Man_sw _led1	Led1_mode[2:0]		
0x2D	LED34_MOD E	Man- sw_led4	Led4_mode[2:0]			Man- sw_led3	Led3_mode[2:0]		
0x2E	MAN_SEQ_ CFG	man_mo de	man_sw _sdmult	man_sw _sdpol	man_sw _itg	diode_ctrl[2:0]			seq_en
0xA2	LEDSTATUS	Not used	Not used	Not used	Not used	led4_sup ply_low	led3_sup ply_low	led2_sup ply_low	led1_sup ply_low
0x1A	PD_CFG	Not used	Not used	pd4	pd3	pd2	pd1	pd_i1	pd_i0
0x1B	PDOFFX_LE DOFF	pdoffx_ledoff[7:0]							
0x1C	PDOFFX_LE DON	pdoffx_ledon[7:0]							
0x1D	PD_AMPRC CFG	pd_ampres[2:0]			pd_ampcap[4:0]				
0x1E	PD_AMPCF G	pd_amp_ en	pd_amp_ auto	pd_ampvo[3:0]				pd_ampcomp[1:0]	
0x1F	PD_THRES HCFG	pd_clipdetect_h_thresh[3:0]				pd_clipdetect_l_thresh[3:0]			
0x30	SEQ_CNT	seq_count[7:0]							

Addr	Name	<D7>	<D6>	<D5>	<D4>	<D3>	<D2>	<D1>	<D0>
0X31	SEQ_DIV	seq_div[7:0]							
0X32	SEQ_START	Not used	Not used	Not used	Not used	Not used	Not used	seq_start_sync	seq_start
0X33	SEQ_PER	seq_period[7:0]							
0X34	SEQ_LED_STA	seq_led_seq_led_stop_start[7:0]							
0X35	SEQ_LED_STO	seq_led_stop[7:0]							
0X36	SEQ_SECLD_STA	seq_secled_start[7:0]							
0X37	SEQ_SECLD_STO	seq_secled_stop[7:0]							
0X38	SEQ_ITG_STA	seq_itg_start[7:0]							
0X39	SEQ_ITG_STO	seq_itg_stop[7:0]							
0X3A	SEQ_SDP1_STA	seq_sdp1_start[7:0]							
0X3B	SEQ_SDP1_STO	seq_sdp1_stop[7:0]							
0X3C	SEQ_SDP2_STA	seq_sdp2_start[7:0]							
0X3D	SEQ_SDP2_STO	seq_sdp2_stop[7:0]							
0X3E	SEQ_SDM1_STA	seq_sdm1_start[7:0]							
0X3F	SEQ_SDM1_STO	seq_sdm1_stop[7:0]							
0X40	SEQ_SDM2_STA	seq_sdm2_start[7:0]							
0X41	SEQ_SDM2_STO	seq_sdm2_stop[7:0]							
0X42	SEQ_ADC	seq_adc[7:0]							
0X43	SEQ_ADC2TIA	seq_adc2tia[7:0]							
0X44	SEQ_ADC3TIA	seq_adc3tia[7:0]							
0X45	SD_SUBS	sd_subs[7:0]							
0X46	SEQ_CFG	Not used	Not used	Not used	Not used	Not used	Not used	Not used	sd_subs_always
0X47	SEQ_ERR	irq_adc_timing_error	Not used	Not used	Not used	Not used	Not used	Not used	Not used
0X60	CYC_COUNTER	cycle_counter[7:0]							
0X60	SEQ_COUNTER	sequence_counter[7:0]							
0X62	SUBS_COUNTER	subs_counter[7:0]							

Addr	Name	<D7>	<D6>	<D5>	<D4>	<D3>	<D2>	<D1>	<D0>
0X50	OFE_CFGA	ofe2_en	ofe1_en	en_bias_ ofe	aa_freq[1:0]		gain_sd[2:0]		
0X51	OFE_CFGB	sd_clipdetect_h_thresh[3:0]				sd_clipdetect_l_thresh[3:0]			
0X52	OFE_CFGC	Not used	prefilter_aa_byp	prefilter_hp_byp	prefilter_gain_byp	prefilter_bypass_en	prefilter_aa_en	prefilter_hp_en	prefilter_gain_en
0X53	OFE_CFGD	Not used	Not used	Not used	Not used	Not used	Not used	ofe_gs_aa[1:0]	
0X54	OFE1_CFGA	ofe1_sd_pol_init	ofe1_sd_en	ofe1_hp_en	ofe1_gai_n_en	ofe1_sd_byp	ofe1_hp_byp	ofe1_gai_n_byp	ofe1_sd_hld
0X55	OFE1_CFGB	Not used	ofe1_gain_g[2:0]			ofe1_sd_bw[1:0]		ofe1_hp_freq[1:0]	
0X58	OFE2_CFGA	ofe2_sd_pol_init	ofe2_sd_en	ofe2_hp_en	ofe2_gai_n_en	ofe2_sd_byp	ofe2_hp_byp	ofe2_gai_n_byp	ofe2_sd_hld
0X59	OFE2_CFGB	Not used	ofe2_gain_g[2:0]			ofe2_sd_bw[1:0]		2fe2_hp_freq[1:0]	
0X20	LTFDATA0_L	ltfdata0[7:0]							
0X21	LTFDATA0_H	ltfdata0[15:8]							
0X22	LTFDATA1_L	ltfdata0[7:0]							
0X23	LTFDATA1_H	ltfdata1[15:8]							
0X24	ITIME	itime[7:0]							
0X25	LTF_CONFIG	infinite_iti me	az_disab le_auto	reserved	reserved	Not used	Not used	ltf_fifo_m ode	ltf_enabl e
0X26	LTF_SEL	Not used	ltf1_sel[2:0]			Not used	ltf0_sel[2:0]		
0X27	LTF_GAIN	Do not use	Do not use	itime_unit[1:0]		ltf_gain[3:0]			
0X28	LTF_CONTR OL	Do not use	Do not use	Do not use	Do not use	Do not use	Do not use	Do not use	ltf_start
0X29	AZ_CONTR OL	Do not use	Do not use	Do not use	Do not use	Do not use	Do not use	az_enabl e_1	az_enabl e_0
0X2A	OFFSET0	offset0[7:0]							
0X2B	OFFSET1	offset0[7:0]							
0X70	AFE_CFG	Do not use	Do not use	Do not use	Do not use	afe_ena b	afe_enab _dac	afe_enab _dac_buf	afe_enab _gainsta ge
0X80	EAF_GST	gpio_gst_in[2:0]			gst_ref[1:0]		gst_gain[2:0]		
0X81	EAF_BIAS	gpio_r_bias[2:0]			Not used	Not used	Not used	Not used	Not used
0X82	EAF_DAC	Do not use	Do not use	Do not use	sigref_on _dac_buf	measure _dac	gpio_dac[2:0]		
0X83	EAF_DAC1_L	dac1_value[]		Not used	Not used	Not used	Not used	Not used	Not used
0X84	EAF_DAC1_H	dac1_value[9:2]							
0X85	EAF_DAC2_L	dac2_value[1:0]		Not used	Not used	Not used	Not used	Not used	Not used
0X86	EAF_DAC2_H	dac2_value[9:2]							

Addr	Name	<D7>	<D6>	<D5>	<D4>	<D3>	<D2>	<D1>	<D0>
0X87	EAF_DAC_CFG	Not used	Not used	Not used	Not used	Not used	Not used	dac_mode[1:0]	
0X5C	ECG_CFGA	ecg_en	Not used	ecg_lp_en	ecg_hp_en	ecg_gain_en	ecg_lp_byp	ecg_hp_byp	ecg_gain_byp
0X5D	ECG_CFGB	Not used	ecg_lp_freq[1:0]		ecg_hp_freq[1:0]		ecg_gain_g[2:0]		
0X5E	ECG_CFGC	Not used	Not used	Not used	Not used	Not used	Not used	ecg_low_leakage_en	ecg_ref_en
0X5F	ECG_CFGD	Not used	Not used	Not used	ecg_lead_sdet_syn_c_adc	ecg_lead_sdet_pol	ecg_lead_sdet_curr[1:0]		ecg_lead_sdet_en
0X68	ADC_THRE_SHOULD	adc_threshold[7:0]							
0X69	ADC_THRE_SHOULD_CFG	Not used	Not used	Not used	Not used	Not used	Not used	adc_thresh_diffential	adc_thresh_tiaonly
0X88	ADC_CFGA	Not used	Not used	Not used	Not used	adc_multi_n[2:0]			adc_multi_mode
0X89	ADC_CFGB	Not used	Not used	adc_clock[2:0]			adc_calibration	ulp	adc_en
0X8A	ADC_CFGC	Not used	Not used	Not used	adc_self_pd	adc_disc_harge	adc_settling_time[2:0]		
0X8B	ADC_CHANNEL_MASK_L	adc_channel_mask_pregain	adc_channel_mask_afe	adc_channel_mask_temp	adc_channel_mask_sd2	adc_channel_mask_ofe2	adc_channel_mask_sd1	adc_channel_mask_ofe1	adc_channel_mask_tia
0X8C	ADC_CHANNEL_MASK_H	Not used	Not used	Not used	Not used	adc_channel_mask_gpio2	adc_channel_mask_gpio3	adc_channel_mask_ecgi	adc_channel_mask_ecgo
0X8E	ADC_DATA_L	adc_data[7:0]							
0X8F	ADC_DATA_H	Not used	Not used	adc_data[13:8]					
0X78	FIFO_CFG	Not used	Not used	fifo_threshold[5:0]					
0X79	FIFO_CNTRL	Not used	Not used	Not used	Not used	Not used	Not used	Not used	fifo_clear
0XA3	FIFOSTATUS	fifooverflow	FifoLevel[6:0]						
0XFE	FIFOL	FifoL[7:0]							
0XFF	FIFOH	FifoH[7:0]							
0x00	CONTROL	Not used	Not used	Not used	Not used	Not used	Not used	osc_en	ldo_en
0X08	GPIO_A	Not used	Not used	Not used	Not used	gpio3_a	gpio2_a	gpio1_a	gpio0_a
0X09	GPIO_E	Not used	Not used	Not used	Not used	gpio3_e	gpio2_e	gpio1_e	gpio0_e
0X0A	GPIO_O	Not used	Not used	Not used	Not used	gpio3_o	gpio2_o	gpio1_o	gpio0_o
0X0B	GPIO_I	Not used	Not used	Not used	Not used	gpio3_i	gpio2_i	gpio1_i	gpio0_i
0X0C	GPIO_P	gpio3_pd	gpio3_pu	gpio2_pd	gpio2_pu	gpio1_pd	gpio1_pu	gpio0_pd	gpio0_pu
0X0D	GPIO_SR	Not used	Not used	Not used	Not used	gpio3_sr	gpio2_sr	gpio1_sr	gpio0_sr
0X91	SUBID	subid[4:0]					Revision[2:0]		
0X92	ID	id[5:0]						id_reserved[1:0]	

Addr	Name	<D7>	<D6>	<D5>	<D4>	<D3>	<D2>	<D1>	<D0>
0XA0	STATUS	irq_led_supply_low	irq_clipdetect	irq_fifooverflow	irq_fifothreshold	irq_adc_threshold	irq_ltf	irq_sequencer	irq_adc
0XA1	CLIPSTATUS	Not used	Not used	Not used	Not used	pd_clipdetect_l	pd_clipdetect_h	sd_clipdetect_l	sd_clipdetect_h
0XA2	LEDSTATUS	Not used	Not used	Not used	Not used	led4_supply_low	led3_supply_low	led2_supply_low	led1_supply_low
0XA8	INTENAB	irq_led_supply_low_enab	irq_clipdetect_enab	irq_fifooverflow_ena	irq_fifothreshold_enab	irq_adc_threshold_enab	irq_ltf_enab	irq_sequencer_enab	irq_adc_enab
0XA9	INTR	irq_led_supply_low_intr	irq_clipdetect_intr	irq_fifooverflow_intr	irq_fifothreshold_intr	irq_adc_threshold_intr	irq_ltf_intr	irq_sequencer_intr	irq_adc_intr

7.1.1.1 LED Configuration

LED_CFG Register (Address 0x10)

Figure 11:
LED_CFG Register

Addr: 0x10		LED_CFG			
Bit	Bit Name	Default	Access	Bit Description	
6	sigref_en	0	RW	Signal reference: Is required for all analog blocks (except PD_Amp or light-to-frequency operation) 0 ... Disable signal reference 1 ... Enable signal reference	
5:4	sigref_voltage	0	RW	Voltage setting of SIGREF – datasheet parameters are guaranteed only for default value of 0.6 V.	
				Setting	IMAX
				00	0.6 V (default)
				01	0.7 V
				10	0.8 V
				11	0.9 V
3	led4_en	0	RW	0 ... Disables LED4 output source. 1 ... Enables LED4 output source.	
2	led3_en	0	RW	0 ... Disables LED3 output source. 1 ... Enables LED3 output source.	
1	led2_en	0	RW	0 ... Disables LED2 output source. 1 ... Enables LED2 output source.	
0	led1_en	0	RW	0 ... Disables LED1 output source. 1 ... Enables LED1 output source.	

The LED_CURR defines the LED output current.



Attention

It is recommended to configure the current only when the output is not active, as there is no latch implemented to keep the 10 bits consistent. New values are applied directly and immediately.

LED1_CURRL Register (Address 0x12)**Figure 12:**
LED1_CURRL Register

Addr: 0x12		LED1_CURRL		
Bit	Bit Name	Default	Access	Bit Description
7:6	Curr1[1:0]	0	RW	LED1 output current lower 2 bits
5:0	Not used	0	RW	Not used

LED1_CURRH Register (Address 0x13)**Figure 13:**
LED1_CURRH Register

Addr: 0x13		LED1_CURRH		
Bit	Bit Name	Default	Access	Bit Description
7:0	Curr1[9:2]	0	RW	LED1 output current upper 8 bits Coding for curr1[9:0]: 000h ... 786 μ A 001h ... 883 μ A (1 LSB=97 μ A) 002h ... 980 μ A 166h ... 35 mA 3FFh ... 100 mA

LED2_CURRL Register (Address 0x14)**Figure 14:**
LED2_CURRL Register

Addr: 0x14		LED2_CURRL		
Bit	Bit Name	Default	Access	Bit Description
7:6	Curr2[1:0]	0	RW	LED2 output current lower 2 bits
5:0	Not used	0	RW	Not used

LED2_CURRH Register (Address 0x15)**Figure 12:**
LED2_CURRH Register

Addr: 0x15		LED2_CURRH		
Bit	Bit Name	Default	Access	Bit Description
7:0	Curr2[9:2]	0	RW	LED2 output current upper 8 bits Coding for curr1[9:0]: 000h ... 786 μ A 001h ... 883 μ A (1 LSB=97 μ A) 002h ... 980 μ A 166h ... 35 mA 3FFh ... 100 mA

LED3_CURRL Register (Address 0x16)**Figure 13:**
LED3_CURRL Register

Addr: 0x16		LED3_CURRL		
Bit	Bit Name	Default	Access	Bit Description
7:6	Curr3[1:0]	0	RW	LED3 output current lower 2 bits
5:0	Not used	0	RW	Not used

LED3_CURRH Register (Address 0x17)**Figure 14:**
LED3_CURRH Register

Addr: 0x17		LED3_CURRH		
Bit	Bit Name	Default	Access	Bit Description
7:0	Curr3[9:2]	0	RW	LED3 output current upper 8 bits Coding for curr1[9:0]: 000h ... 786 μ A 001h ... 883 μ A (1 LSB=97 μ A) 002h ... 980 μ A 166h ... 35 mA 3FFh ... 100 mA

LED4_CURRL Register (Address 0x18)**Figure 15:**
LED4_CURRL Register

Addr: 0x18		LED4_CURRL		
Bit	Bit Name	Default	Access	Bit Description
7:6	Curr4[1:0]	0	RW	LED4 output current lower 2 bits
5:0	Not used	0	RW	Not used

LED4_CURRH Register (Address 0x19)**Figure 16:**
LED4_CURRH Register

Addr: 0x19		LED4_CURRH		
Bit	Bit Name	Default	Access	Bit Description
7:0	Curr4[9:2]	0	RW	LED4 output current upper 8 bits Coding for curr1[9:0]: 000h ... 786 μ A 001h ... 883 μ A (1 LSB=97 μ A) 002h ... 980 μ A 166h ... 35 mA 3FFh ... 100 mA

LED12_MODE Register (Address 0x2c)**Figure 17:**
LED12_MODE Register

Addr: 0x2c		LED12_MODE		
Bit	Bit Name	Default	Access	Bit Description
7	Man-sw_led2	0	RW	0 ... LED output D2 disabled. (High impedance) 1 ... LED output D2 enabled
6.4	Led2_mode	0	RW	LED2 mode
				Settings
				Behavior
				000 Always OFF
				001 Always ON when sequencer is active

Addr: 0x2c		LED12_MODE		
Bit	Bit Name	Default	Access	Bit Description
				010 Controlled by sequencer
				011 Controlled by sequencer, only ON in even iterations: 0, 2, 4 etc.
				100 Controlled by sequencer, only ON in odd iterations: 1, 3, 5 etc.
				101 Controlled by sequencer, only ON in every fourth iteration, starting at 1: 1, 5, 9 etc.
				110 Controlled by sequencer: secondary LED timing
				111 Do not use
3	Man_sw_led1	0	RW	0 ... LED output D1 disabled. (High impedance) 1 ... LED output D1 enable
2.0	Led1_mode	0	RW	LED1 mode
				Settings Behavior
				000 Always OFF
				001 Always ON when sequencer is active
				010 Controlled by sequencer
				011 Controlled by sequencer, only ON in even iterations: 0, 2, 4 etc.
				100 Controlled by sequencer, only ON in odd iterations: 1, 3, 5 etc.
				101 Controlled by sequencer, only ON in every fourth iteration, starting at 1: 1, 5, 9 etc.
				110 Controlled by sequencer: secondary LED timing
				111 Do not use

LED34_MODE Register (Address 0x2d)**Figure 18:**
LED34_MODE Register

Addr: 0x2d		LED34_MODE		
Bit	Bit Name	Default	Access	Bit Description
7	Man-sw_led4	0	RW	0 ... LED output D4 disabled. (High impedance) 1 ... LED output D4 enabled
6.4	Led4_mode	0	RW	LED4 mode
				Settings Behavior
				000 Always OFF
				001 Always ON when sequencer is active
				010 Controlled by sequencer
				011 Controlled by sequencer, only ON in even iterations: 0, 2, 4 etc.
				100 Controlled by sequencer, only ON in odd iterations: 1, 3, 5 etc.
				101 Controlled by sequencer, only ON in every fourth iteration, starting at 1: 1, 5, 9 etc.
				110 Controlled by sequencer: secondary LED timing
				111 Do not use
3	Man_sw_led3	0	RW	0 ... LED output D3 disabled. (High impedance) 1 ... LED output D3 enable
2.0	Led3_mode	0	RW	LED3 mode
				Settings Behavior
				000 Always OFF
				001 Always ON when sequencer is active
				010 Controlled by sequencer
				011 Controlled by sequencer, only ON in even iterations: 0, 2, 4 etc.
				100 Controlled by sequencer, only ON in odd iterations: 1, 3, 5 etc.
				101 Controlled by sequencer, only ON in every fourth iteration, starting at 1: 1, 5, 9 etc.
				110 Controlled by sequencer: secondary LED timing
				111 Do not use

The MAN_SEQ_CFG register is used to configure the operation of the optical front end

MAN_SEQ_CFG Register (Address 0x2e)

Figure 19:
MAN_SEQ_CFG Register

Addr: 0x2e		MAN_SEQ_CFG			
Bit	Bit Name	Default	Access	Bit Description	
7	man_mode	0	RW	0 ... Enables Sequencer 1 ... Enables Manual control of optical front end	
6	man_sw_sd_mult	0	RW	If man_mode=1 0 ... Disables synchronous demodulator multiplication 1 ... Enables synchronous demodulator multiplication	
5	man_sw_sd_pol	0	RW	If man_mode=1 0 ... Negative polarity in synchronous demodulator multiplication 1 ... Positive polarity in synchronous demodulator multiplication	
4	man_sw_itg	0	RW	If man_mode=1 0 ... All integrator capacitors are shorted. Integrator is reset 1 ... Integrator capacitors are charging up. Integrator is running	
3:1	diode_ctrl	0	RW	Connection of Photodiodes PD1, PD2, PD3, PD4 to the photodiode amplifier. 0 ... PD1-PD4 are connected 1 ... PD1 synchronous to LED1, PD2 sync/to LED2 PD3 sync/to LED3, PD4 sync/to LED4 2 ... PD1 synchronous to LED1, PD2 sync/to LED1 PD3 sync/to LED2, PD4 sync/to LED2 3 ... PD1 synchronous to LED1, PD2 sync/to LED1 PD3 sync/to LED4, PD4 sync/to LED4 4 ... SPO2 mode *(obsolete): (negedge(sdm1) or negedge(sdp1)) - PD1=0 PD2=0 PD3=1 PD4=1; (negedge(sdm2) or negedge(sdp2)) - PD1=1 PD2=1 PD3=0 PD4=0 Note that PD_CFG.pdX takes precedence - to turn OFF one photo diode, the respective bit has to be de-asserted in the PD_CFG register.	
				PD_CFG.pdX	diode_ctrl
				Photo Diode 1	Photo Diode 2
				Photo Diode 3	Photo Diode 4
				0	xx
				1	0
				OFF	OFF
				ON	ON

Addr: 0x2e		MAN_SEQ_CFG							
Bit	Bit Name	Default	Access	Bit Description					
				1	1	LED1	LED2	LED3	LED4
				1	2	LED1	LED1	LED2	LED2
				1	3	LED1	LED1	LED4	LED4
				1	4	SPO2 mode (obsolete)			
				1	5..7	Do not use			
0	seq_en	0	RW	0 ... Disables sequencer 1 ... Enables sequencer					

LEDSTATUS Register (Address 0xa2)

Figure 20:
LEDSTATUS Register

Addr: 0xa2		LED4_CURRL							
Bit	Bit Name	Default	Access	Bit Description					
7:4	NA	0	RO	Not Used					
3	led4_supply_low	0	RO	If this bit is asserted, LED4 voltage has been too low.					
2	led3_supply_low	0	RO	If this bit is asserted, LED3 voltage has been too low.					
1	led2_supply_low	0	RO	If this bit is asserted, LED2 voltage has been too low.					
0	led1_supply_low	0	RO	If this bit is asserted, LED1 voltage has been too low.					

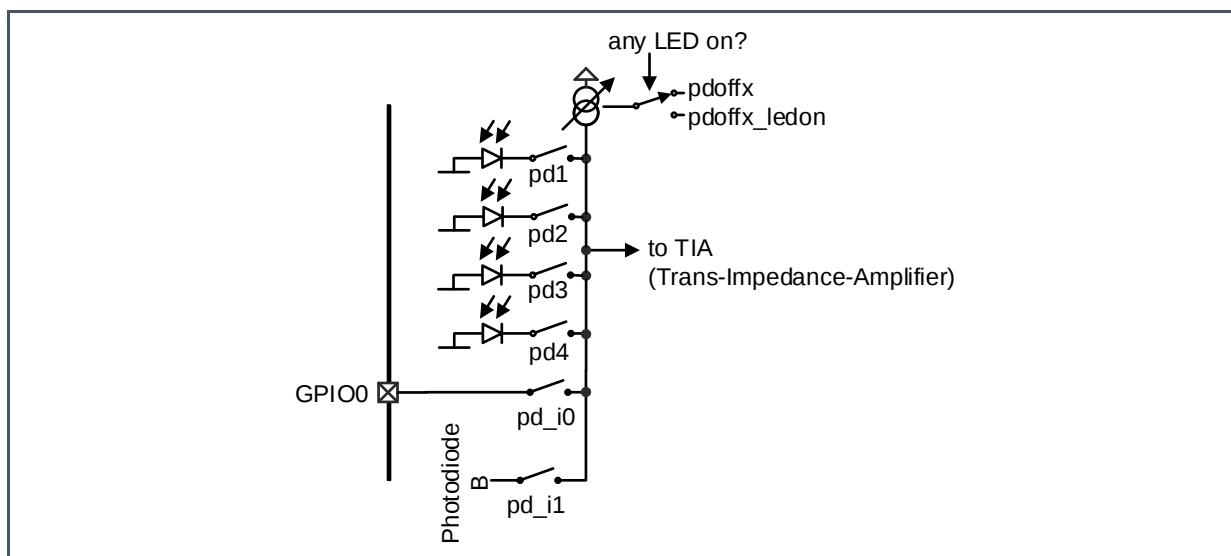
An asserted bit can be cleared by writing a '1' to the irq_led_supply_low bit.

7.1.2 Photodiode Selection

In order to have flexible arrangement of the use photodiodes, PD1-PD4 can be individually connected to the photodiode amplifier input. The optional offset current allows cancellation of constant light sources like sunlight. In case of an external photodiode or any other sensor with (low) current output, the pins GPIO0 and GPIO1 can be used as input.

Additionally the sequencer can control the diodes – see diode_ctrl described in register MAN_SEQ_CFG.

Figure 21:
Photodiode Selection



PD_CFG Register (Address 0x1a)

Figure 22:
PD_CFG Register

Addr: 0x1a		PD_CFG		
Bit	Bit Name	Default	Access	Bit Description
7:6	NA	0	RW	Not Used
5	pd4	0	RW	0 ... Photodiode PD4 is disconnected from photo amplifier 1 ... Photodiode PD4 is connected to photo amplifier (as defined in diode_ctrl)
4	pd3	0	RW	0 ... Photodiode PD3 is disconnected from photo amplifier 1 ... Photodiode PD3 is connected to photo amplifier (as defined in diode_ctrl)
3	pd2	0	RW	0 ... Photodiode PD2 is disconnected from photo amplifier 1 ... Photodiode PD2 is connected to photo amplifier (as defined in diode_ctrl)
2	pd1	0	RW	0 ... Photodiode PD1 is disconnected from photo amplifier 1 ... Photodiode PD1 is connected to photo amplifier (as defined in diode_ctrl)

Addr: 0x1a		PD_CFG		
Bit	Bit Name	Default	Access	Bit Description
1	pd_i1	0	RW	0... Photodiode B (see Photodiode Characteristics) disconnected from TIA input 1... Photodiode B (see Photodiode Characteristics) connected to TIA input; set ltf1_sel=0 and ltf2_sel=0.
0	pd_i0	0	RW	0 ... GPIO0-input is disconnected from photo amplifier 1 ... GPIO0-input is connected to photo amplifier; set gpio_a[0]=1.

PDOFFX_LEDOFF Register (Address 0x1b)

Figure 23:
PDOFFX_LEDOFF Register

Addr: 0x1b		PDOFFX_LEDOFF		
Bit	Bit Name	Default	Access	Bit Description
7:0	pdoffx_ledoff	0	RW	Input offset current if all LEDs are OFF (all sw_led* sequencer outputs are zero) loffset = pdoffx_ledoff*10 nA 0 ... Offset source is turned OFF

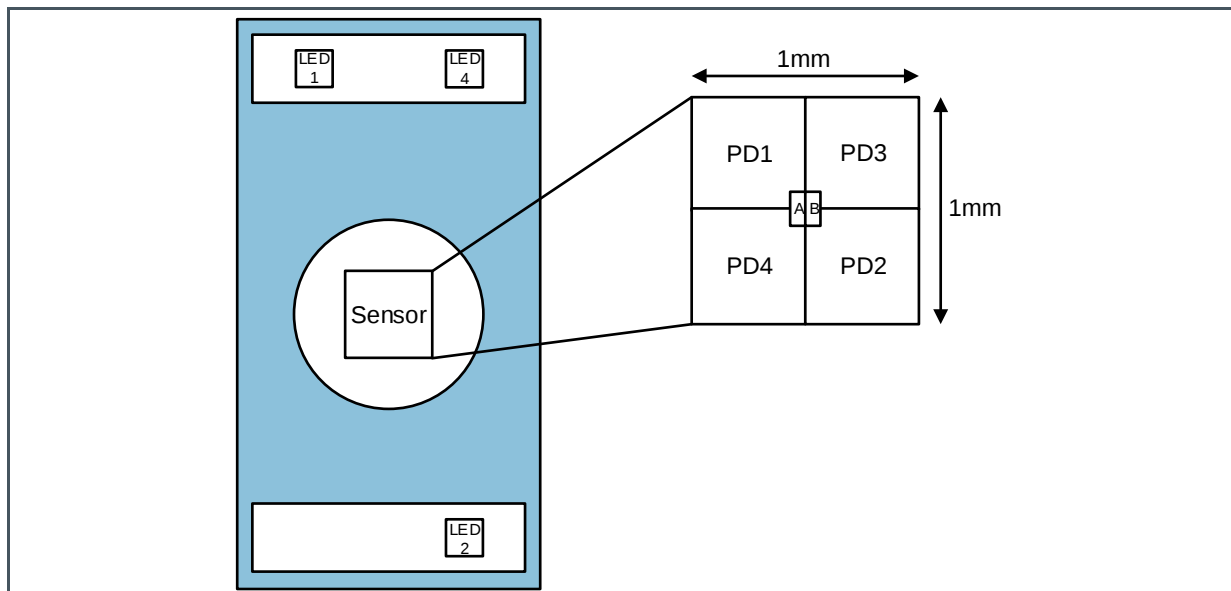
PDOFFX_LEDON Register (Address 0x1c)

Figure 24:
PDOFFX_LEDON Register

Addr: 0x1c		PDOFFX_LEDON		
Bit	Bit Name	Default	Access	Bit Description
7:0	pdoffx_ledon	0	RW	Input offset current if at least one LED is ON (one or more sw_led* sequencer outputs are non-zero) loffset = pdoffx_ledon*10nA 0 ... Offset source is turned OFF

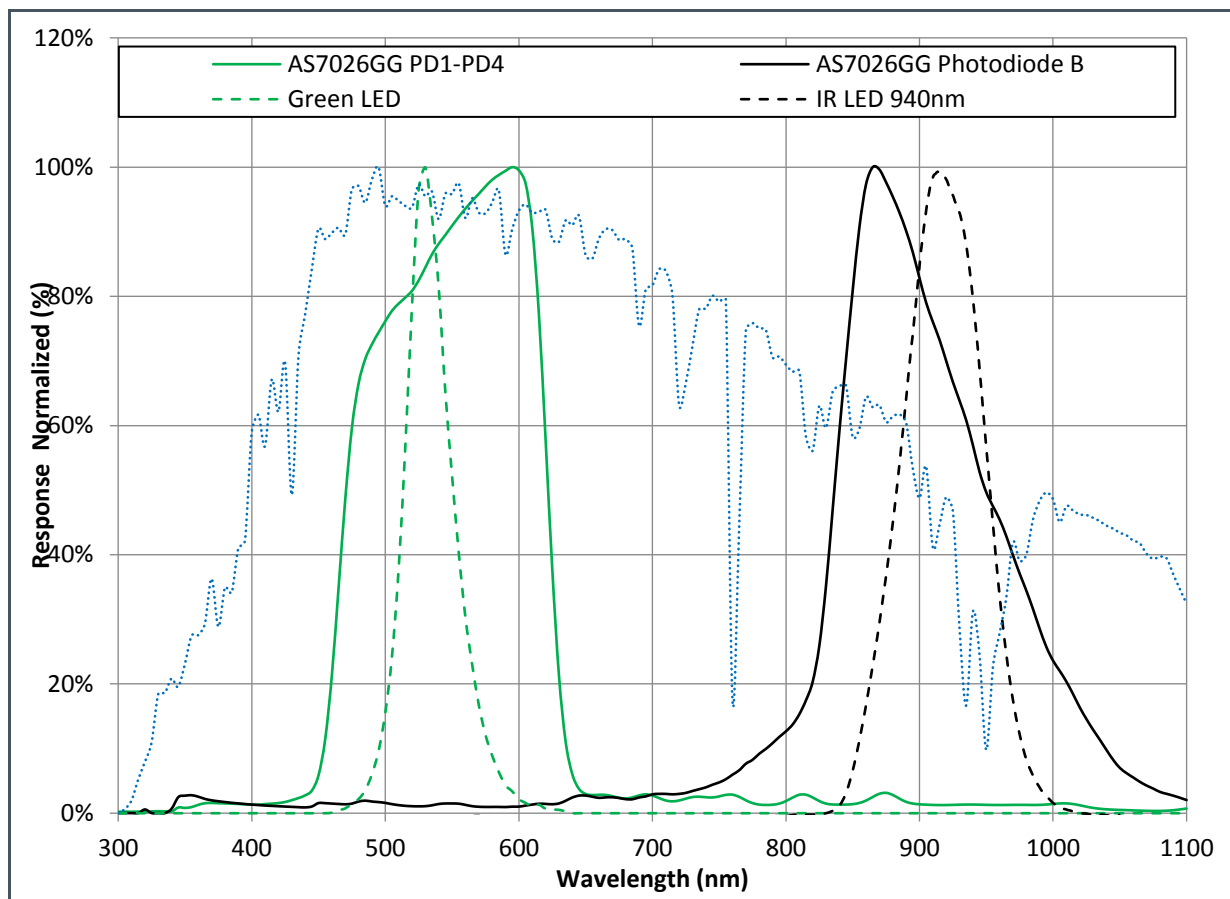
7.1.3 Photodiode Characteristics

Figure 25:
Photodiode Arrangement –Orientation as in Figure 2



For operation and characteristics of photodiode 'A' and photodiode 'B' see section 7.1.12 Light-to-Frequency Mode.

Figure 26:
Photodiode Sensitivity (solid green and black) and LED Emission Spectrum (dotted green and dotted black)



Information

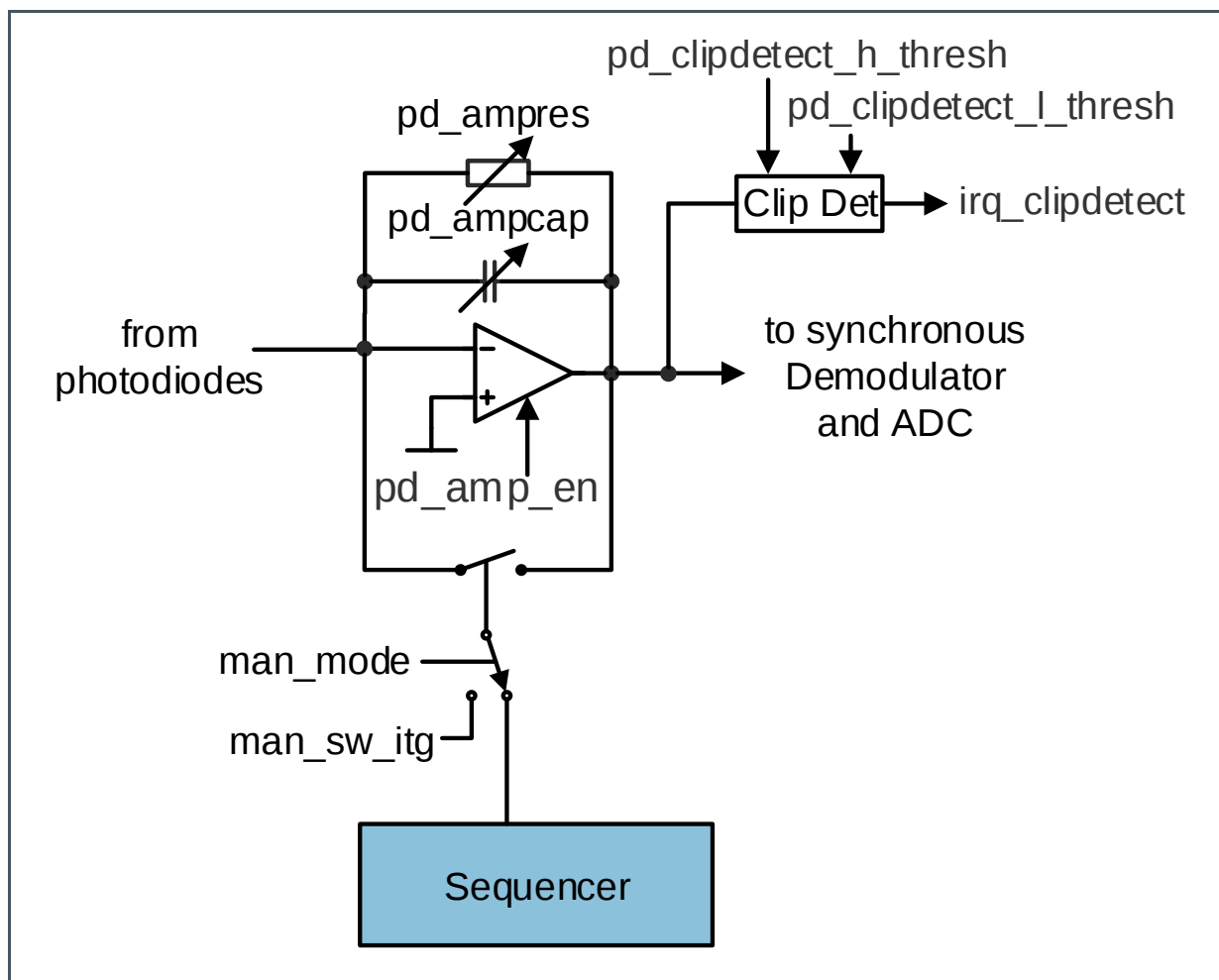
All 4 photodiodes used $pd1/2/3/4=1$; perpendicular light source and no diffuser used on AS7026GG; due to the difference in photodiode size the absolute response for Photodiode B (0.01 mm^2) is much lower compared to PD1-PD4 (0.8 mm^2)

7.1.4 Photodiode Transimpedance Amplifier (TIA)

The photodiode amplifier can be configured in three different modes:

- Photocurrent to frequency converter – see section 7.1.12 Light-to-Frequency Mode
- Photocurrent to voltage converter
- Photocurrent integrator

Figure 27:
Transimpedance Amplifier (TIA)



The integration time t_{INT} is defined either by the sequencer (`man_mode=0`) or manually through the bit `sw_itg` if `man_mode=1`.

Use following settings for the programming of the TIA:

Figure 28:
TIA Programming

pd_ampres	pd12341	pd_ampcap	pd_ampcomp	pd_ampvo	gain
1	1...4	13	1	15	1 V/μA
2	1...4	7	1	15	2 V/μA
3	1...4	5	1	15	3 V/μA
4	1...2	2	0	15	5 V/μA
	3...4	3			
5	1...2	2	0	15	7 V/μA
	3...4	3			
6	1	1	0	15	10 V/μA
	2...4	2			
7	1...2	1	0	15	15 V/μA
	3...4	2			
Low Bandwidth Mode					
5	1...4	31	3	15	7 V/μA
Integrating Mode (pd_ampres=0)					
0	1...4	10	3	15	1 V/pQ
0	1...4	20	3	15	1/2V/pQ
0	1...4	30	3	15	1/3V/pQ

7.1.5 Photodiode TIA Registers

PD_AMPRCCFG Register (Address 0x1d)

Figure 29:
PD_AMPRCCFG Register

Addr: 0x1d		PD_AMPRCCFG																				
Bit	Bit Name	Default	Access	Bit Description																		
				Feedback resistor																		
				<table><tr><th>Setting</th><th>Resistance</th></tr><tr><td>0</td><td>No resistor in feedback of amplifier – photocurrent integrator</td></tr><tr><td>1</td><td>1 MΩ</td></tr><tr><td>2</td><td>2 MΩ</td></tr><tr><td>3</td><td>3 MΩ</td></tr><tr><td>4</td><td>5 MΩ</td></tr><tr><td>5</td><td>7 MΩ</td></tr><tr><td>6</td><td>10 MΩ</td></tr><tr><td>7</td><td>15 MΩ</td></tr></table>	Setting	Resistance	0	No resistor in feedback of amplifier – photocurrent integrator	1	1 MΩ	2	2 MΩ	3	3 MΩ	4	5 MΩ	5	7 MΩ	6	10 MΩ	7	15 MΩ
Setting	Resistance																					
0	No resistor in feedback of amplifier – photocurrent integrator																					
1	1 MΩ																					
2	2 MΩ																					
3	3 MΩ																					
4	5 MΩ																					
5	7 MΩ																					
6	10 MΩ																					
7	15 MΩ																					
7:5	pd_ampres	0	RW																			
4:0	pd_ampcap			Feedback capacitor																		

The PD_AMPCFG register is used to configure the operating mode of the photoamplifier.

PD_AMPCFG Register (Address 0x1e)

Figure 30:
PD_AMPCFG Register

Addr: 0x1e		PD_AMPCFG		
Bit	Bit Name	Default	Access	Bit Description
7	pd_amp_en	0	RW	0 ... Activates power down mode of photo-amplifier 1 ... Enables photo-amplifier (direct or automatic pd_amp_auto mode) also set en_bias_ofe=1
6	pd_amp_auto	0	RW	0 ... Normal TIA mode 1 ... Enable TIA only when seq_itg is set (i.e. controlled by sequencer itg setting) also set en_bias_ofe=1

Addr: 0x1e		PD_AMPCFG		
Bit	Bit Name	Default	Access	Bit Description
5:2	pd_ampvo	1	RW	OpAmp offset. Can be used to limit signal in darkness and to shorten rise times
1:0	pd_ampcomp	3	RW	OpAmp compensation, depending on gain and number of used photo diodes Capacitor = $pd_ampcap \cdot 0.1pF$

PD_THRESHCFG Register (Address 0x1f)

Figure 31:
PD_THRESHCFG Register

Addr: 0x1f		PD_THRESHCFG		
Bit	Bit Name	Default	Access	Bit Description
7:4	pd_clipdetect_h_thresh	0	RW	If the voltage on the output of the TIA exceed this threshold the irq_clipdetect interrupt is asserted. The threshold is defined as 0 ... 1824 mV 1 ... 1748 mV 2 ... 1672 mV 3 ... 1596 mV 4 ... 1520 mV 5 ... 1444 mV 6 ... 1368 mV 7 ... 1292 mV 8 ... 1216 mV 9 ... 1140 mV 10 ... 1064 mV 11... 988 mV 12 ... 912 mV 13 ... 836 mV 14 ... 760 mV 15 ... 684 mV
3:0	pd_clipdetect_l_thresh	0	RW	If the voltage on the output of the TIA falls below this threshold the irq_clipdetect interrupt is asserted. The threshold is defined as 0 ... 67 mV 1 ... 143 mV 2 ... 219 mV

Addr: 0x1f		PD_THRESHCFG		
Bit	Bit Name	Default	Access	Bit Description
				3 ... 295 mV
				4 ... 371 mV
				5 ... 447 mV
				6 ... 523 mV
				7 ... 599 mV
				8 ... 675 mV
				9 ... 751 mV
				10 ... 827 mV
				11 ... 903 mV
				12 ... 979 mV
				13 ... 1055 mV
				14 ... 1131 mV
				15 ... 1207 mV

7.1.6 Voltage Mode of the Photodiode Amplifier

The output voltage of the photodiode amplifier is depending on the feedback component.

Equation 1:

$$U_{out} = I_{photo} \cdot R_{fb} \quad \text{Feedback resistor}$$

Equation 2:

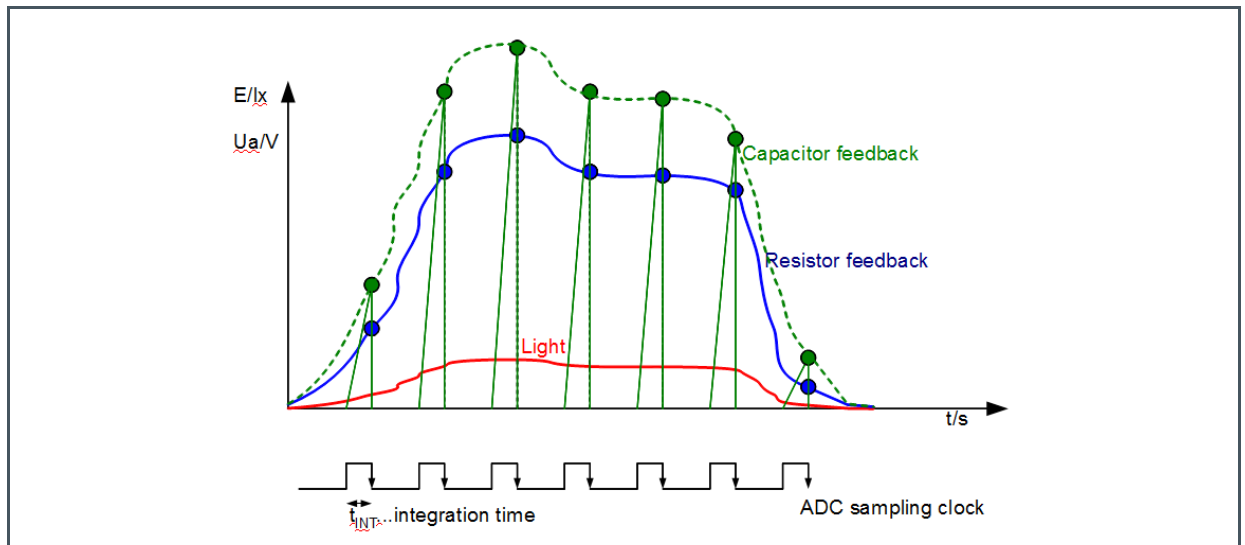
$$U_{out} = I_{photo} \cdot \frac{t_{INT}}{C_{fb}} \quad \text{Feedback capacitor}$$



Information

The integration time t_{INT} is defined either by the sequencer ($man_mode=0$) or manually through the bit sw_itg if $man_mode=1$. For the synchronous demodulator only use the resistive feedback.

Figure 32:
Difference Between Resistive and Capacitive Feedback



- (1) **Green:** Capacitive Integration
Green Dotted: Effective Value from Capacitive Mode
Blue: Resistive Feedback
Red: Light Intensity

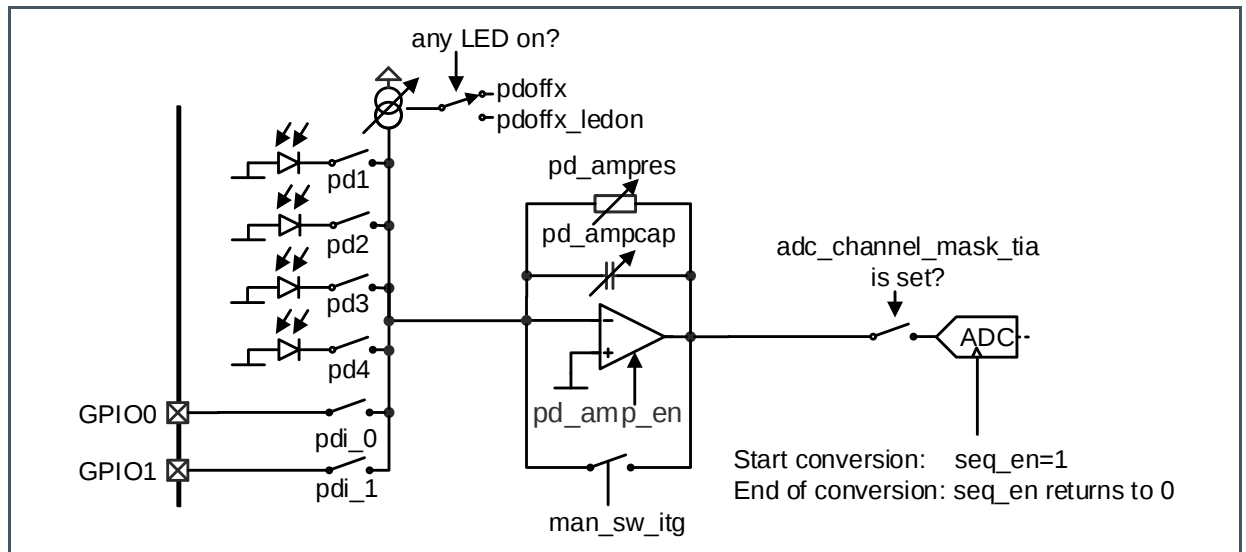
7.1.7 Optical Front End Operating Modes

Once the photodiode amplifier is configured the measurement can be done in two different ways. Either the LED-outputs, the photodiode amplifier and the ADC are controlled manually by means of register bits, or they are controlled by a built in sequencer.

Manual Operation of the Optical Frontend:

The optical front end can be manually controlled via the register `man_mode=1`

Figure 33:
Optical Frontend



(1) Applies only if man_mode=1.

For manual operation of the LEDs and its current sinks see 6.1.2 LED-Driver

7.1.8 Sequencer

In order to synchronize the LED-currents, the integration time and the ADC-sampling time, a built in sampling sequencers can be used. The sequencer generates the 8-bit-timings based on a 1 μ s clock which can be pre-scaled with seq_div. The results of the analog to digital conversion are automatically stored in a pipeline buffer or in register adc_data and the ADC FIFO.

The timings can be programmed with following registers (apply for man_mode=0):

Figure 34:
Timing Registers

Register	Description
seq_div	Divider of the 1 μ s input clock for all sequencer timings
seq_count	Number of measurements in one sequence
seq_start	Writing 1 starts the sequencer, 0 stops the sequencer
seq_period	Time of one measurement cycle
seq_led_start	Start time of the LED drivers within one cycle
seq_led_stop	Stop time of the LED drivers within one cycle
seq_secled_start	Start time of the secondary LED drivers within one cycle (used for SpO2)
seq_secled_stop	Stop time of the secondary LED drivers within one cycle (used for SpO2)

Register	Description
seq_itg_start	Start time of the integrator
seq_itg_stop	Stop time of the integrator
seq_sdp1_start	Start time of the synchronous demodulator's 1 positive multiplication
seq_sdp1_stop	Stop time of the synchronous demodulator's 1 positive multiplication
seq_sdm1_start	Start time of the synchronous demodulator's 1 negative multiplication
seq_sdm1_stop	Stop time of the synchronous demodulator's 1 negative multiplication
seq_sdp2_start	Start time of the synchronous demodulator's 2 positive multiplication
seq_sdp2_stop	Stop time of the synchronous demodulator's 2 positive multiplication
seq_sdm2_start	Start time of the synchronous demodulator's 2 negative multiplication
seq_sdm2_stop	Stop time of the synchronous demodulator's 2 negative multiplication
seq_adc	Sampling position of the ADC
seq_adc2tia, seq_adc3tia	If the TIA channel is selected allow a second (and third) conversion within this cycle.
sd_subs, sd_subs_always	Synchronous demodulator subsampling ratio between sequencer frequency and ADC sampling frequency.
ulp	Ultra low power bit for the sequencer. If this bit is set and sd_subs>0, it disables the LED pulses and powers off the TIA in all sequences but the one where the TIA is sampled. This bit can be used to optimize the power consumption of the LEDs and the AS7026GG (This bit is located in ADC_CFGB Register bit 1)
irq_adc_timing_error	The sequencer setup caused a timing error on ADC conversion.

Applies only If man_mode=1

- The lowest data value of all registers except seq_count and seq_div is 1.

The diagram illustrates the internal architecture of the TIA-SDM, divided into two main functional blocks: the TIA (Transimpedance Amplifier) and the Sequencer.

TIA Block:

- Inputs:** Receives signals from GPIO0, GPIO1, and GPIO3. GPIO0 and GPIO1 are connected to a bank of switches (pd1, pd2, pd3, pd4, pdi_0, pdi_1). GPIO3 is connected to a bank of switches (VD12, VD3).
- Core Components:**
 - pd_ampres, pd_ampcap, pd_amp_en:** These components form the TIA core, which is connected to a **Prefilter**.
 - Sync. Demodulator 1+2 + BP Filter + Gain Stages:** This block processes the signal from the Prefilter, including a **Clip Det** (clip detector) and **h/l vth** (high/low threshold) inputs.
 - ADC and FiFo:** The output of the demodulator is converted by an **ADC** and stored in a **FiFo** (First In, First Out) buffer.
- Control Signals:** The TIA is controlled by **reset**, **seq_itg_start**, **seq_itg_stop**, **clk**, **seq_led_start**, **seq_led_stop**, **seq_secled_start**, **seq_secled_stop**, **seq_div**, and **1μs**.

Sequencer Block:

- Control Logic:** The Sequencer contains a **register diode_ctrl** and a **led_x_mode x=1...4** block.
- Sequencing Logic:** It includes a **seq_itg_start** and **seq_itg_stop** block, a **seq_led_start** and **seq_led_stop** block, and a **seq_secled_start** and **seq_secled_stop** block.
- ADC Logic:** The Sequencer controls the ADC via **seq_adc**, **seq_adc2tia**, and **seq_adc3tia** signals.
- Run / Stop Logic:** This block manages the **Run / Reset** state and the **seq_count cycles**.
- Counter:** A **Counter 0...seq_period-1** is used to generate the **seq_div** signal.
- Outputs:** The Sequencer outputs **seq_start** and **read status** signals.

7.1.1.9 Sequencer Registers

SEQ_CNT Register (Address 0x30)

Figure 36:
SEQ_CNT Register

Addr: 0x30		SEQ_CNT		
Bit	Bit Name	Default	Access	Bit Description
7:0	seq_count	0	RW	Number of measurements in one sequence. If seq_count = 0x0 the sequencer is running continuously if started by seq_start=1 or seq_start_sync=1. This register is reset by disabling/enabling of seq_start=0 (but not by osc_off=1)

SEQ_DIV Register (Address 0x31)

Figure 37:
SEQ_DIV Register

Addr: 0x31		SEQ_DIV		
Bit	Bit Name	Default	Access	Bit Description
7:0	seq_div	0	RW	Divider value Sequencer time increment tclk = (seq_div + 1) * 1 μ s

The SEQ_DIV register sets the input divider for the main clock.

SEQ_START (Address 0x32)

Figure 38:
SEQ_START Register

Addr: 0x32		SEQ_START		
Bit	Bit Name	Default	Access	Bit Description
7:2	Not used	0	R_PUSH	Not used

Addr: 0x32		SEQ_START		
Bit	Bit Name	Default	Access	Bit Description
1	seq_start_sync	0	R_PUSH	Similar to seq_start, but the sequencer will wait for overflow of the frequency divider that feeds all the switched-cap filters. This means 1) That it could take anything between 0 and 8 ms before the sequencer actually starts. 2) That the generated frequencies are in phase with the sequencer. For this to have any effect, the sequencer period should be selected with the selected frequencies (sd_bw, hp_freq) in mind.
0	seq_start	0	R_PUSH	Writing 1 starts the sequencer(s) in the according to the configuration and upon rising edge of seq_start ADC selects first channel. Writing 0 stops the sequencer(s). In manual mode, writing 1 starts one ADC conversion but does not initialize the ADC channel selection. Reading returns 1 if the sequencer is running (sequencer mode), respectively if the ADC is converting (manual mode)

With the SEQ_START register sets the configured sequencer can be started

SEQ_PER (Address 0x33)

Figure 39:
SEQ_PER Register

Addr: 0x33		SEQ_PER		
Bit	Bit Name	Default	Access	Bit Description
7:0	seq_period	0	RW	t_period Sequencer period $T = t_period * (seq_div+1) * 1 \mu s$

The SEQ_PER register sets one measurement cycle of the sequencer.

SEQ_LED_STA (Address 0x34)

Figure 40:
SEQ_LED_STA Register

Addr: 0x34		SEQ_LED_STA		
Bit	Bit Name	Default	Access	Bit Description
7:0	seq_led_start	0	RW	LED start time

The SEQ_LED register sets the LED drive timing. Data is stored as 16-bit value.

SEQ_LED_STO (Address 0x35)

Figure 41:
SEQ_LED_STO Register

Addr: 0x35		SEQ_LED_STO		
Bit	Bit Name	Default	Access	Bit Description
7:0	seq_led_stop	0	RW	LED stop time

The SEQ_LED register sets the LED drive timing. Data is stored as 16-bit value.

SEQ_SECLEL_STA (Address 0x36)

Figure 42:
SEQ_SECLEL_STA Register

Addr: 0x36		SEQ_SECLEL_STA		
Bit	Bit Name	Default	Access	Bit Description
7:0	seq_secled_start	0	RW	Secondary LED start time

The SEQ_LED register sets the secondary LED drive timing which is used in ledX_mode 6 only. Data is stored as 16-bit value.

SEQ_SECLEL_STO (Address 0x37)

Figure 43:
SEQ_SECLEL_STO Register

Addr: 0x37		SEQ_SECLEL_STO		
Bit	Bit Name	Default	Access	Bit Description
7:0	seq_secled_stop	0	RW	Secondary LED stop time

SEQ_ITG_STA (Address 0x38)

Figure 44:
SEQ_ITG_STA Register

Addr: 0x38		SEQ_ITG_STA		
Bit	Bit Name	Default	Access	Bit Description
7:0	seq_itg_start	0	RW	Integrator start time (start time=1 and stop time=0 means that it's - by default - always ON) Turning OFF the integrator actually means discharge the capacitor. This is only useful in capacitive integration mode, without the synchronous demodulator.

The SEQ_ITG register sets the photoamplifier integration time. Data is stored as 16-bit value.

SEQ_ITG_STO (Address 0x39)

Figure 45:
SEQ_ITG_STO Register

Addr: 0x39		SEQ_ITG_STO		
Bit	Bit Name	Default	Access	Bit Description
7:0	seq_itg_stop	0	RW	Integrator stop time

SEQ_SDP1_STA (Address 0x3a)

Figure 46:
SEQ_SDP1_STA Register

Addr: 0x3a		SEQ_SDP1_STA		
Bit	Bit Name	Default	Access	Bit Description
7:0	seq_sdp1_start	0	RW	Positive multiplication start time 1

The SEQ_SDP register sets the synchronous demodulator positive multiplication time. Data is stored as 16-bit value.

SEQ_SDP1_STO (Address 0x3b)

Figure 47:
SEQ_SDP1_STO Register

Addr: 0x3b		SEQ_SDP1_STO		
Bit	Bit Name	Default	Access	Bit Description
7:0	seq_sdp1_stop	0	RW	Positive multiplication stop time 1

SEQ_SDP2_STA (Address 0x3c)

Figure 48:
SEQ_SDP2_STA Register

Addr: 0x3c		SEQ_SDP2_STA		
Bit	Bit Name	Default	Access	Bit Description
7:0	seq_sdp2_start	0	RW	Positive multiplication start time 2

The SEQ_SDP register sets the synchronous demodulator positive multiplication time. Data is stored as 16-bit value.

SEQ_SDP2_STO (Address 0x3d)

Figure 49:
SEQ_SDP2_STO Register

Addr: 0x3d		SEQ_SDP2_STO		
Bit	Bit Name	Default	Access	Bit Description
7:0	seq_sdp2_stop	0	RW	Positive multiplication stop time 2

SEQ_SDM1_STA (Address 0x3e)

Figure 50:
SEQ_SDM1_STA Register

Addr: 0x3e		SEQ_SDM1_STA		
Bit	Bit Name	Default	Access	Bit Description
7:0	seq_sdm1_start	0	RW	Negative multiplication start time 1

The SEQ_SDM1 register sets the synchronous demodulator negative multiplication time 1. Data is stored as 16-bit value

SEQ_SDM1_STO (Address 0x3f)

Figure 51:
SEQ_SDM1_STO Register

Addr: 0x3f		SEQ_SDM1_STO		
Bit	Bit Name	Default	Access	Bit Description
7:0	seq_sdm1_stop	0	RW	Negative multiplication stop time 1

SEQ_SDM2_STA (Address 0x40)

Figure 52:
SEQ_SDM2_STA Register

Addr: 0x40		SEQ_SDM2_STA		
Bit	Bit Name	Default	Access	Bit Description
7:0	seq_sdm2_start	0	RW	Negative multiplication start time 2

The SEQ_SDM2 register sets the synchronous demodulator negative multiplication time 2. Data is stored as 16-bit value

SEQ_SDM2_STO (Address 0x41)

Figure 53:
SEQ_SDM2_STO Register

Addr: 0x41		SEQ_SDM2_STA		
Bit	Bit Name	Default	Access	Bit Description
7:0	seq_sdm2_stop	0	RW	Negative multiplication stop time 2

SEQ_ADC (Address 0x42)

Figure 54:
SEQ_ADC Register

Addr: 0x42		SEQ_ADC		
Bit	Bit Name	Default	Access	Bit Description
7:0	seq_adc	0	RW	ADC sampling time The ADC conversion needs to be finished before the sequencer period ends otherwise ADC samples can be lost.

The SEQ_ADC register defines the time when the ADC starts sampling during each measurement cycle.

SEQ_ADC2TIA (Address 0x43)**Figure 55:**
SEQ_ADC2TIA Register

Addr: 0x43		SEQ_ADC2TIA		
Bit	Bit Name	Default	Access	Bit Description
7:0	seq_adc2tia	0	RW	ADC second sampling time for TIA: If this time is non-zero, an ADC conversion is started at the given cycle, but only if adc_sel is currently selecting TIA. For all other channels, there is only a single ADC conversion executed in the sequencer period. Warning: If non-zero, seq_adc must be non-zero as well, and seq_adc2tia bigger than seq_adc. The difference must be high enough so that the second ADC conversion is started after the first ADC conversion has finished. Also, if the seq_adc2tia features is used, there is the additional restriction that the second ADC conversion has to be finished before the end of the sequencer period.

SEQ_ADC3TIA (Address 0x44)**Figure 56:**
SEQ_ADC3TIA Register

Addr: 0x44		SEQ_ADC3TIA		
Bit	Bit Name	Default	Access	Bit Description
7:0	seq_adc3tia	0	RW	ADC third sampling time for TIA: same as seq_adc2tia. Also must make sure to not overlap ADC conversions! Also, adc3tia must be after adc2tia

SEQ_ADC3TIA (Address 0x45)

Figure 57:
SD_SUBS Register

Addr: 0x45		SD_SUBS		
Bit	Bit Name	Default	Access	Bit Description
7:0	sd_subs	0	RW	Synchronous demodulator subsampling ratio between sequencer frequency and ADC sampling frequency. $\text{ADC-Fsample} = \frac{\text{Sequencer_Frequency}}{(\text{sd_subs}+1)}$ When setting to 0, then in every sequencer iteration the ADC will run. When setting to 1, then the first sequencer iteration will not trigger the ADC, but the second one will. Setting to N will make N iterations without ADC, followed by one iteration with the ADC measurement executed. It is recommended to use the ADC interrupt in this case and not the sequencer interrupt. Also see sd_subs_always which significantly affects this mechanism.

SEQ_CFG (Address 0x46)

Figure 58:
SEQ_CFG Register

Addr: 0x46		SEQ_CFG		
Bit	Bit Name	Default	Access	Bit Description
7:1	Not Used	0	RW	Not Used
0	sd_subs_always	0	RW	If this bit is asserted, all sequencer periods are subject to subsampling as defined in SD_SUBS. If this bit is zero, then only the first period of an "ADC cycle" is duplicated sd_subs times, all other periods are regular. One "ADC cycle" is the time from the sequence in which adc_sel is pointing to the "smallest" adc channel up and including the sequence of the "largest" adc channel.

SEQ_CFG (Address 0x47)**Figure 59:**
SEQ_ERR Register

Addr: 0x47		SEQ_ERR		
Bit	Bit Name	Default	Access	Bit Description
7	irq_adc_timing_error	0	SS_WC	The ADC was started by the sequencer (or manually) while it was still converting. This does not flag an interrupt but when playing with the sequencer settings we suggest to check this flag to make sure that there is no problem with the sequencer programming
6:0	Not Used	0	RW	Not Used

CYC_COUNTER (Address 0x60)**Figure 60:**
CYC_COUNTER Register

Addr: 0x60		CYC_COUNTER		
Bit	Bit Name	Default	Access	Bit Description
7:0	cycle_counter	0	RO	Current cycle counter value

The SEQ_COUNTER register shows the current value of the sequence counter and period counter

SEQ_COUNTER (Address 0x61)**Figure 61:**
SEQ_COUNTER Register

Addr: 0x61		SEQ_COUNTER		
Bit	Bit Name	Default	Access	Bit Description
7:0	sequence_counter	0	RO	Current sequence counter value

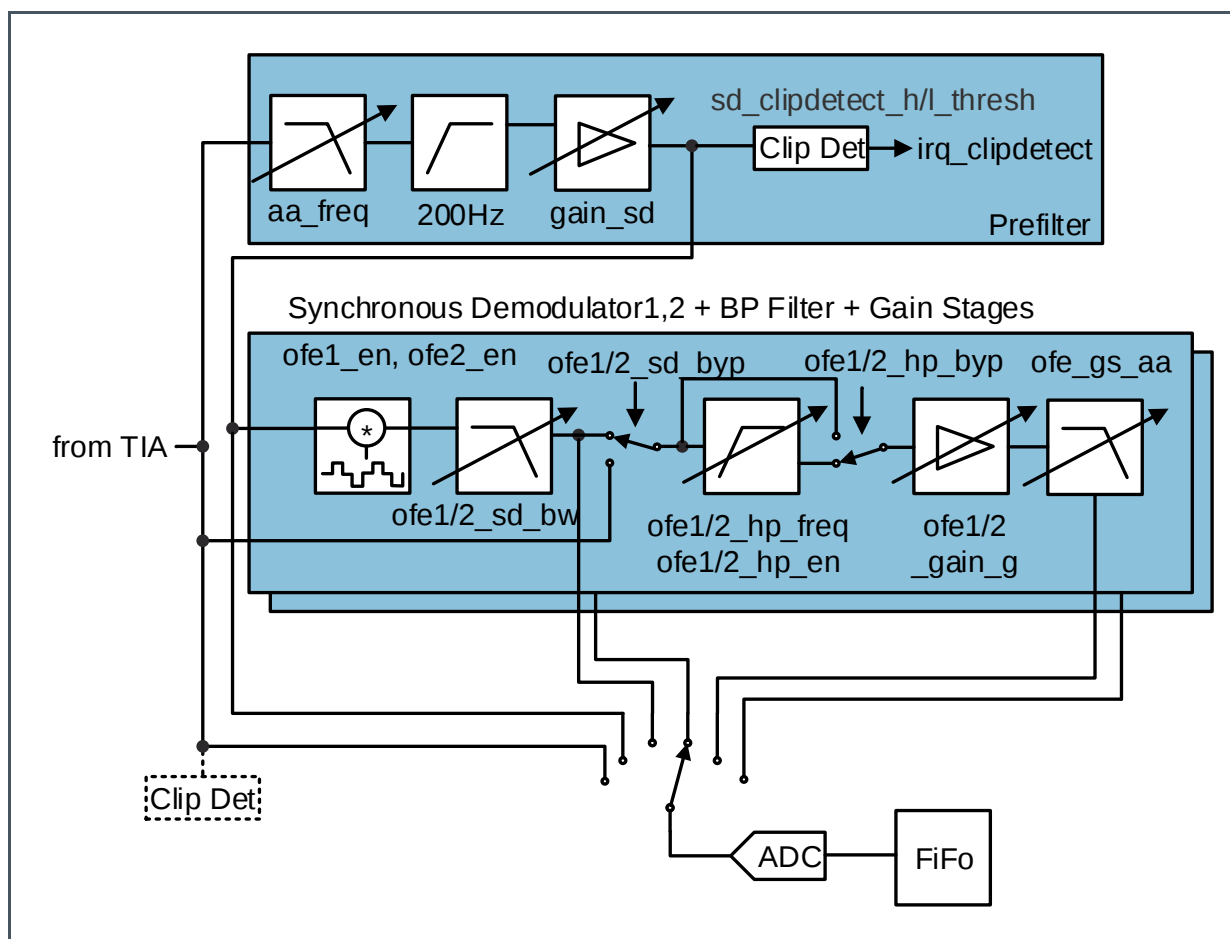
SUBS_COUNTER (Address 0x62)

Figure 62:
SUBS_COUNTER Register

Addr: 0x62		SUBS_COUNTER			
Bit	Bit Name	Default	Access	Bit Description	
7:0	subs_counter	0	RO	Current subsampling counter value	

7.1.10 Optical Signal Conditioning

Figure 63:
Optical Signal Conditioning



Synchronous Demodulator

Two optional synchronous demodulators can be used to detect small optical signals in the presence of large unwanted noise (ambient light). Since the detector synchronizes to the LED frequency, the demodulator can only be used if the measurement sequencer is running.

It includes input filter (high pass at 200 Hz, adjustable low pass) and an 2nd order adjustable output low pass. The demodulator itself multiplies the signal by +1 / 0 / -1 with a timing which is controlled by the sequencer.



Information

The optical signal conditioning stage needs `sigref_en=1` for operation.

High Pass Filter

Two optional high pass filters can be used to remove unwanted DC-components from the signal and allow further amplification. In order to guarantee fast settling times of the filter, four cutoff frequencies can be chosen.

Gain Stage

Two optional gain stages can be used to amplify the signal after the DC-component has been removed.

7.1.11 Optical Signal Conditioning Registers

OFE_CFGA (Address 0x50)

Figure 64:
OFE_CFGA Register

Addr: 0x50		OFE_CFGA					
Bit	Bit Name	Default	Access	Bit Description			
7	ofe2_en	0	RW	Enable OFE2			
6	ofe1_en	0	RW	Enable OFE1			
5	en_bias_ofe	0	RW	Enable bias for OFE and TIA			
4:3				aa_freq	0	RW	Anti-aliasing filter cut-OFF frequency
				Settings		Signal	
				0		10kHz	
				1		20kHz	
				2		40kHz	
				3		60kHz	
2:0				gain_sd	0	RW	SD gain
				Settings		Normal Gain	
				0		1	
				1		2	
				2		4	
				3		8	
				4		16	
				5		32	
				6		64	
				7		Reserved	

OFE_CFGB (Address 0x50)

Figure 65:
OFE_CFGB Register

Addr: 0x51		OFE_CFGB			
Bit	Bit Name	Default	Access	Bit Description	
7:4	sd_clipdetect_h_thresh	0	RW	If the voltage on the output of the gain_sd stage (input of synchronous demodulator) exceed this threshold the irq_clipdetect interrupt is asserted. The threshold is defined as:	
				0 ... 1824 mV	
				1 ... 1748 mV	
				2 ... 1672 mV	
				3 ... 1596 mV	
				4 ... 1520 mV	
				5 ... 1444 mV	
				6 ... 1368 mV	
				7 ... 1292 mV	
				8 ... 1216 mV	
				9 ... 1140 mV	
				10 ... 1064 mV	
				11 ... 988 mV	
				12 ... 912 mV	
				13 ... 836 mV	
				14 ... 760 mV	
				15 ... 684 mV	
3:0	sd_clipdetect_l_thresh	0	RW	If the voltage on the output of the gain_sd stage (input of synchronous demodulator) falls below this threshold the irq_clipdetect interrupt is asserted. The threshold is defined as:	
				0 ... 67 mV	
				1 ... 143 mV	
				2 ... 219 mV	
				3 ... 295 mV	
				4 ... 371 mV	
				5 ... 447 mV	
				6 ... 523 mV	
				7 ... 599 mV	
				8 ... 675 mV	
				9 ... 751 mV	
				10 ... 827 mV	
				11 ... 903 mV	

Addr: 0x51		OFE_CFGB		
Bit	Bit Name	Default	Access	Bit Description
				12 ... 979 mV
				13 ... 1055 mV
				14 ... 1131 mV
				15 ... 1207 mV

OFE_CFGC (Address 0x52)

Figure 66:
OFE_CFGC Register

Addr: 0x52		OFE_CFGC		
Bit	Bit Name	Default	Access	Bit Description
7	Not used	0	RW	Not used
6	prefilter_aa_byp	0	RW	0 ... Anti aliasing filter (aa_filter) is used 1 ... Bypass anti aliasing filter
5	prefilter_hp_byp	0	RW	0 ... Use 200 Hz high pass filter 1 ... Bypass 200 Hz high pass filter
4	prefilter_gain_byp	0	RW	0 ... Use gain_sd stage 1 ... Bypass gain_sd stage
3	prefilter_bypass_en	0	RW	0 ... Use prefilter unless any of the above register is set 1 ... Bypass complete prefilter
2	prefilter_aa_en	0	RW	0 ... Anti aliasing filter (aa_filter) is OFF 1 ... Anti aliasing filter is ON
1	prefilter_hp_en	0	RW	0 ... 200 Hz high pass filter is OFF 1 ... 200 Hz high pass filter is ON
0	prefilter_gain_en	0	RW	0 ... gain_sd stage is OFF 1 ... gain_sd stage is ON

OFE_CFGD (Address 0x53)

Figure 67:
OFE_CFGD Register

Addr: 0x53		OFE_CFGD		
Bit	Bit Name	Default	Access	Bit Description
7:2	Not used	0	RW	Not used
1:0	ofe_gs_aa	0	RW	OFE anti aliasing
				Setting
				Nominal Gain
				0 Bypass
				1 fc=100 kHz
				2 fc=10 kHz
				3 fc=826 Hz

OFE1_CFGA (Address 0x54)

Figure 68:
OFE1_CFGA Register

Addr: 0x54		OFE1_CFGA		
Bit	Bit Name	Default	Access	Bit Description
7	ofe1_sd_pol_init	0	RW	The low level driver shall ensure that this register is 0 if one of the seq_sdm pulses is first, and is 1 if the seq_sdp is first within a sequence.
6	ofe1_sd_en	0	RW	0 ... Power down of the Synchronous demodulator 1 ... Enable Synchronous demodulator
5	ofe1_hp_en	0	RW	0 ... Power down of the high pass filter 1 ... Enable high pass filter
4	ofe1_gain_en	0	RW	0 ... Power down of the Gain stage 1 ... Enable Gain stage
3	ofe1_sd_byp	0	RW	0 ... Synchronous demodulator is used 1 ... Synchronous demodulator is bypassed
2	ofe1_hp_byp	0	RW	0 ... HP filter is used 1 ... HP filter is bypassed
1	ofe1_gain_byp	0	RW	0 ... Gain stage is used 1 ... Gain stage is bypassed

Addr: 0x54		OFE1_CFGA		
Bit	Bit Name	Default	Access	Bit Description
0	ofe1_sd_hld	0	RW	SD hold 0 ... Output of synchronous demodulator is forced to SIGREF if not set to +1 or -1 1... Output of synchronous demodulator is tristated if not set to +1 or -1

OFE1_CFGB (Address 0x55)

Figure 69:
OFE1_CFGB Register

Addr: 0x55		OFE1_CFGB			
Bit	Bit Name	Default	Access	Bit Description	
7	Not used	0	RW	Not used	
6:4	ofe1_gain_g	0	RW	Gain	
				Setting	Gain
				0	1
				1	2
				2	4
				3	8
				4	16
				5	32
				6	64
				7	Do not use
3:2	ofe1_sd_bw	0	RW	Synchronous demodulator low pass filter	
				Setting	Frequency
				0	10 Hz
				1	20 Hz
				2	40 Hz
				3	80 Hz

Addr: 0x55		OFE1_CFGB												
Bit	Bit Name	Default	Access	Bit Description										
				High pass filter cutoff frequency										
				<table><tr><th>Setting</th><th>Cutoff Frequency</th></tr><tr><td>0</td><td>0.33 Hz</td></tr><tr><td>1</td><td>1.32 Hz</td></tr><tr><td>2</td><td>5.28 Hz</td></tr><tr><td>3</td><td>10.56 Hz</td></tr></table>	Setting	Cutoff Frequency	0	0.33 Hz	1	1.32 Hz	2	5.28 Hz	3	10.56 Hz
Setting	Cutoff Frequency													
0	0.33 Hz													
1	1.32 Hz													
2	5.28 Hz													
3	10.56 Hz													
1:0	ofe1_hp_freq	0	RW											

OFE2_CFGA (Address 0x58)

Figure 70:
OFE2_CFGA Register

Addr: 0x58		OFE2_CFGA		
Bit	Bit Name	Default	Access	Bit Description
7	ofe2_sd_pol_init	0	RW	The low level driver shall ensure that this register is 0 if one of the seq_sdm pulses is first, and is 1 if the seq_sdp is first within a sequence.
6	ofe2_sd_en	0	RW	0 ... Power down of the synchronous demodulator 1 ... Enable Synchronous demodulator
5	ofe2_hp_en	0	RW	0 ... Power down of the high pass filter 1 ... Enable high pass filter
4	ofe2_gain_en	0	RW	0 ... Power down of the Gain stage 1 ... Enable Gain stage
3	ofe2_sd_byp	0	RW	0 ... Synchronous demodulator is used 1 ... Synchronous demodulator is bypassed
2	ofe2_hp_byp	0	RW	0 ... HP filter is used 1 ... HP filter is bypassed
1	ofe2_gain_byp	0	RW	0 ... Gain stage is used 1 ... Gain stage is bypassed
0	ofe2_sd_hld	0	RW	SD hold 0 ... Output of synchronous demodulator is forced to SIGREF if not set to +1 or -1 1... Output of synchronous demodulator is tristated if not set to +1 or -1

OFE2_CFGB (Address 0x59)

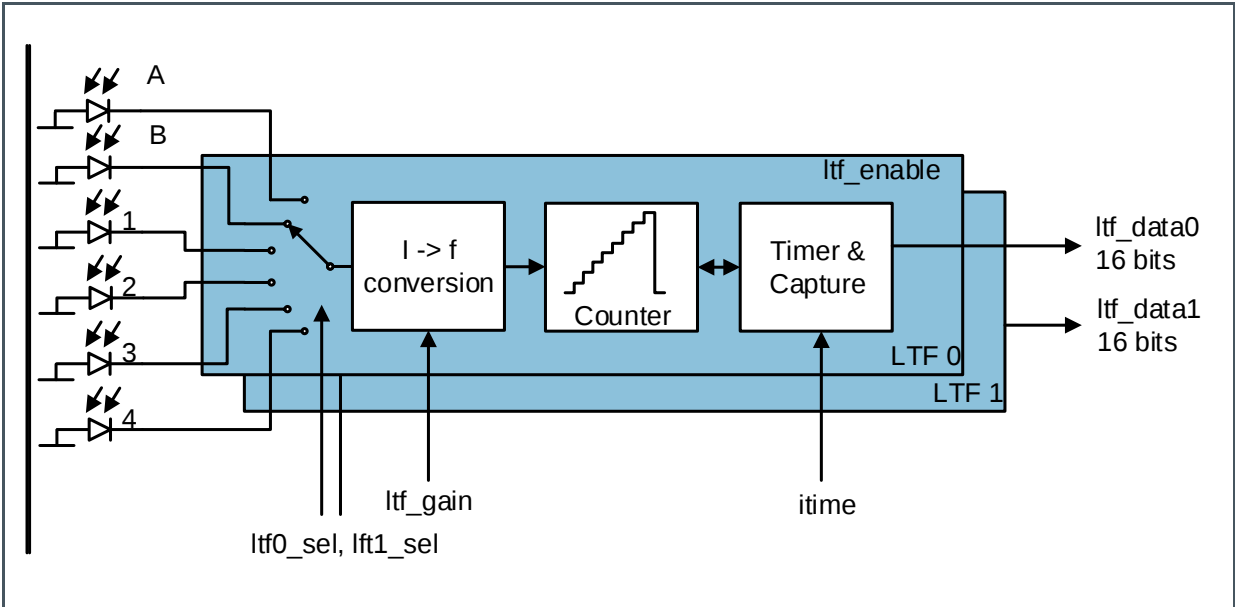
Figure 71:
OFE2_CFGB Register

Addr: 0x59		OFE2_CFGB																				
Bit	Bit Name	Default	Access	Bit Description																		
7	Not used	0	RW	Not used																		
				Gain																		
				<table><tr><th>Setting</th><th>Gain</th></tr><tr><td>0</td><td>1</td></tr><tr><td>1</td><td>2</td></tr><tr><td>2</td><td>4</td></tr><tr><td>3</td><td>8</td></tr><tr><td>4</td><td>16</td></tr><tr><td>5</td><td>32</td></tr><tr><td>6</td><td>64</td></tr><tr><td>7</td><td>Do not use</td></tr></table>	Setting	Gain	0	1	1	2	2	4	3	8	4	16	5	32	6	64	7	Do not use
Setting	Gain																					
0	1																					
1	2																					
2	4																					
3	8																					
4	16																					
5	32																					
6	64																					
7	Do not use																					
6:4	ofe2_gain_g	0	RW																			
				Synchronous demodulator low pass filter																		
				<table><tr><th>Setting</th><th>Frequency</th></tr><tr><td>0</td><td>10 Hz</td></tr><tr><td>1</td><td>20 Hz</td></tr><tr><td>2</td><td>40 Hz</td></tr><tr><td>3</td><td>80 Hz</td></tr></table>	Setting	Frequency	0	10 Hz	1	20 Hz	2	40 Hz	3	80 Hz								
Setting	Frequency																					
0	10 Hz																					
1	20 Hz																					
2	40 Hz																					
3	80 Hz																					
3:2	ofe2_sd_bw	0	RW																			
				High pass filter cutoff frequency																		
				<table><tr><th>Setting</th><th>Cutoff Frequency</th></tr><tr><td>0</td><td>0.33 Hz</td></tr><tr><td>1</td><td>1.32 Hz</td></tr><tr><td>2</td><td>5.28 Hz</td></tr><tr><td>3</td><td>10.56 Hz</td></tr></table>	Setting	Cutoff Frequency	0	0.33 Hz	1	1.32 Hz	2	5.28 Hz	3	10.56 Hz								
Setting	Cutoff Frequency																					
0	0.33 Hz																					
1	1.32 Hz																					
2	5.28 Hz																					
3	10.56 Hz																					
1:0	ofe2_hp_freq	0	RW																			

7.1.12 Light-to-Frequency Mode

The LTF (light-to-frequency, or FM, frequency mode) mode.

Figure 72:
Light-to-Frequency Mode Internal Circuit



- (1) Do not use diodes which are connected to the TIA (register pd_a, pd_b, pd1...4) at the same time when ltf_en is enabled on the same diode.

LTFDATA0_L (Address 0x20)

Figure 73:
LTFDATA0_L Register

Addr: 0x20		LTFDATA0_L		
Bit	Bit Name	Default	Access	Bit Description
7:0	ltfdata0[7:0]	0	RO	LTF result channel 0 low byte. Software must make sure that the LTF integration is not running when accessing the LTFDATA registers. These are the direct counter registers, they are not latched. If buffering is required, consider using FIFO mode.

LTFDATA0_H (Address 0x21)

Figure 74:
LTFDATA0_H Register

Addr: 0x21		LTFDATA0_H		
Bit	Bit Name	Default	Access	Bit Description
7:0	ltfdata0[15:8]	0	RO	LTF result channel 0 high byte

LTFDATA1_L (Address 0x22)

Figure 75:
LTFDATA1_L Register

Addr: 0x22		LTFDATA1_L		
Bit	Bit Name	Default	Access	Bit Description
7:0	ltfdata0[7:0]	0	RO	LTF result channel 1 low byte. Software must make sure that the LTF integration is not running when accessing the LTFDATA registers. If buffering is required, consider using FIFO mode.

LTFDATA1_H (Address 0x23)

Figure 76:
LTFDATA1_H Register

Addr: 0x23		LTFDATA1_H		
Bit	Bit Name	Default	Access	Bit Description
7:0	ltfdata1[15:8]	0	RO	LTF result channel 1 high byte

ITIME (Address 0x24)

Figure 77:
ITIME Register

Addr: 0x24		ITIME		
Bit	Bit Name	Default	Access	Bit Description
7:0	itime	0	RW	LTF integration time. MODCLK is 2/3MHz (666.67 kHz). One LSB of itime is 3.072 ms (2048 MODCLK cycles). 0=3.072 ms ... 255=786.432 ms Using the itime_unit register (see below), the unit of itime can be reduced by 2, 4, or 8. This shorter integration times can be selected (required for flicker detection), but it can also be used to increase the resolution of itime. For example, if 50ms integration time are desired, the best value for regular itime would be 15 (=16 periods=49.152 ms). However, but setting itime_unit=2 (LSB=768 μs), one can select 64 (=65 periods=49.9 ms) Warning: selecting an integration time smaller than 3.072 ms will reduce the resolution of the conversion, as the maximum Itfdata value is not 1024 (10 bits) anymore, but 512 (9 bits) in case of 1.536 ms integration time, 256 (8 bits) for 768 μs and 128 (7 bits) for 384 μs

LTF_CONFIG (Address 0x25)

Figure 78:
LTF_CONFIG Register

Addr: 0x25		LTF_CONFIG		
Bit	Bit Name	Default	Access	Bit Description
7	infinite_itime	0	RW	If this is asserted, then integration does not stop. The ITIME setting is ignored. Use with watch the ltfdata counters. (Warning: must be filtered in software to prevent inconsistent upper/lower byte). It's implemented as a count disable on the integration counter, so when resetting bit to 0 again, the itime counter will continue and results can be read afterwards through the regular mechanisms (ltfdata or FIFO) This is intended for very long integration times - as the timing is controlled by software/I ² C, accuracy fully depends on the system and I ² C master.
6	az_disable_auto	0	RW	0: Run autozero on both channels every time FM mode is activated for the first time after ENAB is being asserted. 1: Do not run autozero automatically. Autozero can only be activated manually (AZ_CONTROL)
5:4	reserved	0	RW	Reserved – leave at 0
1	ltf_fifo_mode	0	RW	Run LTF integrations back to back, the LTF modulator is running continuously (the modulators are not reset between integrations cycles). After each integration, the result gets written to the FIFO. The FIFO is being filled automatically, FIFO threshold interrupt is flagged as configured. The first item read from the FIFO is from channel 0, the next one from channel 1, etc. Note that there is no ltf_done interrupt triggered after each integration. A FIFO threshold of 1 can be used to generate an interrupt for each result. irq_ltf_enab should be kept asserted to avoid missing an ltf_sat interrupt. Do not enable ADC/sequencer FIFO mode and ltf_fifo_mode at the same time, corrupted data would be the result. Make sure to empty the FIFO in time, if the FIFO is full, new data is not being stored in the FIFO. Source of data read from the FIFO after an overflow condition is undefined (can be from channel 0 or channel 1) Stop the procedure by clearing this bit.

Addr: 0x25		LTF_CONFIG		
Bit	Bit Name	Default	Access	Bit Description
0	ltf_enable	0	RW	This bit must be asserted for any LTF function (powers up the LTF clock tree)

LTF_SEL (Address 0x26)

Figure 79:
LTF_SEL Register

Addr: 0x26		LTF_SEL																				
Bit	Bit Name	Default	Access	Bit Description																		
7	Do not use	0	RW	Do not use																		
				Select the sensor diode for LTF1																		
				<table><tr><th>Setting</th><th>Source</th></tr><tr><td>0</td><td>A</td></tr><tr><td>1</td><td>A/16</td></tr><tr><td>2</td><td>B</td></tr><tr><td>3</td><td>B/16</td></tr><tr><td>4</td><td>PD1</td></tr><tr><td>5</td><td>PD2</td></tr><tr><td>6</td><td>PD3</td></tr><tr><td>7</td><td>PD4</td></tr></table>	Setting	Source	0	A	1	A/16	2	B	3	B/16	4	PD1	5	PD2	6	PD3	7	PD4
Setting	Source																					
0	A																					
1	A/16																					
2	B																					
3	B/16																					
4	PD1																					
5	PD2																					
6	PD3																					
7	PD4																					
6:4	ltf1_sel	2	RW																			
				Select the sensor diode for LTF0																		
				<table><tr><th>Setting</th><th>Source</th></tr><tr><td>0</td><td>A</td></tr><tr><td>1</td><td>A/16</td></tr><tr><td>2</td><td>B</td></tr><tr><td>3</td><td>B/16</td></tr><tr><td>4</td><td>PD1</td></tr><tr><td>5</td><td>PD2</td></tr><tr><td>6</td><td>PD3</td></tr><tr><td>7</td><td>PD4</td></tr></table>	Setting	Source	0	A	1	A/16	2	B	3	B/16	4	PD1	5	PD2	6	PD3	7	PD4
Setting	Source																					
0	A																					
1	A/16																					
2	B																					
3	B/16																					
4	PD1																					
5	PD2																					
6	PD3																					
7	PD4																					
2:0	ltf0_sel	0	RW																			

LTF_GAIN (Address 0x27)

Figure 80:
LTF_GAIN Register

Addr: 0x27		LTF_GAIN					
Bit	Bit Name	Default	Access	Bit Description			
7:6	Do not use	0	RW	Do not use			
5:4				itime_unit	0	RW	Select the itime unit. See ITIME register description (Figure 77).
				Setting		Behavior	
				0		Normal, time LSB=3.072 ms	
				1		/2, LSB=1.536 ms	
				2		/4, time LSB=768 μs	
				3		/8, time LSB=384 μs	
3:0				ltf_gain	0	RW	Select the gain
				Setting		Gain	
				0		0.25	
				1		0.5	
				2		1	
				3		2	
				4		4	
				5		8	
				6		16	
				7		24	
				8		64	
				9-15		Reserved – do not use	

LTF_CONTROL (Address 0x28)

Figure 81:
LTF_CONTROL Register

Addr: 0x28		LTF_CONTROL		
Bit	Bit Name	Default	Access	Bit Description
7:1	Do not use	0	R_PUSH	Do not use

Addr: 0x28		LTF_CONTROL		
Bit	Bit Name	Default	Access	Bit Description
0	ltf_start	0	R_PUSH	Writing 1 starts the counter, and it will run for the specified time (itime). Afterwards it stops automatically and interrupt is flagged. writing 0 to the counter stops it as well. reading the value returns whether the counter is running. If ltf_fifo_mode is non-zero, then FM conversions are done continuously until a 0 is written to this bit again.

AZ_CONTROL (Address 0x29)

Figure 82:
AZ_CONTROL Register

Addr: 0x29		AZ_CONTROL		
Bit	Bit Name	Default	Access	Bit Description
7:2	Do not use	0	RW_SM	Do not use
1	az_enable_1	0	RW_SM	Writing a '1' to this register starts the AZ engine for channel 1. This is usually not necessary, as AZ is executed automatically before the first LTF integration (unless az_disable_auto is set) The bit is cleared to '0' automatically when the AZ has finished. You cannot write a '0' to this register.
0	az_enable_0	0	RW_SM	The same as az_enable_1, but for channel 0.

OFFSET0 (Address 0x2a)

Figure 83:
OFFSET0 Register

Addr: 0x2a		OFFSET0		
Bit	Bit Name	Default	Access	Bit Description
7:0	offset0[7:0]	0	RW_SM	This register holds the value of the offset on the channel 0 OpAmp. It can be overwritten, and it gets overwritten by the auto-zero mechanism. The value is in sign/magnitude encoding. The value is ± 127, sign/magnitude

OFFSET1 (Address 0x2b)

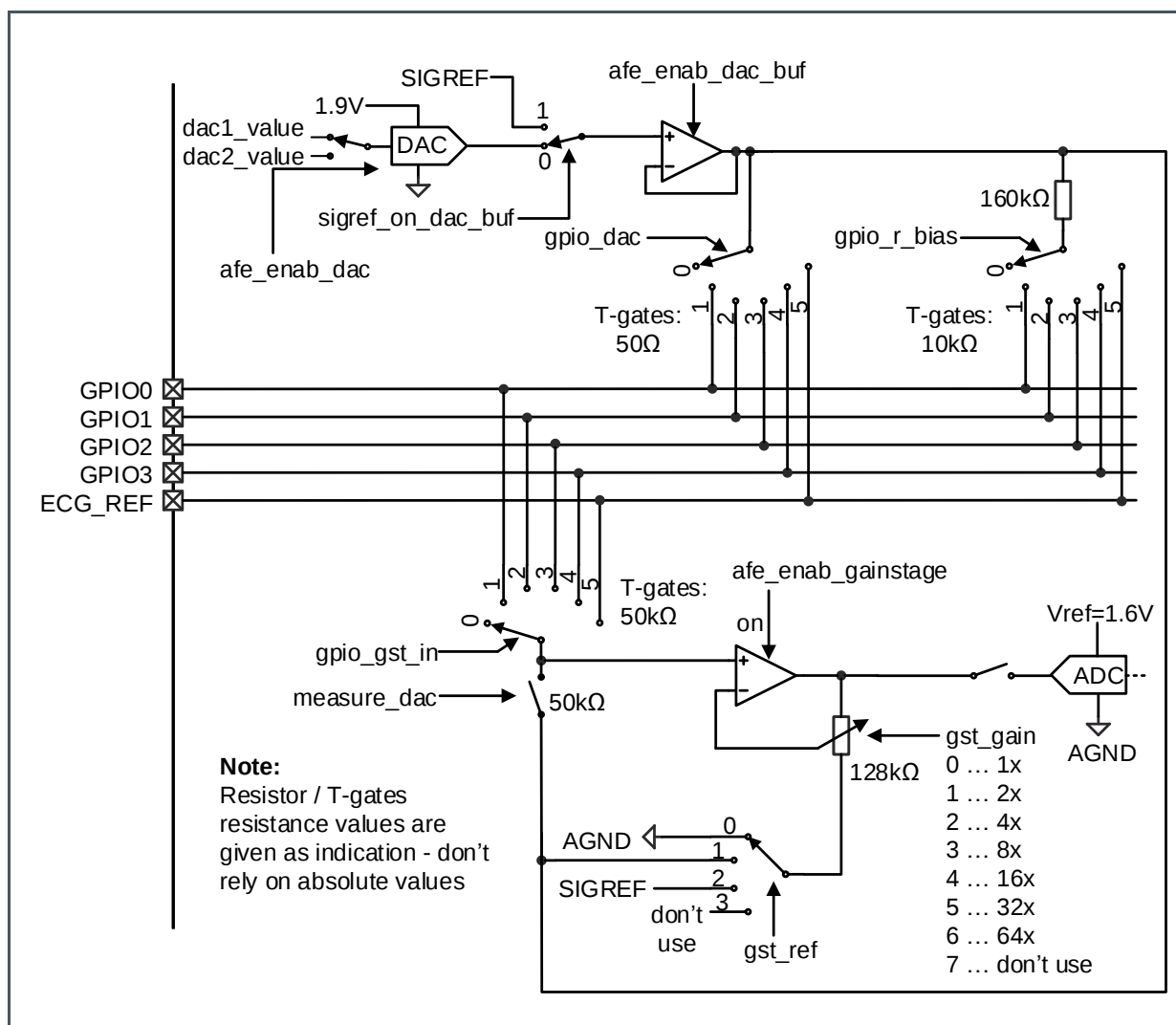
Figure 84:
OFFSET1 Register

Addr: 0x2b		OFFSET1		
Bit	Bit Name	Default	Access	Bit Description
7:0	offset0[7:0]	0	RW_SM	This register holds the value of the offset on the channel 1 OpAmp. It can be overwritten, and it gets overwritten by the auto-zero mechanism. The value is in sign/magnitude encoding. The value is ± 127 , sign/magnitude

7.1.13 Electrical Analog Front End

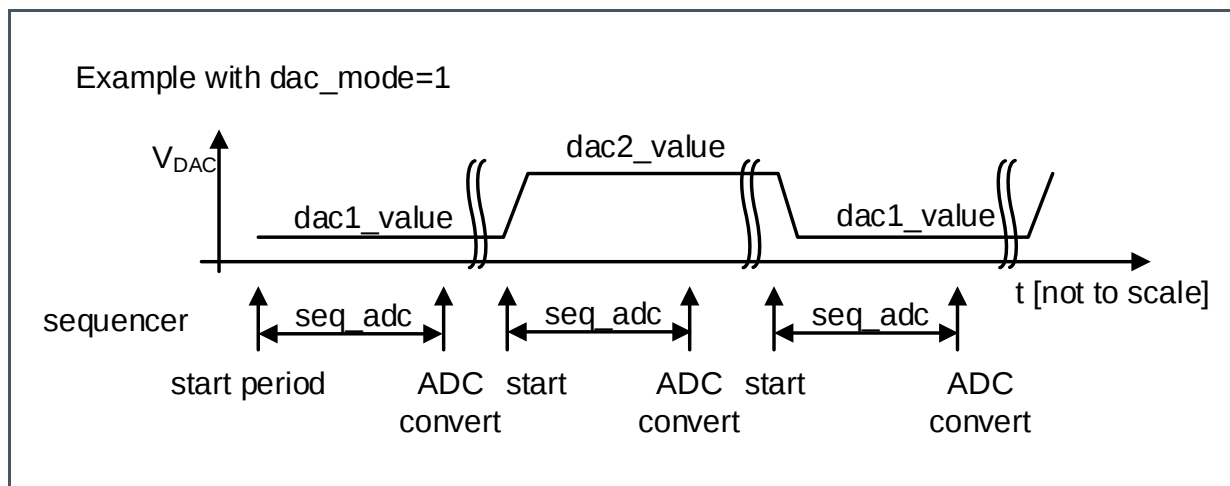
The electrical analog front end consists of three identical signal paths with independent settings of bias condition, gain and offset.

Figure 85:
Electrical Analog Front End Internal Circuit



DAC Switching

Figure 86:
Electrical Analog Front End DAC Level Switching



If bit `dac_mode` is not zero, the DAC switches its codes between `dac1_value` and `dac2_value` on the beginning of every/every 2nd/every 4th sequencer cycle where the ADC is converting the electrical frontend channel. ADC conversions of any other channel do not switch the DAC.

Input Pins

Four general purpose pins and `ECG_REF` can be used either as configurable GPIO pin or as analog input pins for the electrical analog front end. The analog inputs can be configured to setup different amplifier topologies.

7.1.14 EAF (Electrical Analog Frontend) Registers

AFE_CFG (Address 0x70)

Figure 87:
AFE_CFG Register

Addr: 0x70		AFE_CFG		
Bit	Bit Name	Default	Access	Bit Description
7:4	Do not use	0	RW	Do not use
3	afe_enab	0	RW	0 ... EAF bias deactivated 1 ... EAF bias activated (need to be set for any functions of the EAF are used).
2	afe_enab_dac	0	RW	0 ... DAC inside the EAF OFF 1 ... DAC inside the EAF ON
1	afe_enab_dac_buf	0	RW	0 ... DAC buffer OFF 1 ... DAC buffer ON
0	afe_enab_gainstage	0	RW	0 ... Gain stage in EAF OFF 1 ... Gain stage in EAF ON

The AFE_CFG register is used to configure the analog frontend.

EAF_GST (Address 0x80)

Figure 88:
EAF_GST Register

Addr: 0x80		EAF_GST			
Bit	Bit Name	Default	Access	Bit Description	
7:5	gpio_gst_in	0	RW	Gain stage input selection	
				Setting	Meaning
				0	Not connected
				1	GPIO0
				2	GPIO1
				3	GPIO2
				4	GPIO3
				5	ECG_REF

Addr: 0x80		EAF_GST			
Bit	Bit Name	Default	Access	Bit Description	
4:3				Gain stage reference voltage	
				Setting	Meaning
				0	AGND
				1	DAC buffer
				2	SIGREF
				3	Reserved
2:0				Gain stage gain	
				Setting	Meaning
				0	1
				1	2
				2	4
				3	8
				4	16
				5	32
				6	64
				7	Reserved

The EAF register is used to configure the electrical frontend

EAF_BIAS (Address 0x81)

Figure 89:
EAF_BIAS Register

Addr: 0x81		EAF_BIAS			
Bit	Bit Name	Default	Access	Bit Description	
7:5				Resistive biasing	
				Setting	Meaning
				0	No resistive biasing
				1	Resistive biasing on GPIO0
				2	Resistive biasing on GPIO1
				3	Resistive biasing on GPIO2
				4	Resistive biasing on GPIO3
				5	Resistive biasing on ECG_REF
gpio_r_bias	0	RW			

Addr: 0x81		EAF_BIAS		
Bit	Bit Name	Default	Access	Bit Description
4:0	Not used	0	RW	Do not use

EAF_DAC (Address 0x82)

Figure 90:
EAF_DAC Register

Addr: 0x82		AFE_CFG																
Bit	Bit Name	Default	Access	Bit Description														
7:5	Do not use	0	RW	Do not use														
4	sigref_on_dac_buf	0	RW	If asserted, connect SIGREF to DAC buffer.														
3	measure_dac	0	RW	If this bit is asserted, the DAC output is connected to the gain stage input (independent of gpio_gst_in selection, therefore the DAC output is measurable on the GPIO pin)														
				DAC on GPIO														
				<table><tr><th>Setting</th><th>Meaning</th></tr><tr><td>0</td><td>No DAC biasing</td></tr><tr><td>1</td><td>DAC on GPIO0</td></tr><tr><td>2</td><td>DAC on GPIO1</td></tr><tr><td>3</td><td>DAC on GPIO2</td></tr><tr><td>4</td><td>DAC on GPIO3</td></tr><tr><td>5</td><td>DAC on ECG_REF</td></tr></table>	Setting	Meaning	0	No DAC biasing	1	DAC on GPIO0	2	DAC on GPIO1	3	DAC on GPIO2	4	DAC on GPIO3	5	DAC on ECG_REF
Setting	Meaning																	
0	No DAC biasing																	
1	DAC on GPIO0																	
2	DAC on GPIO1																	
3	DAC on GPIO2																	
4	DAC on GPIO3																	
5	DAC on ECG_REF																	
2:0	gpio_dac	0	RW															

EAF_DAC1_L (Address 0x83)

Figure 91:
EAF_DAC1_L Register

Addr: 0x83		EAF_DAC1_L		
Bit	Bit Name	Default	Access	Bit Description
7:6	dac1_value[1:0]	0	RW	DAC value 1 (2LSB)
5:0	Not used	0	RW	Not used

The EAF_DAC1/2_L/H registers is used to configure the dac value. See bit dac_mode for selection of dac register 1 or 2

EAF_DAC1_H (Address 0x84)

Figure 92:
EAF_DAC1_H Register

Addr: 0x84		EAF_DAC1_H		
Bit	Bit Name	Default	Access	Bit Description
7:0	dac1_value[9:2]	0	RW	DAC value 1 (upper 8 bits) 10-bit value: 0x000 ... 0 V 0x3FF ... 1.9 V

EAF_DAC2_L (Address 0x85)

Figure 93:
EAF_DAC2_L Register

Addr: 0x85		EAF_DAC2_L		
Bit	Bit Name	Default	Access	Bit Description
7:6	dac2_value[1:0]	0	RW	DAC value 2 (2LSB)
5:0	Not used	0	RW	Not used

EAF_DAC2_H (Address 0x86)

Figure 94:
EAF_DAC2_H Register

Addr: 0x86		EAF_DAC2_H		
Bit	Bit Name	Default	Access	Bit Description
7:0	dac2_value[9:2]	0	RW	DAC value 1 (upper 8 bits) 10-bit value: 0x000 ... 0 V 0x3FF ... 1.9 V

EAF_DAC_CFG (Address 0x87)

Figure 95:
EAF_DAC_CFG Register

Addr: 0x85		EAF_DAC_CFG												
Bit	Bit Name	Default	Access	Bit Description										
7:2	Not used	0	RW	Not used										
1:0	dac_mode	0	RW	DAC mode The EAF has a DAC that can be switched out on GPIOs.										
				dac_mode 0 uses statically dac1_value, the other modes switch dynamically between the two values.										
				The system switches from one value to the next always at the beginning of a sequence in which the ADC will sample the AFE channel.										
				<table><tr><th>Setting</th><th>Meaning</th></tr><tr><td>0</td><td>1-1-1-1-1-1-1-1-1-1-</td></tr><tr><td>1</td><td>1-2-1-2-1-2-1-2-1-2-</td></tr><tr><td>2</td><td>1-1-2-2-1-1-2-2-1-1-2-</td></tr><tr><td>3</td><td>1-1-1-1-2-2-2-2-1-1-1-</td></tr></table>	Setting	Meaning	0	1-1-1-1-1-1-1-1-1-1-	1	1-2-1-2-1-2-1-2-1-2-	2	1-1-2-2-1-1-2-2-1-1-2-	3	1-1-1-1-2-2-2-2-1-1-1-
				Setting	Meaning									
				0	1-1-1-1-1-1-1-1-1-1-									
				1	1-2-1-2-1-2-1-2-1-2-									
2	1-1-2-2-1-1-2-2-1-1-2-													
3	1-1-1-1-2-2-2-2-1-1-1-													

Possible Configurations of Every Amplifier Stage

Figure 96:
Non Inverting Amplifier with Offset and Input Voltage Divider (Temperature Sensor)

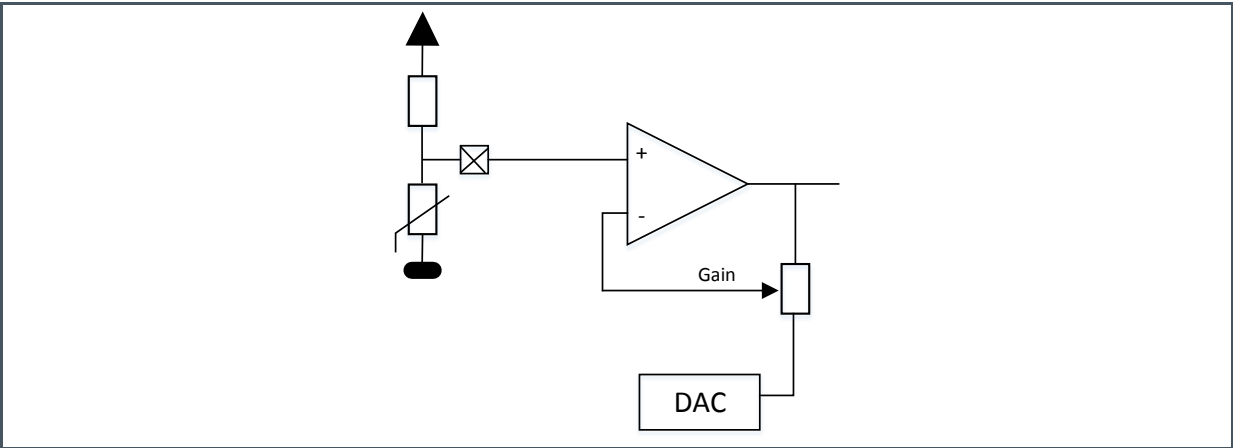


Figure 97:
Non Inverting Amplifier with Current Source and Offset (Temperature Sensor)

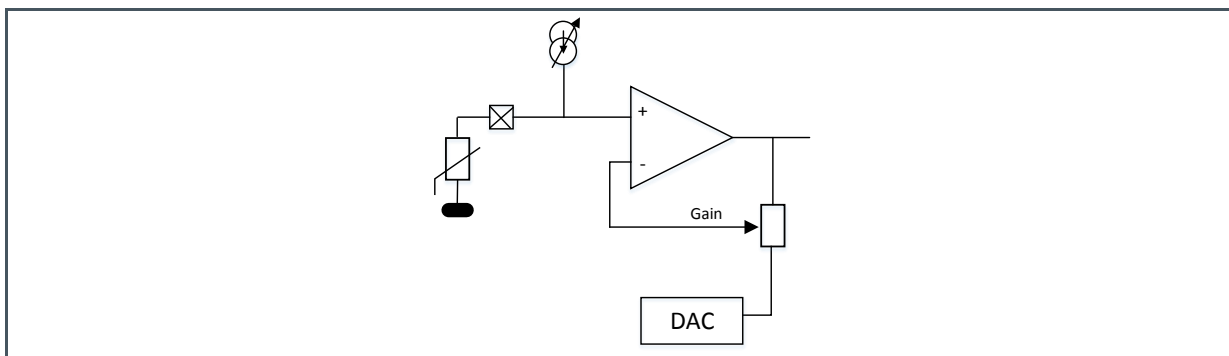


Figure 98:
Non Inverting Amplifier with Current Source and Reference Path (Temperature Sensor)

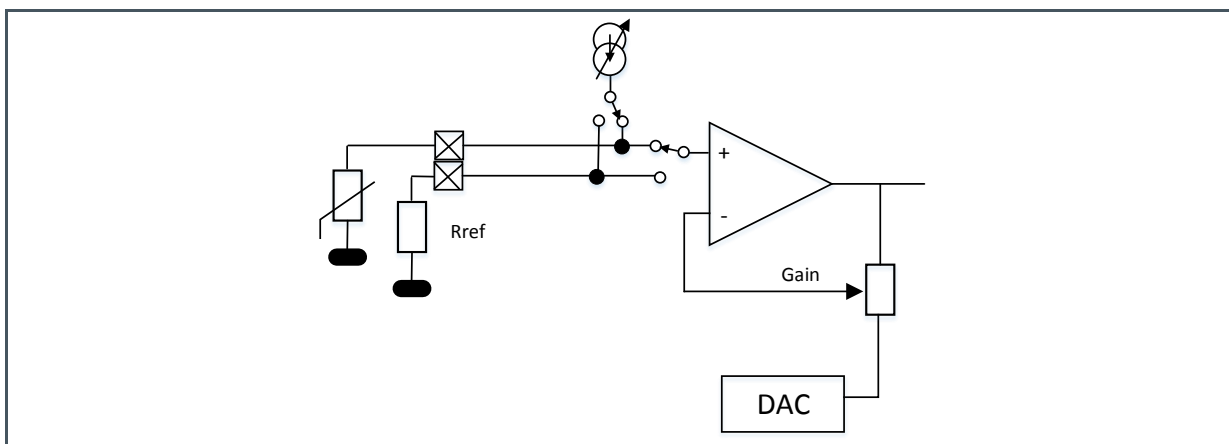


Figure 99:
Non Inverting Amplifier High Impedance, GND Referenced

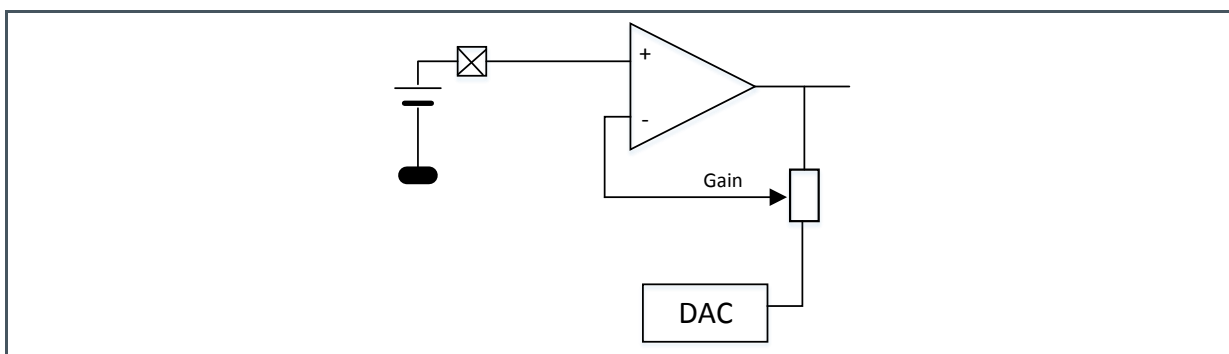


Figure 100:
Non Inverting Amplifier with DC-Blocking, Referenced to $V_{\text{ADCRef}}/2$

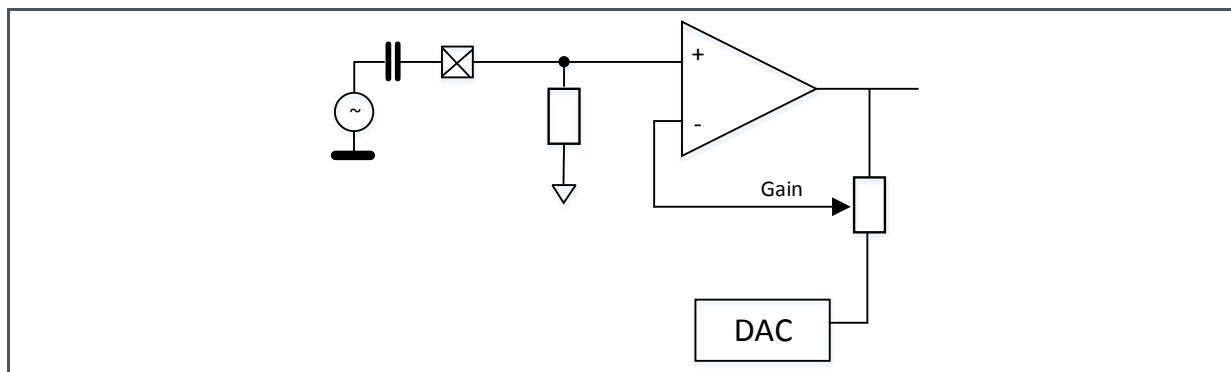
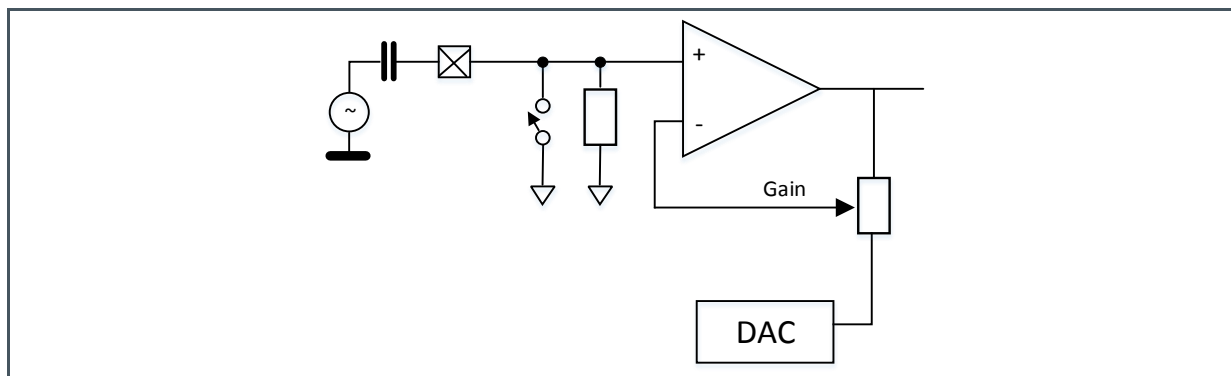
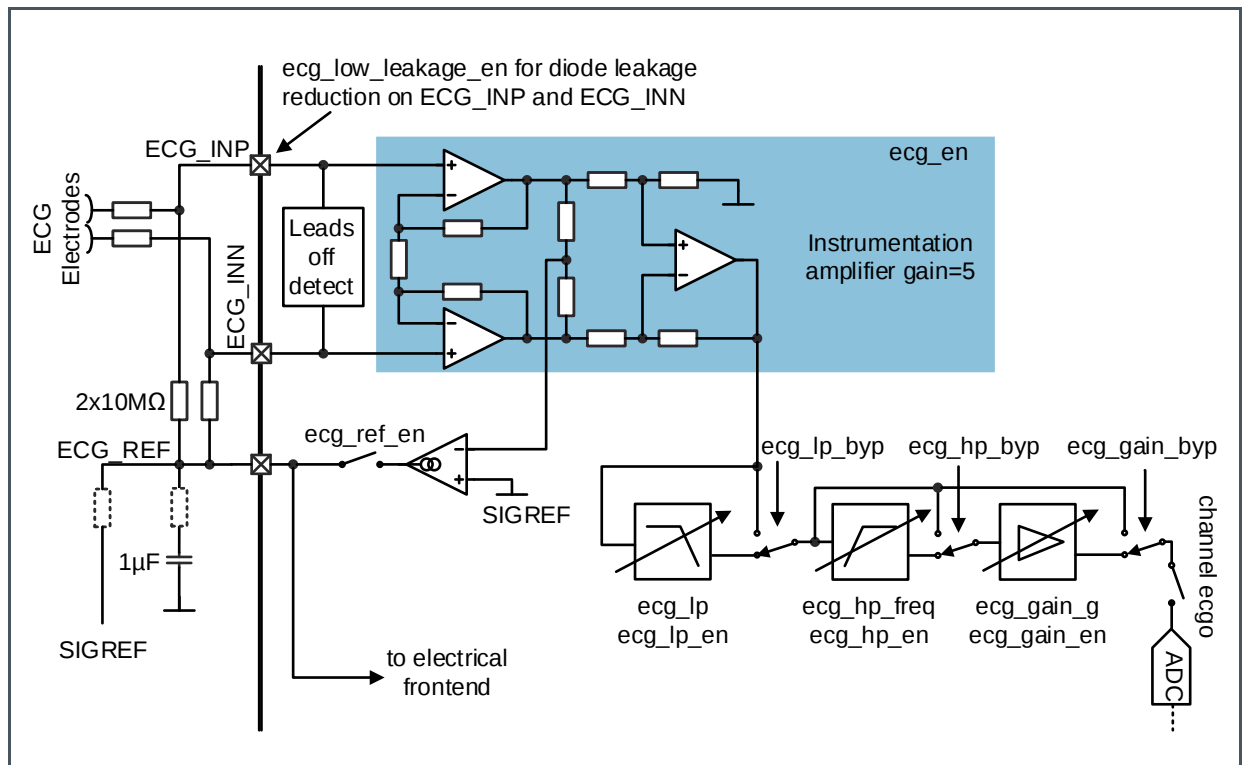


Figure 101:
Non Inverting Amplifier with DC-Blocking and Fast Settling Time, Referenced to $\text{ADCRef} / 2$



7.1.15 ECG Amplifier

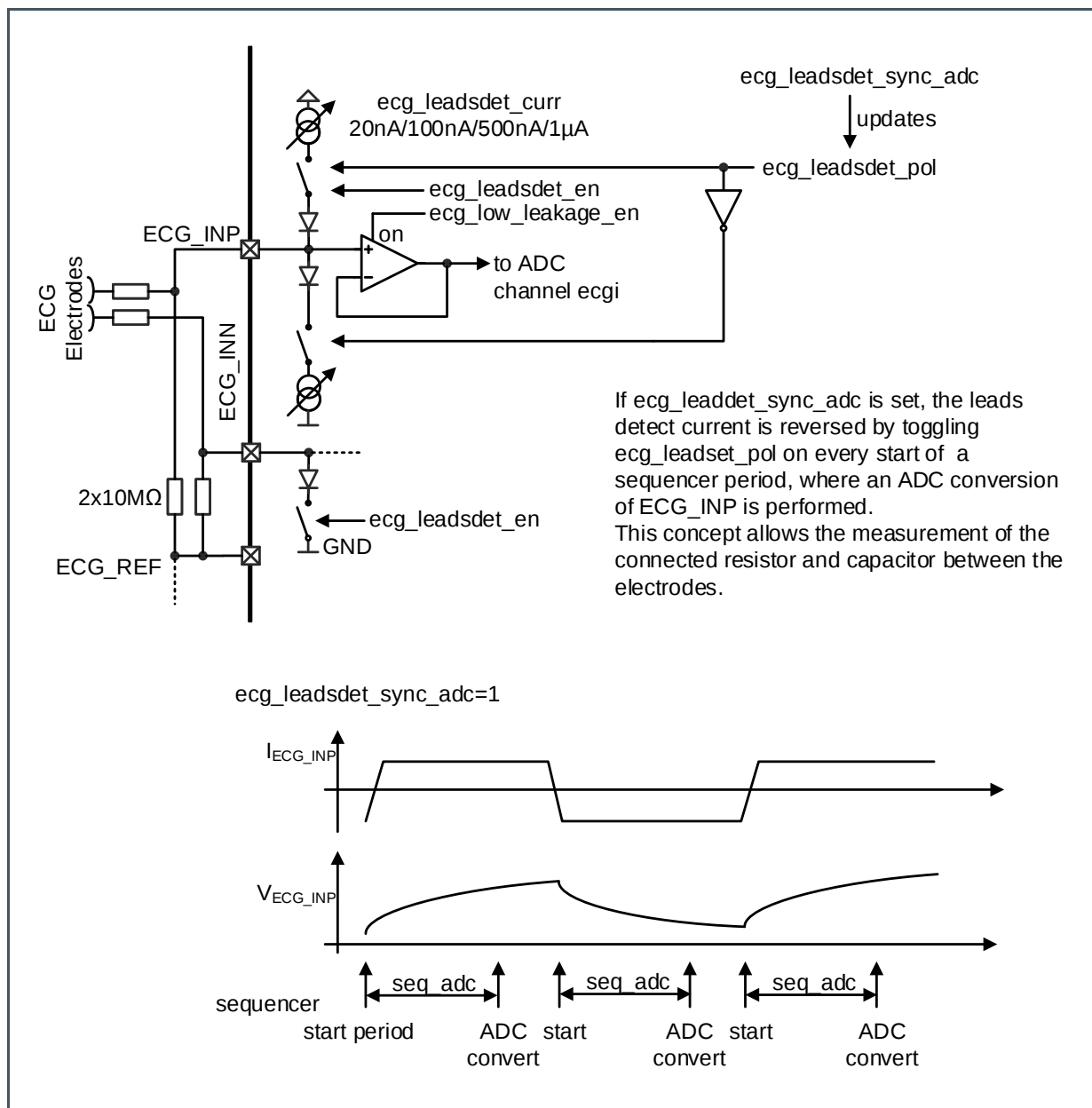
Figure 102:
ECG Amplifier



The ECG (electro cardiogram) amplifier is a high impedance, low noise instrumentation amplifier with analog circuitry to bandpass filter the signal and amplify it before converting it with the ADC.

ECG Lead OFF Detection

Figure 103:
ECG Lead OFF Detection Internal Circuit



The ECG lead OFF detection can be used for detection if the user actually touches the leads. It is a circuitry to measure the capacitor and/or resistance between the two lead inputs ECG_INP and ECG_INN.

7.1.16 ECG Registers

ECG_CFGA (Address 0x5c)

Figure 104:
ECG_CFGA Register

Addr: 0x5c		ECG_CFGA		
Bit	Bit Name	Default	Access	Bit Description
7	ecg_en	0	RW	Enable ECG instrumentation amplifier
6	not used	0	RW	Do not rely on the content of this register
5	ecg_lp_en	0	RW	0 ... LP filter disabled 1 ... LP filter enabled
4	ecg_hp_en	0	RW	0 ... Power down of the high pass filter 1 ... Enable high pass filter
3	ecg_gain_en	0	RW	0 ... Power down of the Gain stage 1 ... Enable Gain stage
2	ecg_lp_byp	0	RW	0 ... LP stage is used 1 ... LP stage is bypassed
1	ecg_hp_byp	0	RW	0 ... HP filter is used 1 ... HP filter is bypassed
0	ecg_gain_byp	0	RW	0 ... Gain stage is used 1 ... Gain stage is bypassed

ECG_CFGB (Address 0x5d)

Figure 105:
ECG_CFGB Register

Addr: 0x5d		ECG_CFGB												
Bit	Bit Name	Default	Access	Bit Description										
7	Not used	0	RW	Do not used										
				ECG low pass filter										
				<table><tr><th>Setting</th><th>Frequency</th></tr><tr><td>0</td><td>40 Hz</td></tr><tr><td>1</td><td>80 Hz</td></tr><tr><td>3</td><td>160 Hz</td></tr><tr><td>4</td><td>320 Hz</td></tr></table>	Setting	Frequency	0	40 Hz	1	80 Hz	3	160 Hz	4	320 Hz
Setting	Frequency													
0	40 Hz													
1	80 Hz													
3	160 Hz													
4	320 Hz													
6:5	ecg_lp_freq	0	RW											

Addr: 0x5d		ECG_CFGB				
Bit	Bit Name	Default	Access	Bit Description		
4:3				High pass filter cutoff frequency		
				Setting	Filter Frequency	Cutoff Frequency
				0	122 Hz	0.33 Hz
				1	488 Hz	1.32 Hz
				2	1935 Hz	5.28 Hz
				3	3906 Hz	10.56 Hz
				Gain		
				Setting	Gain	
				0	1	
				1	2	
2:0				2	4	
				3	8	
				4	16	
				5	32	
				6	64	
				7	128	

ECG_CFGC (Address 0x5e)

Figure 106:
ECG_CFGC Register

Addr: 0x5e		ECG_CFGC		
Bit	Bit Name	Default	Access	Bit Description
7:2	Not used	0	RW	Do not use
1	ecg_low_leakage_en	0	RW	Enable ECG leakage compensation
0	ecg_ref_en	0	RW	ECG Reference Feedback Amplifier Enable

ECG_CFGD (Address 0x5f)

Figure 107:
ECG_CFGD Register

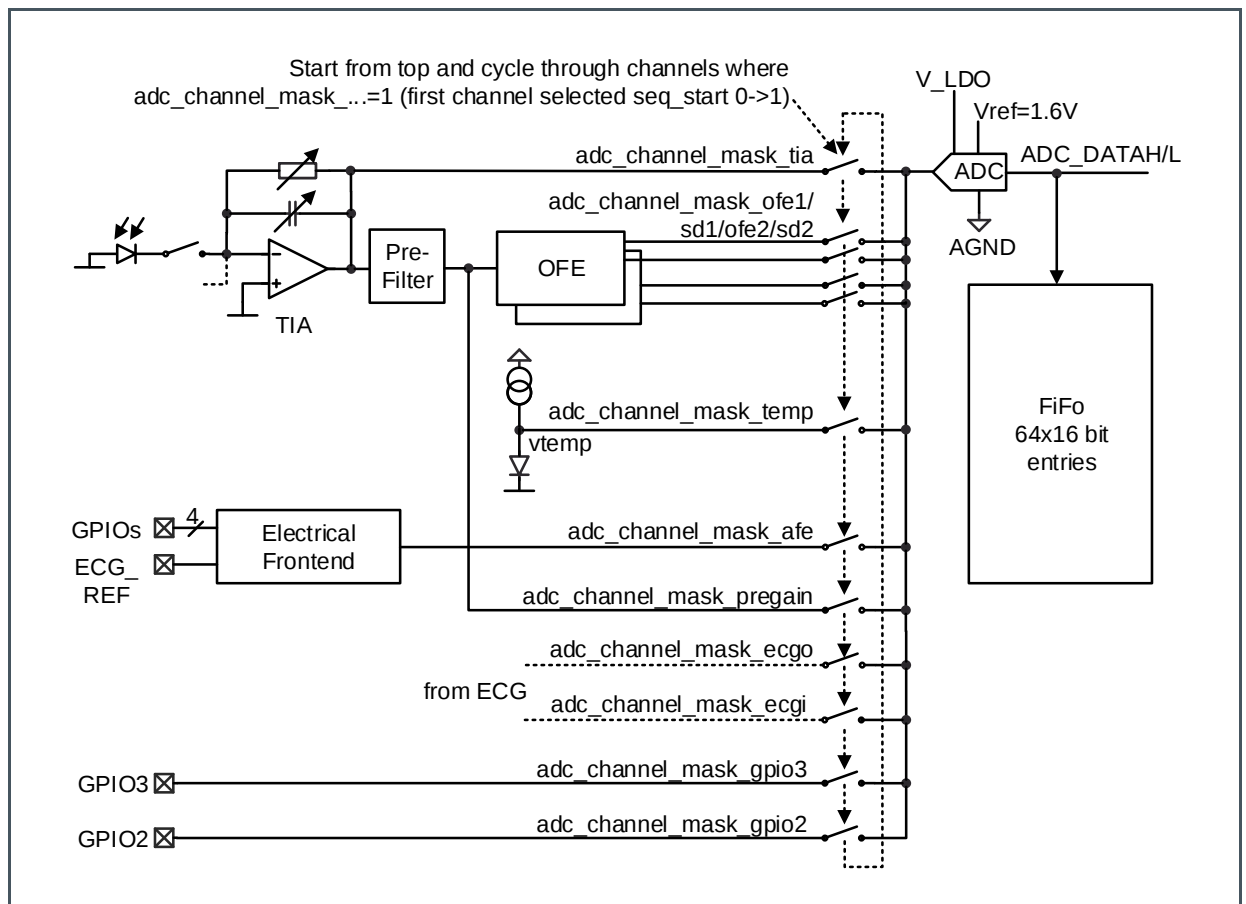
Addr: 0x5f		ECG_CFGD			
Bit	Bit Name	Default	Access	Bit Description	
7:5	Not used	0	RW	Do not used	
4	ecg_leadsdet_sync_adc	0	RW	ECG Leads Detection Automatic Update. If this is asserted, then ecg_leadsdet_pol is inverted automatically at the start of a sequence (at count=2) if in this sequence the ADC will convert the ECGi channel.	
3	ecg_leadsdet_pol	0	RW	ECG Leads Detection Polarity. Can be written to manually if ecg_leadsdet_sync_adc is clear, otherwise it is automatically toggled.	
2:1	ecg_leadsdet_curr	0	RW	ECG Leads Detection Current	
				Setting	Current
				0	20 nA
				1	100 nA
				2	500 nA
				3	1 μA
0	ecg_leadsdet_en	0	RW	ECG Leads Detection Enable	

7.1.17 ADC and FIFO

The ADC is a 14-bit successive-approximation register (SAR) type. It supports 14-bit with conversion time up to 50 ksp/s.

The ADC is started by the sequencer and its timing or in manual mode (man_mode=1) by setting seq_start=1 (seq_start stays '1' as long as the conversion runs). The AS7026GG can be configured to trigger an interrupt upon end of conversion.

Figure 108:
ADC Internal Circuit and Multiplexer



For best accuracy, the ADC can be optionally calibrated.



Information

If GPIO2 or GPIO3 is used as ADC input, there is no anti-aliasing filter in front of the ADC (needs to be added externally).

7.1.18 ADC Threshold

At the output of the ADC converter a digital threshold can be enabled. If the output of the ADC exceeds the threshold `adc_threshold`, it triggers an interrupt. This mechanism can be used to identify if an object is in proximity of the sensor and then to interrupt the host. In cases where no object is detected, the host can be sleeping therefore reducing power consumption of the system.

For detailed description of the threshold calculation see the register `ADC_THRESHOLD` and `ADC_THRESHOLD_CFG` description

7.1.19 ADC Registers

ADC_THRESHOLD (Address 0x68)

Figure 109:
ADC_THRESHOLD Register

Addr: 0x68		ADC_THRESHOLD		
Bit	Bit Name	Default	Access	Bit Description
7:0	adc_threshold	0xff	RW	If the ADC returns a value above adc_threshold (not equal), then the adc_threshold interrupt can be triggered. Note that when only the upper 8 bits are compared, the lower 6 bits are ignored. A value of 0xff can therefore never trigger the interrupt

ADC_THRESHOLD_CFG (Address 0x69)

Figure 110:
ADC_THRESHOLD_CFG Register

Addr: 0x69		ADC_THRESHOLD_CFG		
Bit	Bit Name	Default	Access	Bit Description
7:2	Not Used	0	RW	Not used
1	adc_thresh_differential	0	RW	If adc_thresh_tiaonly is asserted and any of seq_adc[23]tia is non-zero, meaning that there are two or three ADC TIA measurements in one sequencer period, then the second is subtracted from the first, and the <i>difference</i> is being compared to the adc_threshold.
0	adc_thresh_tiaonly	0	RW	Normally, the adc_threshold works regardless of the adc channel. If this bit is set, then the threshold is only checked if the adc channel is TIA

ADC_CFGA (Address 0x88)**Figure 111:**
ADC_CFGA Register

Addr: 0x88		ADC_CFGA																				
Bit	Bit Name	Default	Access	Bit Description																		
7:4	Not Used	0	RW	Not used																		
3:1	adc_multi_n	0	RW	Defines number of samples that are taken in multimode (adc_multimode =1)																		
				<table><tr><th>Setting</th><th>Number of Samples per ADC Conversion Command</th></tr><tr><td>0</td><td>2</td></tr><tr><td>1</td><td>4</td></tr><tr><td>2</td><td>8</td></tr><tr><td>3</td><td>16</td></tr><tr><td>4</td><td>32</td></tr><tr><td>5</td><td>48</td></tr><tr><td>6</td><td>64</td></tr><tr><td>7</td><td>96</td></tr></table>	Setting	Number of Samples per ADC Conversion Command	0	2	1	4	2	8	3	16	4	32	5	48	6	64	7	96
Setting	Number of Samples per ADC Conversion Command																					
0	2																					
1	4																					
2	8																					
3	16																					
4	32																					
5	48																					
6	64																					
7	96																					
0	adc_multimode	0	RW	0 ... If ADC is started one sample is measured 1 ... If ADC is started multiple samples are stored in sequence in the FIFO. The number of samples is defined with "adc_multi_n".																		

**Information**

If the ADC is triggered with the sequencer, the very first ADC conversion after seq_en=1 stores the number of samples according to above table. All subsequent samples use one sample less (e.g. 7 instead of 8).

ADC_CFGB (Address 0x89)

Figure 112:
ADC_CFGB Register

Addr: 0x89		ADC_CFGB					
Bit	Bit Name	Default	Access	Bit Description			
7:6	Not Used	0	RW	Not used			
5:3	adc_clock	0	RW	ADC clock divider: The ADC clock is freely configurable			
				Setting	Periods	µs	kHz
				0	2	1	1000
				1	4	2	500
				2	6	3	333
				3	8	4	250
				4	10	5	200
				5	12	6	167
				6	14	7	143
				7	16	8	125
2	adc_calibration	0	RW	To activate the optional self calibration, this bit must be asserted, and an ADC “conversion” has to be started in manual mode (man_mode=1) by asserting seq_start.			
1	ulp	0	RW	Ultra low power bit for the sequencer. If this bit is set and sd_subs>0, it disables the LED pulses and powers off the TIA in all sequences but the one where the TIA is sampled.			
0	adc_en	0	RW	0 ... Reset ADC 1 ... Enable ADC Warning: In reset state the ADC clears its calibration data. Re-calibration is necessary next time it is enabled again.			

ADC_CFGC (Address 0x8a)

Figure 113:
ADC_CFGC Register

Addr: 0x8a		ADC_CFGC		
Bit	Bit Name	Default	Access	Bit Description
7:5	Not Used	0	RW	Not used

Addr: 0x8a		ADC_CFGC					
Bit	Bit Name	Default	Access	Bit Description			
4	adc_selfpd	0	RW	1 ... Power down the ADC when not converting; use this to conserve power, but set adc_settling_time to minimum 64us to permit settling of the ADC reference buffer. 0 ... Always enable ADC			
3	adc_discharge	0	RW	0: Suppress ADC capacitor discharging – use with caution 1: Discharge ADC capacitor before tracking If asserted, the capacitor is discharged before the tracking phase. If zero, the discharge phase is suppressed and the tracking phase is started one cycle earlier			
2:0	adc_settling_time	0	RW	ADC settling time: Use with synchronous demodulator. It defines the number of ADC clock cycles the sampling window is kept open additionally. If the gain stage in the optical frontend is used (gain_byp=0), set this to minimum 8 μs. If adc_selfpd=1, set this to minimum 64 μs.			
				Setting	Periods	μs (@500 kHz)	μs (@250 kHz)
				0	0	0	0
				1	4	8	16
				2	8	16	32
				3	16	32	64
				4	32	64	128
				5	64	128	256
				6	128	256	512
7	256	512	1 ms				

ADC_CHANNEL_MASK_L (Address 0x8b)

Figure 114:
ADC_CHANNEL_MASK_L Register

Addr: 0x8b		ADC_CHANNEL_MASK_L		
Bit	Bit Name	Default	Access	Bit Description
7	adc_channel_mask_pregain	0	RW	Pregain channel selection
6	adc_channel_mask_afe	0	RW	Electrical front end

Addr: 0x8b		ADC_CHANNEL_MASK_L		
Bit	Bit Name	Default	Access	Bit Description
5	adc_channel_mask_temp	0	RW	Temperature measurement
4	adc_channel_mask_sd2	0	RW	Synchronous modulator 2 output just before the gain stage
3	adc_channel_mask_ofe2	0	RW	Synchronous modulator 2 output after the gain stage
2	adc_channel_mask_sd1	0	RW	Synchronous modulator 1 output just before the gain stage
1	adc_channel_mask_ofe1	0	RW	Synchronous modulator 1 output after the gain stage
0	adc_channel_mask_tia	0	RW	Transimpedance amplifier output

The adc channel is chosen automatically from the bits within the adc_channel_mask_* set. It starts from right and finishes left (LSB->MSB) and wraps back from the most significant asserted bit to the least significant of the asserted bits. After every ADC conversion it switches to the next enabled channel, (except around the adc2tia/adc3tia cases). See register description FIFOH and FIFOL for encoding of the first channel in the data stream.

This applies to both, manual mode and sequencer mode. In sequencer mode, it starts with the smallest channel when the sequencer is being started. In manual mode, the adc_sel is reset with every write to either ADC_CHANNEL_MASK_L or ADC_CHANNEL_MASK_H.

ADC_CHANNEL_MASK_H (Address 0x8c)

Figure 115:
ADC_CHANNEL_MASK_H Register

Addr: 0x8c		ADC_CHANNEL_MASK_H		
Bit	Bit Name	Default	Access	Bit Description
7:4	Not used	0	RW	Not used
3	adc_channel_mask_gpio2	0	RW	GPIO2 input – set gpio2_a=1 and Write 0x47 to register 0xC6 Write 0x0C to register 0xC2 Write 0x0C to register 0xC3
2	adc_channel_mask_gpio3	0	RW	GPIO3 input – set gpio3_a=1 and Write 0x47 to register 0xC6 Write 0x0C to register 0xC2 Write 0x0C to register 0xC3
1	adc_channel_mask_ecgi	0	RW	ECG amplifier input – use for leads off detection

Addr: 0x8c		ADC_CHANNEL_MASK_H		
Bit	Bit Name	Default	Access	Bit Description
0	adc_channel_mask_ecgo	0	RW	ECG amplifier output – amplified ECG signal

ADC_DATA_L (Address 0x8e)

Figure 116:
ADC_DATA_L Register

Addr: 0x8e		ADC_DATA_L		
Bit	Bit Name	Default	Access	Bit Description
7:0	adc_data[7:0]	0	RO	Current ADC output: low byte

The ADC_DATA register shows the current raw output of the ADC.

ADC_DATA_H (Address 0x8f)

Figure 117:
ADC_DATA_H Register

Addr: 0x8f		ADC_DATA_H		
Bit	Bit Name	Default	Access	Bit Description
7:6	Not used	0	RO	Not used
5:0	adc_data[13:8]	0	RO	Current ADC output: high byte warning: there is no latch mechanism implemented to guarantee consistency if the ADC is possibly running when reading this register, then the data can be corrupted - use the FIFO to guarantee data consistency

7.1.20 FIFO Registers

FIFO_CFG (Address 0x78)

Figure 118:
FIFO_CFG Register

Addr: 0x78		FIFO_CFG		
Bit	Bit Name	Default	Access	Bit Description
7:6	Not used	0	RW	Not used
5:0	fifo_threshold	0	RW	FIFO threshold. The fifo_threshold interrupt is flagged if there are more than this many entries in the FIFO. 0 ... Interrupt with 1 (16-bit) entry in FIFO 63 ... Interrupt when FIFO is full but one; note that the FIFO is 64 entries deep

FIFO_CNTRL (Address 0x79)

Figure 119:
FIFO_CNTRL Register

Addr: 0x79		FIFO_CNTRL		
Bit	Bit Name	Default	Access	Bit Description
7:1	Not used	0	RW	Not used
0	fifo_clear	0	PUSH1	Write a 1 here to clear the FIFO. Can be useful when switching from one sequencer mode to another to make sure that there are no old FIFO entries left

FIFOSTATUS (Address 0xa3)

Figure 120:
FIFOSTATUS Register

Addr: 0xa3		FIFOSTATUS		
Bit	Bit Name	Default	Access	Bit Description
7	fifooverflow	0	RO	FIFO overflow indicator
6:0	fifolevel	0	RO	FIFO fill level (0 ...64)

FIFOL (Address 0xfe)

Figure 121:
FIFOL Register

Addr: 0xfe		FIFOL		
Bit	Bit Name	Default	Access	Bit Description
7:0	fifol	0	PUSHPOP	Low byte of FIFO

FIFOL can be read out with single reads (2 consecutive I²C addresses have to be read to get one FIFO entry) or with block-read (up to 2 x fifo_depth values can be read in a single block-read)

Upon reading of FIFOH, it automatically advances the internal read pointer and decreases FIFO level. If reading beyond end of FIFO, data will return 00h. There is no underrun flag, this is not an error condition.

Use **ams** SDK functions to read from the FIFO register to keep the reading in synchronization with the ADC channel selection. If synchronization is no concern use [fifoh[7:0] : fifol[7:2]] as ADC result as the ADC data is multiplied by x4 before it is pushed in to the FIFO. FIFOL[0] is used as an ADC first channel indication. The first channel indication bit toggles upon every new entry unless the first ADC channel is transmitted. Then toggling can be stopped for up to 5 FIFO entries and the very first stopping indicates the first ADC channel. To allow encoding of any number of ADC channels, the first ADC channel encoding is dropped from time to time.

FIFOH (Address 0xff)

Figure 122:
FIFOH Register

Addr: 0xff		FIFOH		
Bit	Bit Name	Default	Access	Bit Description
7:0	fifoh	0	PUSHPOP	High byte of FIFO

See Interrupts for the actual FIFO interrupt.

7.1.21 Digital Interface

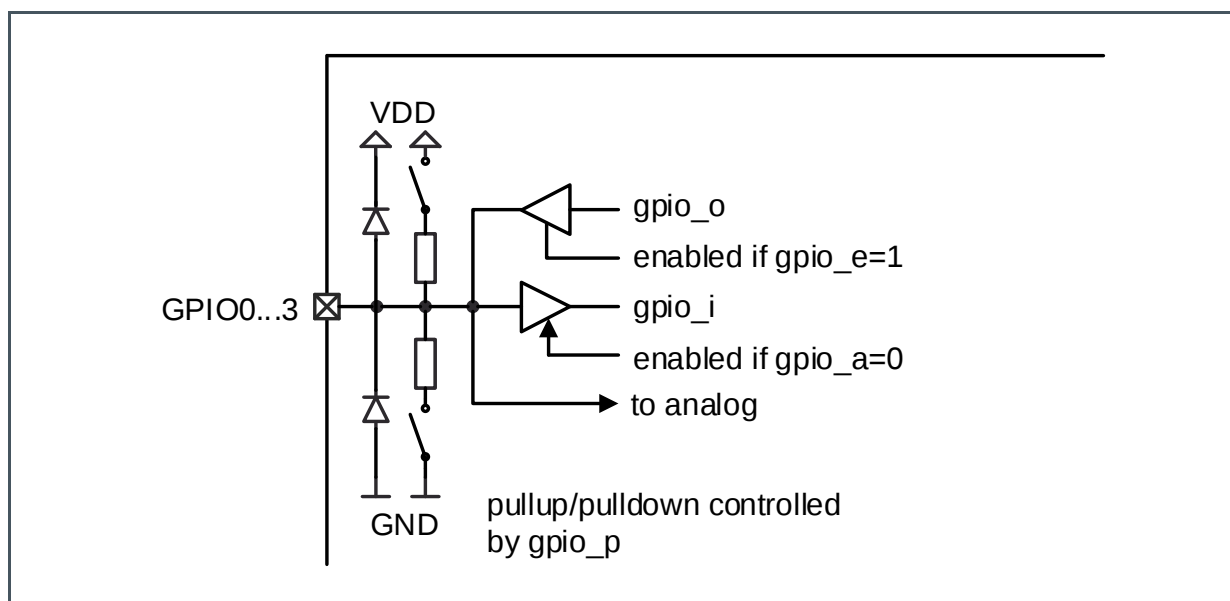
Power Management

After setting the pin ENABLE=1 the AS7026GG registers can be accessed by the I²C interface. Before enabling any additional function (current source, TIA, ADC...) set the bit ldo_en=1 to set the internal LDO to normal mode.

For operating the ADC or the sequencer enable the oscillator by setting osc_en=1

GPIO Pins

Figure 123:
GPIO Internal Circuit



Interrupts

An interrupt output pin INT can be used to interrupt the host. Following interrupt sources are possible:

irq_adc: End of ADC conversion

irq_sequencer: End of sequencer sequence reached.

irq_ltf: A light-to-frequency conversion is finished.

irq_adc_threshold: ADC threshold triggered – see ADC Threshold.

irq_fifothreshold: FIFO almost full (as defined in bit fifo_threshold)

irq_fifooverflow: FIFO overflow (error condition, data is lost)

irq_clipdetect: TIA output and/or SD output exceeded threshold– see details in CLIPSTATUS

irq_led_supply_low: led supply low comparator triggered – see details in LEDSTATUS

Depending on the setting in register INTENAB each of the above interrupt source can assert INT output pin (active low).

7.2 I²C

The AS7026GG includes an I²C slave using an I²C address of 0x30 (7-bit format; R/W bit has to be added) respectively 60 h (8-bit format for writing) and 61 h (8-bit format for reading). It expects external pullup resistors.

7.2.1 I²C Serial Control Interface

I²C Feature List

- Fast mode (400 kHz) and standard mode (100 kHz) support
- 7+1-bit addressing mode
- Write formats: Single-Byte-Write, Page-Write
- Read formats: Current-Address-Read, Random-Read, Sequential-Read
- SDA input delay and SCL spike filtering by integrated RC-components

I²C Protocol

Figure 124:
I²C Symbol Definition

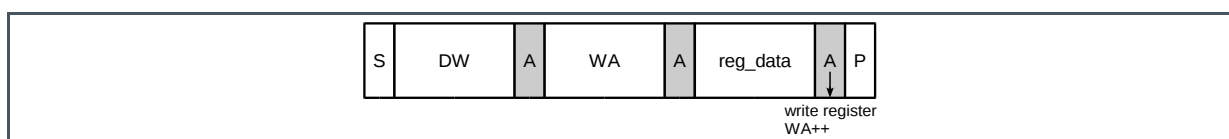
Symbol	Definition	RW	Note
S	Start condition after stop	R	1-bit
Sr	Repeated start	R	1-bit
DW	Device address for write	R	0110 0000b (60 h)
DR	Device address for read	R	0110 0001b (61 h)
WA	Word address	R	8-bit
A	Acknowledge	W	1-bit
N	No Acknowledge	R	1-bit
reg_data	Register data/write	R	8-bit
data (n)	Register data/read	W	8-bit
P	Stop condition	R	1-bit
WA++	Increment word address internally	R	During acknowledge

I²C Symbol Definition: Shows the symbols used in the following mode descriptions.

I²C Write Access

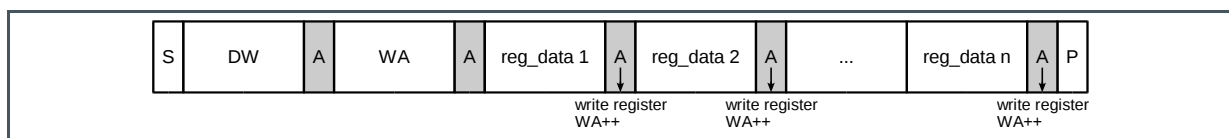
Byte Write and Page Write formats are used to write data to the slave

Figure 125:
I²C Byte Write



I²C Byte Write: Shows the format of an I²C byte write access

Figure 126:
I²C Page Write

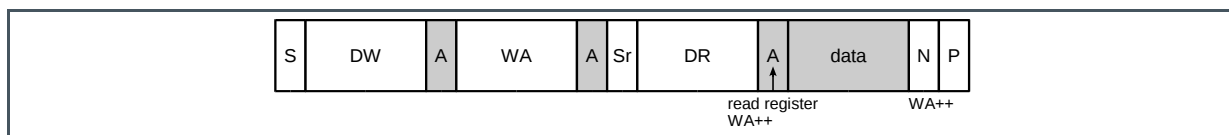


I²C Page Write: Shows the format of an I²C page write access

The transmission begins with the START condition, which is generated by the master when the bus is in IDLE state (the bus is free). The device-write address is followed by the word address. After the word address any number of data bytes can be sent to the slave. The word address is incremented internally, in order to write subsequent data bytes on subsequent address locations.

For reading data from the slave device, the master has to change the transfer direction. This can be done either with a repeated START condition followed by the device-read address, or simply with a new transmission START followed by the device-read address, when the bus is in IDLE state. The device-read address is always followed by the 1st register byte transmitted from the slave. In Read mode any number of subsequent register bytes can be read from the slave. The word address is incremented internally.

Figure 127:
I²C Random Read



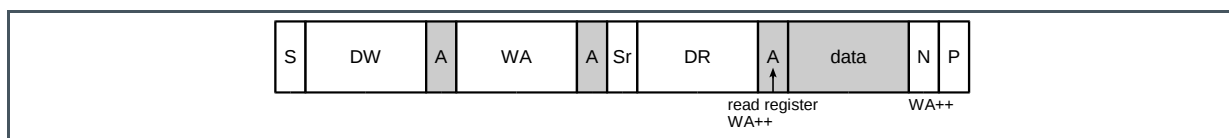
I²C Random Read: Shows the format of an I²C random read access

Random Read and Sequential Read are combined formats. The repeated START condition is used to change the direction after the data transfer from the master.

The word address transfer is initiated with a START condition issued by the master while the bus is idle. The START condition is followed by the device-write address and the word address.

In order to change the data direction a repeated START condition is issued on the 1st SCL pulse after the acknowledge bit of the word address transfer. After the reception of the device-read address, the slave becomes the transmitter. In this state the slave transmits register data located by the previous received word address vector. The master responds to the data byte with a not-acknowledge, and issues a STOP condition on the bus.

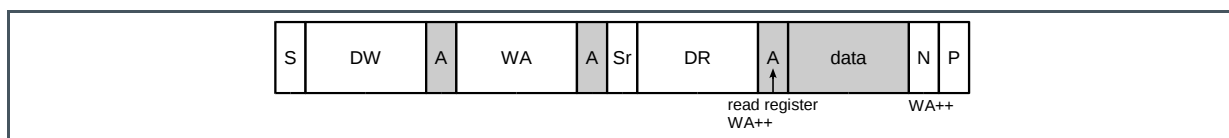
Figure 128:
I²C Sequential Read



I²C Sequential Read: Shows the format of an I²C sequential read access

Sequential Read is the extended form of Random Read, as more than one register-data bytes are transferred subsequently. In difference to the Random Read, for a sequential read the transferred register-data bytes are responded by an acknowledge from the master. The number of data bytes transferred in one sequence is unlimited (consider the behavior of the word-address counter). To terminate the transmission the master has to send a not-acknowledge following the last data byte and generate the STOP condition subsequently.

Figure 129:
I²C Current Address Read



I²C Current Address Read: Shows the format of an I²C current address read access.

To keep the access time as small as possible, this format allows a read access without the word address transfer in advance to the data transfer. The bus is idle and the master issues a START condition followed by the Device-Read address. Analogous to Random Read, a single byte transfer is terminated with a not-acknowledge after the 1st register byte. Analogous to Sequential Read an unlimited number of data bytes can be transferred, where the data bytes has to be responded with an acknowledge from the master. For termination of the transmission the master sends a not-acknowledge following the last data byte and a subsequent STOP condition.

CONTROL (Address 0x00)

Figure 130:
CONTROL Register

Addr: 0x00		CONTROL			
Bit	Bit Name	Default	Access	Bit Description	
7:2	Not Used	0	RW	Not Used	

Addr: 0x00		CONTROL		
Bit	Bit Name	Default	Access	Bit Description
1	osc_en	0	RW	Enable the oscillator. The oscillator must be enabled for any analog block (ADC, sequencer, optical frontend, sequencer); not mandatory for current sinks or ECG amplifier
0	ldo_en	0	RW	If the EN input is not asserted, the chip is in reset If asserted, I ² C transactions are possible. Upon assertion of ldo_en, the reference and the LDO are enabled The LDO must be enabled for anything but plain I ² C register read/write

GPIO_A (Address 0x08)

Figure 131:
GPIO_A Register

Addr: 0x08		GPIO_A		
Bit	Bit Name	Default	Access	Bit Description
7:4	Not Used	0	RW	Not Used
3	gpio3_a	0	RW	1=Put GPIO3 in analog mode; set this bit when used for an analog function e.g. the electrical frontend. If set execute following I ² C commands (otherwise an internal pulldown will be enabled) in this sequence: Write 0x47 to register 0xC6 Write 0x0C to register 0xC2 Write 0x0C to register 0xC3
2	gpio2_a	0	RW	1=Put GPIO2 in analog mode If set execute following I ² C commands (otherwise an internal pulldown will be enabled) in this sequence: Write 0x47 to register 0xC6 Write 0x0C to register 0xC2 Write 0x0C to register 0xC3
1	gpio1_a	0	RW	1=Put GPIO1 in analog mode
0	gpio0_a	0	RW	1=Put GPIO0 in analog mode

GPIO_E (Address 0x09)**Figure 132:**
GPIO_E Register

Addr: 0x09		GPIO_E		
Bit	Bit Name	Default	Access	Bit Description
7:4	Not Used	0	RW	Not Used
3	gpio3_e	0	RW	GPIO3 output enabled if set
2	gpio2_e	0	RW	GPIO2 output enabled if set
1	gpio1_e	0	RW	GPIO1 output enabled if set
0	gpio0_e	0	RW	GPIO0 output enabled if set

GPIO_O (Address 0x0a)**Figure 133:**
GPIO_O Register

Addr: 0x0a		GPIO_O		
Bit	Bit Name	Default	Access	Bit Description
7:4	Not Used	0	RW	Not Used
3	gpio3_o	0	RW	If gpio3_e=1, gpio3_o defines the output state of GPIO3
2	gpio2_o	0	RW	If gpio2_e=1, gpio2_o defines the output state of GPIO2
1	gpio1_o	0	RW	If gpio1_e=1, gpio1_o defines the output state of GPIO1
0	gpio0_o	0	RW	If gpio0_e=1, gpio0_o defines the output state of GPIO0

GPIO_I (Address 0x0b)**Figure 134:**
GPIO_I Register

Addr: 0x0b		GPIO_I		
Bit	Bit Name	Default	Access	Bit Description
7:4	Not Used	0	RO	Not Used

Addr: 0x0b		GPIO_I		
Bit	Bit Name	Default	Access	Bit Description
3	gpio3_i	0	RO	The digital value sensed on GPIO3
2	gpio2_i	0	RO	The digital value sensed on GPIO2
1	gpio1_i	0	RO	The digital value sensed on GPIO1
0	gpio0_i	0	RO	The digital value sensed on GPIO0

GPIO_P (Address 0x0c)

Figure 135:
GPIO_P Register

Addr: 0x0c		GPIO_P		
Bit	Bit Name	Default	Access	Bit Description
7	gpio3_pd	0	RW	GPIO3 pulldown configuration 0: No pulldown on GPIO3 1: Pulldown to GND on GPIO3
6	gpio3_pu	0	RW	GPIO3 pullup configuration 0: No pullup on GPIO3 1: Pullup to VDD on GPIO3
5	gpio2_pd	0	RW	GPIO2 pulldown configuration
4	gpio2_pu	0	RW	GPIO2 pullup configuration
3	gpio1_pd	0	RW	GPIO1 pulldown configuration
2	gpio1_pu	0	RW	GPIO1 pullup configuration
1	gpio0_pd	0	RW	GPIO0 pulldown configuration
0	gpio0_pu	0	RW	GPIO0 pullup configuration

GPIO_SR (Address 0x0d)

Figure 136:
GPIO_SR Register

Addr: 0x0c		GPIO_SR		
Bit	Bit Name	Default	Access	Bit Description
7:4	Not Used	0	RW	Not Used

Addr: 0x0c		GPIO_SR		
Bit	Bit Name	Default	Access	Bit Description
3	gpio3_sr	0	RW	GPIO3 slew rate configuration 0: Default slew rate 1: Increased slew rate
2	gpio2_sr	0	RW	GPIO2 slew rate configuration
1	gpio1_sr	0	RW	GPIO1 slew rate configuration
0	gpio0_sr	0	RW	GPIO0 slew rate configuration

SUBID (Address 0x91)

Figure 137:
SUBID Register

Addr: 0x91		SUBID		
Bit	Bit Name	Default	Access	Bit Description
7:3	subid	NA	RO	Defines product version. Do not rely on bits defined as 'X'. 1XXXXb
2:0	Revision	NA	RO	Reserved. Do no use and do not rely that the content stays the same for each device.

ID (Address 0x92)

Figure 138:
ID Register

Addr: 0x92		ID			
Bit	Bit Name	Default	Access	Bit Description	
7:2	id	0	RO	Part Number Identification.	
				Value	Meaning
				110011	AS7026GG
1:0	id_reserved	0	RO	Reserved. Do no use and do not rely that the content stays the same for each device	

STATUS (Address 0xa0)**Figure 139:**
STATUS Register

Addr: 0xa0		STATUS		
Bit	Bit Name	Default	Access	Bit Description
7	irq_led_supply_low	0	R_PUSH1	Check LEDSTATUS
6	irq_clipdetect	0	R_PUSH1	Check CLIPSTATUS
5	irq_fifooverflow	0	R_PUSH1	FIFO overflow (error condition, new data is lost)
4	irq_fifothreshold	0	R_PUSH1	FIFO is almost full (as defined in fifo_threshold, usually 3/4)
3	irq_adc_threshold	0	R_PUSH1	The ADC value was above the programmed adc_threshold register setting
2	irq_ltf	0	R_PUSH1	LTF measurement is done. check LTFSTATUS (or ignore it)
1	irq_sequencer	0	R_PUSH1	All configured sequencer iterations have finished
0	irq_adc	0	R_PUSH1	ADC has finished

The STATUS register shows the current state of the interface. Some bits in here can trigger an interrupt.

An asserted bit can be cleared by writing a '1' to it - in case of irq_led_supply_low and irq_clipdetect, this also clears the underlying condition in the CLIPSTATUS and LEDSTATUS registers.

The FIFO threshold interrupt cannot be cleared directly, but only by lowering the FIFO level. The FIFO overflow interrupt is sticky and must be cleared explicitly.

CLIPSTATUS (Address 0xa1)**Figure 140:**
CLIPSTATUS Register

Addr: 0xa1		CLIPSTATUS		
Bit	Bit Name	Default	Access	Bit Description
7:4	Not used	0	RO	Not Used
3	pd_clipdetect_l			If this bit is asserted, photo diode amplifier has been below the lower threshold

Addr: 0xa1		CLIPSTATUS		
Bit	Bit Name	Default	Access	Bit Description
2	pd_clipdetect_h			If this bit is asserted, photo diode amplifier has been above the upper threshold
1	sd_clipdetect_l			If this bit is asserted, photo diode amplifier has been below the lower threshold
0	sd_clipdetect_h			If this bit is asserted, photo diode amplifier has been above the upper threshold

LEDSTATUS (Address 0xa2)

Figure 141:
LEDSTATUS Register

Addr: 0xa2		LEDSTATUS		
Bit	Bit Name	Default	Access	Bit Description
7:4	Not used	0	RO	Not Used
3	led4_supply_low	0	RO	If this bit is asserted, LED4 voltage has been too low
2	led3_supply_low	0	RO	If this bit is asserted, LED3 voltage has been too low
1	led2_supply_low	0	RO	If this bit is asserted, LED3 voltage has been too low
0	led1_supply_low	0	RO	If this bit is asserted, LED1 voltage has been too low

INTENAB (Address 0xa8)

Figure 142:
INTENAB Register

Addr: 0xa8		INTENAB		
Bit	Bit Name	Default	Access	Bit Description
7	irq_led_supply_low_enab	0	RW	1 ... Enable led supply low interrupt
6	irq_clipdetect_enab	0	RW	1 ... Enable clipdetect interrupt
5	irq_fifooverflow_ena	0	RW	1 ... Enable fifooverflow interrupt

Addr: 0xa8		INTENAB		
Bit	Bit Name	Default	Access	Bit Description
4	irq_fifothreshold_enab	0	RW	1 ... Enable fifothreshold interrupt
3	irq_adc_threshold_enab	0	RW	1 ... Enable irq_adc_threshold as an interrupt source
2	irq_ltf_enab	0	RW	1 ... Enable LTF as an interrupt source
1	irq_sequencer_enab	0	RW	1 ... Enable irq_sequencer as an interrupt source
0	irq_adc_enab	0	RW	1 ... Enable irq_adc as an interrupt source

Each of the STATUS register bits can cause an interrupt (register INTR) if the respective bit is asserted in the INTENAB register

INTR (Address 0xa9)

Figure 143:
INTR Register

Addr: 0xa9		INTR		
Bit	Bit Name	Default	Access	Bit Description
7	irq_led_supply_low_intr	0	RO	
6	irq_clipdetect_intr	0	RO	
5	irq_fifooverflow_intr	0	RO	
4	irq_fifothreshold_intr	0	RO	
3	irq_adc_threshold_intr	0	RO	
2	irq_ltf_intr	0	RO	
1	irq_sequencer_intr	0	RO	
0	irq_adc_intr	0	RO	

The INTR registers shows the bit or bits that are responsible for an asserted interrupt. Effectively, these bits are OR-ed together to drive the interrupt pin INT low (open drain output).

8 Application Information

8.1 Application Examples

The following figure shows the complete integration of the AS7026GG in a mobile optical measurement system for HRM, SpO2, GSR (galvanic skin resistivity) and skin temperature using an NTC.

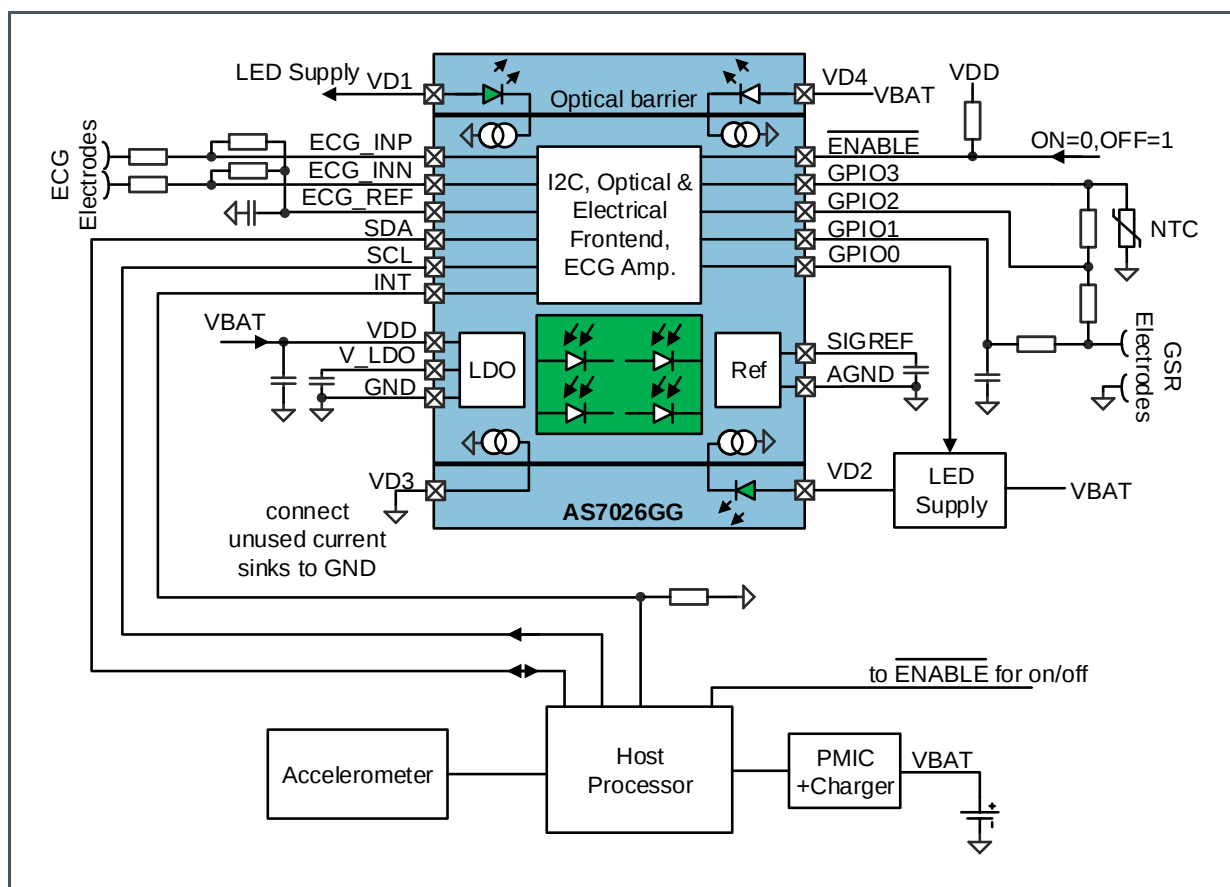
The device can be powered directly by a Li-ion battery as it has its own power management. Nevertheless the I²C interface can be powered by 1.8 V circuitry.



Information

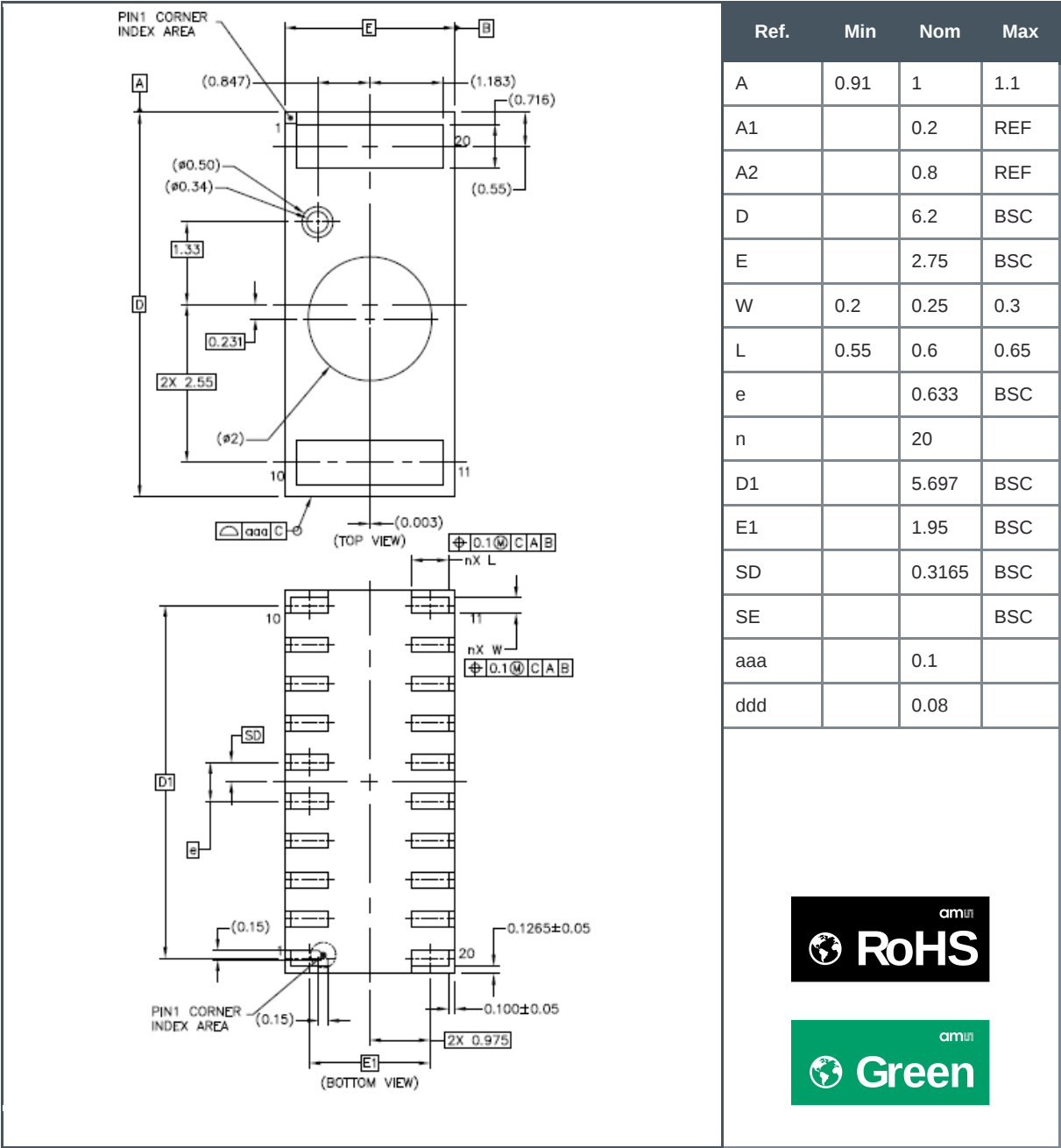
AS7026GG can be used in the same configuration for e.g. a fitness band or a smart watch.

Figure 144:
AS7026GG Optical HRM Measurement System for Wrist Based Application



9 Package Drawings & Markings

Figure 145:
AS7026GG Package Outline Drawing



- (1) All dimensions are in millimeters. Angles in degrees.
- (2) Dimensioning and tolerancing conform to ASME Y14.5M-1994.
- (3) n is the total number of terminals.
- (4) This package contains no lead (Pb).
- (5) This drawing is subject to change without notice.

11 Legal Information

Copyrights & Disclaimer

Copyright ams AG, Tobelbader Strasse 30, 8141 Premstaetten, Austria-Europe. Trademarks Registered. All rights reserved. The material herein may not be reproduced, adapted, merged, translated, stored, or used without the prior written consent of the copyright owner.

Devices sold by ams AG are covered by the warranty and patent indemnification provisions appearing in its General Terms of Trade. ams AG makes no warranty, express, statutory, implied, or by description regarding the information set forth herein. ams AG reserves the right to change specifications and prices at any time and without notice. Therefore, prior to designing this product into a system, it is necessary to check with ams AG for current information. This product is intended for use in commercial applications. Applications requiring extended temperature range, unusual environmental requirements, or high reliability applications, such as military, medical life-support or life-sustaining equipment are specifically not recommended without additional processing by ams AG for each application. This product is provided by ams AG "AS IS" and any express or implied warranties, including, but not limited to the implied warranties of merchantability and fitness for a particular purpose are disclaimed.

ams AG shall not be liable to recipient or any third party for any damages, including but not limited to personal injury, property damage, loss of profits, loss of use, interruption of business or indirect, special, incidental or consequential damages, of any kind, in connection with or arising out of the furnishing, performance or use of the technical data herein. No obligation or liability to recipient or any third party shall arise or flow out of ams AG rendering of technical or other services.

RoHS Compliant & ams Green Statement

RoHS Compliant: The term RoHS compliant means that ams AG products fully comply with current RoHS directives. Our semiconductor products do not contain any chemicals for all 6 substance categories, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, RoHS compliant products are suitable for use in specified lead-free processes.

ams Green (RoHS compliant and no Sb/Br): ams Green defines that in addition to RoHS compliance, our products are free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material).

Important Information: The information provided in this statement represents ams AG knowledge and belief as of the date that it is provided. ams AG bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. ams AG has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. ams AG and ams AG suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

Headquarters

ams AG
Tobelbader Strasse 30
8141 Premstaetten
Austria, Europe
Tel: +43 (0) 3136 500 0

Please visit our website at www.ams.com

Buy our products or get free samples online at www.ams.com/Products

Technical Support is available at www.ams.com/Technical-Support

Provide feedback about this document at www.ams.com/Document-Feedback

For sales offices, distributors and representatives go to www.ams.com/Contact

For further information and requests, e-mail us at ams_sales@ams.com